

LTC1540

with Reference

# FEATURES

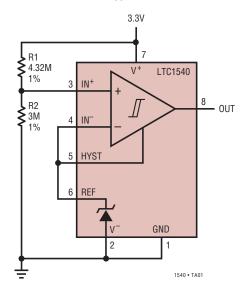
- Ultralow Quiescent Current: 0.3µA Typ
- Reference Output Drives 0.01µF Capacitor
- Adjustable Hysteresis
- Available in 3mm × 3mm × 0.8mm DFN Package
- Wide Supply Range: 2V to 11V
- Input Voltage Range Includes the Negative Supply
- Reference Output Sources Up to 1mA
- TTL/CMOS Compatible Outputs
- 60µs Propagation Delay with 10mV Overdrive
- No Crowbar Current
- 40mA Continuous Source Current
- Pin Compatible with LTC1440, MAX921, MAX931

### **APPLICATIONS**

- Battery-Powered System Monitoring
- Threshold Detectors
- Window Comparators
- Oscillator Circuits

### TYPICAL APPLICATION

Nanopower 2.9V V<sub>CC</sub> Threshold Detector



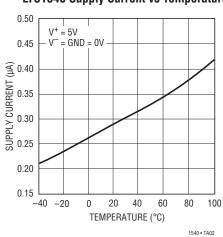
### DESCRIPTION

The LTC®1540 is an ultralow power, single comparator with built-in reference. The comparator's features include less than  $0.6\mu$ A supply current over the commercial temperature range, a  $1.182V \pm 2\%$  reference, programmable hysteresis and TTL/CMOS outputs that sink and source current. The reference output can drive a bypass capacitor of up to  $0.01\mu$ F without oscillation.

Nanopower Comparator

The comparator operates from a single 2V to 11V supply or a dual  $\pm$ 1V to  $\pm$ 5.5V supply. Comparator hysteresis is easily programmed by using two resistors and the HYST pin. Each comparator's input operates from the negative supply to within 1.3V of the positive supply. The comparator output stage can continuously source up to 40mA. By eliminating the cross-conducting current that normally occur when the comparator changes logic states, power supply glitches are eliminated.

The LTC1540 is available in the 8-pin MSOP and SO packages. For space limited applications, the LTC1540 is available in a  $3mm \times 3mm$  low profile (0.8mm) dual fine-pitch leadless package (DFN).



LTC1540 Supply Current vs Temperature

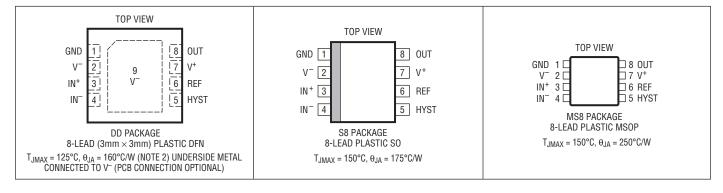
# ABSOLUTE MAXIMUM RATINGS (Note 1)

#### Voltage

V <sup>+</sup> to V <sup>-</sup> , V <sup>+</sup> to GND, GN	D to V <sup>-</sup> 12V to -0.3V
IN⁺, IN⁻, HYST	$(V^+ + 0.3V)$ to $(V^ 0.3V)$
REF	$(V^+ + 0.3V)$ to $(V^ 0.3V)$
	. (V <sup>+</sup> + 0.3V) to (GND - 0.3V)
Current	
IN+, IN <sup>-</sup> , HYST	20mA
REF	20mA
	50mA

OUT Short-Circuit Duration (V <sup>+</sup> $\leq$ 5.5V) Continuous Power Dissipation
Operating Temperature Range
LTC1540C 0°C to 70°C
LTC1540I40°C to 85°C
Storage Temperature Range –65°C to 150°C
(DD Package)65°C to 125°C
Lead Temperature (Soldering, 10 sec)

# PIN CONFIGURATION



## ORDER INFORMATION http://www.linear.com/product/LTC1540#orderinfo

LEAD FREE FINISH	TAPE AND REEL	*PART MARKING	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LTC1540CDD#PBF	LTC1540CDD#TRPBF	LAAS	8-Lead (3mm × 3mm) Plastic DFN	0°C to 70°C
LTC1540IDD#PBF	LTC1540IDD#TRPBF	LAAS	8-Lead (3mm × 3mm) Plastic DFN	-40°C to 85°C
LTC1540CS8#PBF	LTC1540CS8#TRPBF	1540	8-Lead Plastic SO	0°C to 70°C
LTC1540IS8#PBF	LTC1540IS8#TRPBF	15401	8-Lead Plastic SO	-40°C to 85°C
LTC1540CMS8#PBF	LTC1540CMS8#TRPBF	LTCE	8-Lead Plastic MSOP	0°C to 70°C
LTC1540IMS8#PBF	LTC1540IMS8#TRPBF	LTADV	8-Lead Plastic MSOP	-40°C to 85°C

Consult LTC Marketing for parts specified with wider operating temperature ranges. \*The temperature grade is identified by a label on the shipping container. For more information on lead free part marking, go to: http://www.linear.com/leadfree/

For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/. Some packages are available in 500 unit reels through designated sales channels with #TRMPBF suffix.

**ELECTRICAL CHARACTERISTICS** The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications are at  $T_A = 25^{\circ}$ C. V<sup>+</sup> = 5V, V<sup>-</sup> = GND = 0V unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS			MIN	ТҮР	MAX	UNITS
Power Sı	Ipply						I	
V+	Supply Voltage Range			٠	2.0		11.0	V
I <sub>CC</sub>	Supply Current		V, HYST = REF, C-Grade V, HYST = REF, I-Grade	•		0.3	0.68 0.71	μΑ μΑ
Compara	tor	•						
V <sub>OS</sub>	Comparator Input Offset Voltage	V <sub>CM</sub> = 2.5V LTC1540CMS8/I	MS8	•			±12 ±15 ±16	mV mV mV
I <sub>IN</sub>	Input Leakage Current (IN+, IN <sup>-</sup> ) Input Leakage Current (HYST)	$V_{IN}^{+} = V_{IN}^{-} = 2.5$	V	•		±0.01 ±0.02	±1.0 ±1.0	nA nA
V <sub>CM</sub>	Comparator Input Common Mode Range			٠	V-		V+ – 1.3V	V
CMRR	Common Mode Rejection Ratio	V <sup>-</sup> to V <sup>+</sup> – 1.3V				0.1	1	mV/V
PSRR	Power Supply Rejection Ratio	V <sup>+</sup> = 2V to 11V				0.1	1	mV/V
V <sub>HYST</sub>	Hysteresis Input Voltage Range			٠	REF – 50mV		REF	V
t <sub>PD</sub>	Propagation Delay	C <sub>OUT</sub> = 100pF	Overdrive = 10mV Overdrive = 100mV			60 50		μs μs
V <sub>OH</sub>	Output High Voltage	I <sub>0</sub> = -13mA		٠	V <sup>+</sup> - 0.4V			V
V <sub>OL</sub>	Output Low Voltage	l <sub>0</sub> = 1.8mA		٠			GND + 0.4V	V
Referenc	e							
V <sub>REF</sub>	Reference Voltage	No Load	(SO-8)/(DFN) Commercial MS8 Commercial (SO-8)/(MS8)/(DFN) Industrial	•	1.158 1.156 1.152	1.182 1.182 1.182	1.206 1.208 1.212	V V V
$\Delta V_{REF}$	Load Regulation	$0 \le I_{\text{SOURCE}} \le 10$	0μΑ	•		0.5	2.5	mV
		0 ≤ I <sub>SINK</sub> ≤ 10µА		•		0.5	1.5 5	mV mV

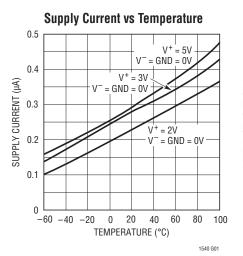
**ELECTRICAL CHARACTERISTICS** The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T<sub>A</sub> = 25°C. V<sup>+</sup> = 3V, V<sup>-</sup> = GND = 0V unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS			MIN	ТҮР	MAX	UNITS
Power Su	ipply	_			1			
V+	Supply Voltage Range				2		11	١
I <sub>CC</sub>	Supply Current		(, HYST = REF, C-Grade /, HYST = REF, I-Grade	•		0.28	0.61 0.64	μ <i>Α</i> μ <i>Α</i>
Compara	tor							
V <sub>OS</sub>	Comparator Input Offset Voltage	V <sub>CM</sub> = 2.5V LTC1540CMS8/IN	158	•			±12 ±15 ±16	mV mV mV
I <sub>IN</sub>	Input Leakage Current (IN+, IN <sup>-</sup> ) Input Leakage Current (HYST)	$V_{IN}^{+} = V_{IN}^{-} = 1.5$	1	•		±0.01 ±0.02	±1 ±1	nA nA
V <sub>CM</sub>	Comparator Input Common Mode Range			•	V-		V+ – 1.3V	V
CMRR	Common Mode Rejection Ratio	V <sup>-</sup> to V <sup>+</sup> – 1.3V				0.1	1	mV/V
PSRR	Power Supply Rejection Ratio	V <sup>+</sup> = 2V to 11V				0.1	1	mV/V
V <sub>HYST</sub>	Hysteresis Input Voltage Range			•	REF – 50mV		REF	V
t <sub>PD</sub>	Propagation Delay	C <sub>OUT</sub> = 100pF	Overdrive = 10mV Overdrive = 100mV			70 60		μs μs
V <sub>OH</sub>	Output High Voltage	$I_0 = -8mA$		•	V+-0.4V			V
V <sub>OL</sub>	Output Low Voltage	I <sub>0</sub> = 0.8mA		•			GND + 0.4V	V
Referenc	e							
V <sub>REF</sub>	Reference Voltage	No Load	(SO-8)/(DFN) Commercial MS8 Commercial (SO-8)/(MS8)/(DFN) Industrial	•	1.158 1.156 1.152	1.182 1.182 1.182	1.206 1.208 1.212	V V V
$\Delta V_{REF}$	Load Regulation	$0 \le I_{\text{SOURCE}} \le 100$	μΑ			0.75	3.5	m٧
		$0 \le I_{SINK} \le 10 \mu A$		•		0.5	1.5 5	m∨ mV

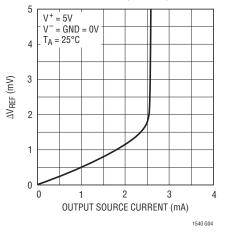
Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: The  $\theta_{JA}$  specified for the DD package is with minimal PCB heat spreading metal. Using expanded metal area on all layers of a board reduces this value.

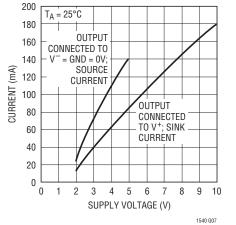
### **TYPICAL PERFORMANCE CHARACTERISTICS**



Reference Voltage Load Regulation (Source)

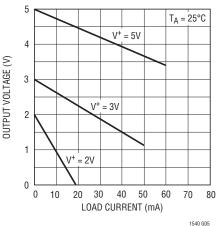


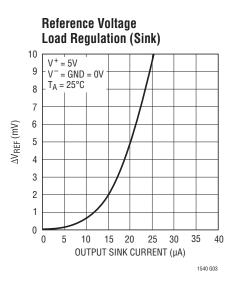
Comparator Short-Circuit Current vs Supply Voltage



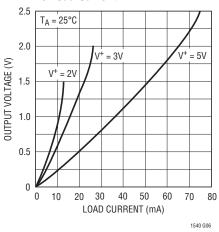
**Reference Voltage vs Temperature** 1.190 V<sup>+</sup> = 5V 1.188  $V^- = GND = 0V$ () 1.186 () 1.184 1.184 1.182 1.180 1.180 1.178 1.178 1.176 1.176 1.174 1.172 20 60 80 -60 -40 -20 0 40 100 TEMPERATURE (°C) 1540 G02

Comparator Output Voltage (High) vs Load Current

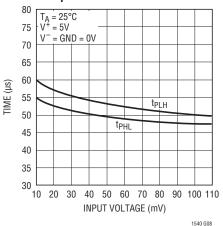




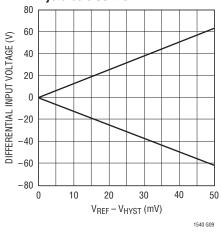
Comparator Output Voltage (Low) vs Load Current



Comparator Response Time vs Input Overdrive



Hysteresis Control



# PIN FUNCTIONS

**GND (Pin 1):** Ground. Connect to  $V^-$  for single supply operation.

**V**<sup>-</sup> (**Pin 2**): Negative Supply. Potential should be more negative than GND. Connect to ground for single supply operation.

**IN<sup>+</sup> (Pin 3)**: Noninverting Comparator Input. Input common mode range from V<sup>-</sup> to V<sup>+</sup> – 1.3V. Input current typically 10pA at 25°C.

**IN<sup>-</sup> (Pin 4):** Inverting Comparator Input. Input common mode range from V<sup>-</sup> to V<sup>+</sup> – 1.3V. Input current typically 10pA at 25°C.

**HYST (Pin 5):** Hysteresis Input. Connect to REF if not used. Input voltage range is from  $V_{REF}$  to  $V_{REF}$  – 50mV.

**REF (Pin 6):** Reference Output. 1.182V with respect to V<sup>-</sup>. Can source up to 1mA and sink  $10\mu A$  at 25°C. Drive 0.01 $\mu$ F bypass capacitor without oscillation.

# **APPLICATIONS INFORMATION**

The LTC1540 is a nanopower comparator with a built-in 1.182V reference. Features include programmable hysteresis, wide supply voltage range (2V to 11V) and the ability of the reference to drive up to a  $0.01\mu$ F capacitor without oscillation. The comparator's CMOS outputs can source up to 40mA while supply current glitches that normally occur when switching logic states, have been eliminated.

### **Power Supplies**

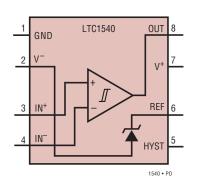
The comparator operates from a single 2V to 11V supply. The LTC1540 includes a separate ground for the comparator output stage, allowing a split supply ranging from  $\pm$  1V to  $\pm$ 5.5V. Connecting V<sup>-</sup> to GND will allow single supply operation. If the comparator output is required to source more than 1mA, or the supply source impedance is high, V<sup>+</sup> should be bypassed with a 0.1µF capacitor.

### **Comparator Inputs**

The comparator inputs can swing from the negative supply, V<sup>-</sup>, to within 1.3V (max) of the positive supply V<sup>+</sup>. The inputs can be forced 300mV below V<sup>-</sup> or above V<sup>+</sup> without damage and the typical input leakage current is only  $\pm 10$ pA.

**V<sup>+</sup> (Pin 7):** Positive Supply operating voltage is from 2V to 11V.

**OUT (Pin 8):** Comparator CMOS Output. Swings from GND to V<sup>+</sup>. Output can source up to 40mA and sink 5mA.



#### **Comparator Output**

The comparator output swings between GND and V<sup>+</sup> to assure TTL compatibility with a split supply. The output is capable of sourcing up to 40mA and sinking up to 5mA while still maintaining nanoampere quiescent currents. The output stage does not generate crowbar switching currents during transitions which helps minimize parasitic feedback through the supply pins.

#### Voltage Reference

The internal bandgap reference has a voltage of 1.182V referenced to V<sup>-</sup>. The reference accuracy is  $\pm 2.0\%$  from 0°C to 70°C. It can source up to 1mA and sink up to 10µA with a 5V supply. The reference can drive a bypass capacitor of up to 0.01µF without oscillation and by inserting a series resistor, capacitance values up to 10µF can be used (Figure 1).

Figure 2 shows the resistor value required for different capacitor values to achieve critical damping. Bypassing the reference can help prevent false tripping of the comparators by preventing glitches on V<sup>+</sup> or reference load transients from disturbing the reference output voltage.

### **APPLICATIONS INFORMATION**

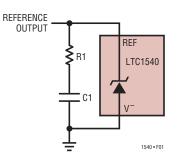


Figure 1. Damping the Reference Output

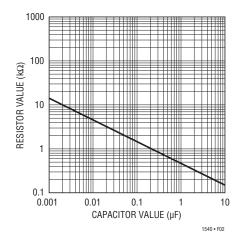


Figure 2. Damping Resistance vs Bypass Capacitor Value

Figure 3 shows the bypassed reference output with a square wave applied to the V<sup>+</sup> pin. Resistors R2 and R3 set 10mV of hysteresis voltage band while R1 damps the reference response. Note that the comparator output doesn't trip.

#### Low Voltage Operation: V<sup>+</sup> = 1.6V

The guaranteed minimum operating voltage is 2V (or  $\pm 1V$ ). As the total supply voltage is reduced below 2V, the performance degrades and the supply current falls. At low supply voltages, the comparator's output drive is reduced and the propagation delay increases. The V<sub>REF</sub> and V<sub>OS</sub> are also slightly worse. The useful input voltage range extends from the negative supply to 0.9V below the positive supply. Test your prototype over the full temperature and supply voltage range if operation below 2V is anticipated. Because of the increase in supply current, operation below 1.5V is not recommended (Figure 4).

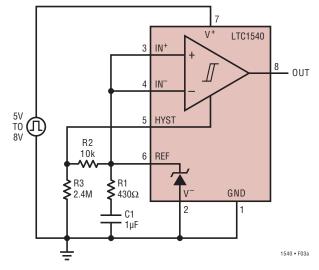


Figure 3a. Power Supply Transient Test Circuit

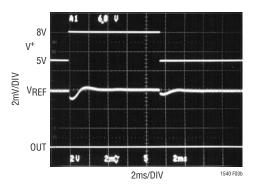


Figure 3b. Power Supply Transient Rejection

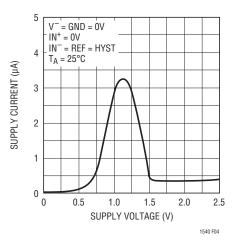


Figure 4. Supply Current vs Supply Voltage

# **APPLICATIONS INFORMATION**

#### Hysteresis

Hysteresis can be added to the LTC1540 by connecting a resistor (R1) between the REF and HYST pins and a second resistor (R2) from HYST to V<sup>-</sup> (Figure 5).

The difference between the upper and lower threshold voltages, or hysteresis voltage band ( $V_{HB}$ ), is equal to twice the voltage difference between the REF and HYST pins.

When more hysteresis is added, the upper threshold increases the same amount as the low threshold decreases. The maximum voltage allowed between REF and HYST pins is 50mV, producing a maximum hysteresis voltage band of 100mV. The hysteresis band may vary by up to

# TYPICAL APPLICATIONS

### Level Detector

The LTC1540 is ideal for use as a nanopower level detector as shown in Figure 6. R1 and R2 form a voltage divider from  $V_{IN}$  to the noninverting comparator input. R3 and R4 set the hysteresis voltage, and R5 and C1 bypass the reference output. The following design procedure can be used to select the component values:

- 1. Choose the  $V_{IN}$  voltage trip level, in this example 4.65V.
- 2. Calculate the required resistive divider ratio.

Ratio =  $V_{REF}/V_{IN}$ 

Ratio = 1.182V/4.65V = 0.254

3. Choose the required hysteresis voltage band at the input  $V_{HBIN}$ , in this example 60mV. Calculate the hysteresis voltage band referred to the comparator input  $V_{HB}$ .

 $V_{HB} = (V_{HBIN})(Ratio)$  $V_{HB} = (60mV)(0.254)$ 

 $V_{HB} = 15.24 mV$ 

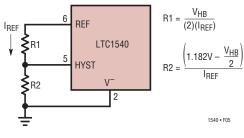
4. Choose the values for R3 and R4 to set the hysteresis.

R4 = 2.4M

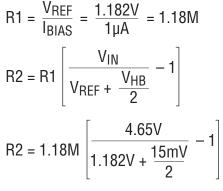
R3 (k
$$\Omega$$
) = 15k, V<sub>HB</sub> (mV) = 15mV

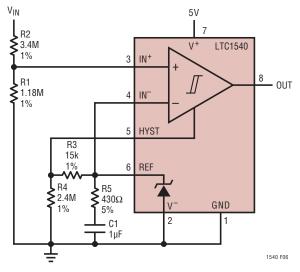
5. Choose the values for R1 and R2 to set the trip point.

15%. If hysteresis is not wanted, the HYST pin should be shorted to REF. Acceptable values for  $I_{REF}$  range from 0.1µA to 5µA. If 2.4M is chosen for R2, then the value of R1 (k $\Omega$ ) is equal to the value of V<sub>HB</sub> (mV).











# TYPICAL APPLICATIONS

#### 3.3V Output Low Dropout Linear Regulator

The LTC1540 can be connected as a micropower ( $I_Q = 5.5\mu$ A at  $V_{IN} = 5V$ ) low dropout linear regulator (Figure 7). When the output is low, Q1 turns on, allowing current to charge output capacitor C1. Local feedback formed by R4, Q1 and Q2 creates a constant-current source from the 5V input to C1. R4, R1 and Q2's V<sub>BE</sub> also provide current limiting in the case of an output short-circuit to ground. C2 reduces output ripple, while the R2-R3 feedback voltage divider establishes the output voltage.

#### Auto Power-Off Source

Figure 8 shows the circuit for a 30mA power supply that has a timed auto power-off function. The comparator output

is the switched power supply output. With a 10mA load, it typically provides a voltage of ( $V_{BAT} - 0.17V$ ). The whole circuit draws a mere 0.8µA of quiescent current with  $V_{BAT} = 5V$ . The three resistor voltage divider programs 50mV of hysteresis for the comparator, and sets the IN<sup>-</sup> voltage at 200mV. This gives an IN<sup>+</sup> trip threshold of approximately 150mV

The RC time constant determines the maximum power-on time of the OUT pin before power down occurs. This period can be approximated by:

t = 4.6RC (seconds)

The actual time will vary with both the leakage current of the capacitor and the input current at the  $IN^+$  pin.

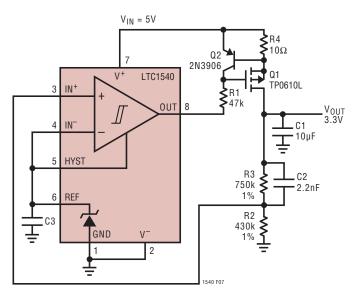


Figure 7. 3.3V Output Low Dropout Linear Regulator

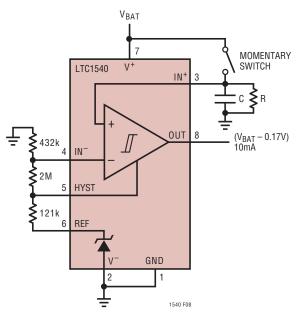


Figure 8. Auto Power-Off Switch Operates on 0.8µA Quiescent Current

# TYPICAL APPLICATIONS

#### **Low-Battery Detect**

Figure 9 shows how to use the LTC1540 for a low-battery detect, drawing only  $1.4\mu A$  at  $V_{BAT} = 2V$ . The circuit is powered by a 2-cell NiCd battery. The  $V_{BAT}$  pin could be as low as 1.6V when the batteries are completely depleted. The electrical specifications of the LTC1540 guarantee operation

down to a supply voltage of 2V, but it is still functional with the supply as low as 1.6V. Some parameters, such as  $V_{REF}$  and  $V_{OS}$ , will be degraded on lower supply voltages. The input voltage range extends from 0.9V below the positive supply to the negative supply.

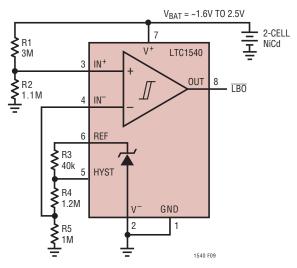
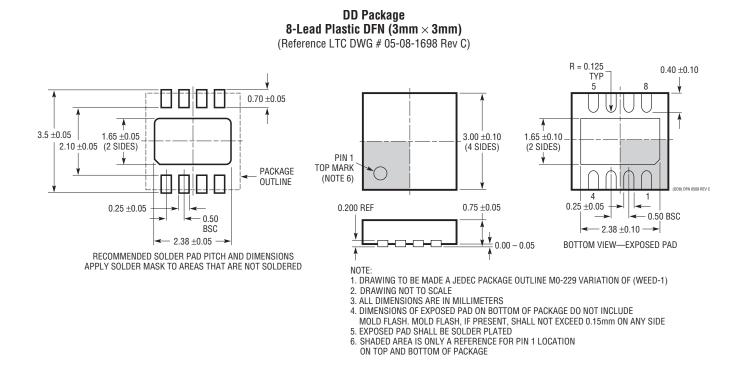


Figure 9. Low-Battery Detect Works Down to 1.6V

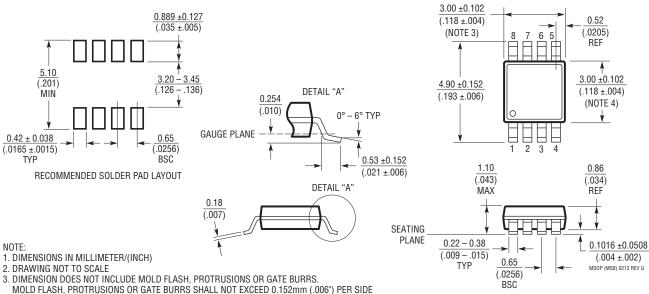
### PACKAGE DESCRIPTION

Please refer to http://www.linear.com/product/LTC1540#packaging for the most recent package drawings.



MS8 Package 8-Lead Plastic MSOP

(Reference LTC DWG # 05-08-1660 Rev G)



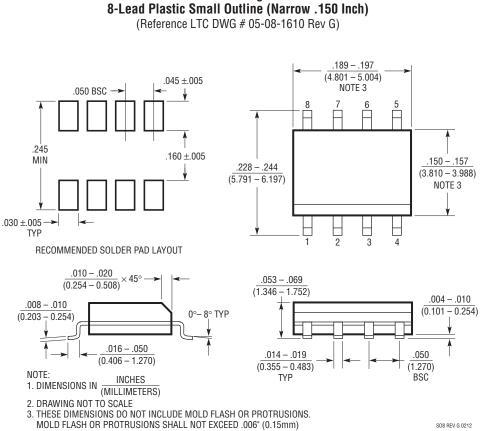
4. DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.

INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.152mm (.006") PER SIDE

5. LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.102mm (.004") MAX

### PACKAGE DESCRIPTION

Please refer to http://www.linear.com/product/LTC1540#packaging for the most recent package drawings.



S8 Package

4. PIN 1 CAN BE BEVEL EDGE OR A DIMPLE

# **REVISION HISTORY** (Revision history begins at Rev B)

REV	DATE	DESCRIPTION	PAGE NUMBER
В	08/17	Reformatted Order Information.	2
		Added web links.	All Pages
		Add Information for analog devices.	All Pages
		Corrected test conditions for supply current specification (I <sub>CC</sub> ).	3

# TYPICAL APPLICATION

### **RF Field Detector**

Figure 10 shows the complete circuit for a field detector which was tested at 445MHz. A transmission line is used to match the detector diode (1N5712) to a quarter-wave whip antenna. The 0.23 $\lambda$  wavelength transmission line section transforms the 1pF (350 $\Omega$ ) diode junction capacitance to a virtual short at the base of the antenna. At the same time it converts the received antenna current to a voltage loop at the diode, giving excellent sensitivity.

The rectified output is monitored by the LTC1540 comparator. The internal reference is used to set up a threshold of about 18mV at the inverting input. A rising edge at the comparator output triggers a one shot that temporarily enables answer back and any other pulsed functions.

The total supply current is 400nA. Among other monolithic one shots, the CD4047 draws the least amount of transient current.

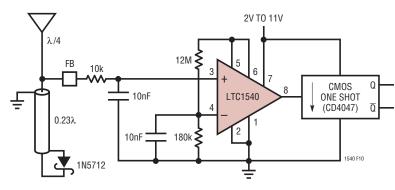


Figure 10. Nanopower Field Detector

# **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS
LT®1178/LT1179	Dual/Quad 17µA Precision Single Supply Op Amps	70μV Max V <sub>OS</sub> , 5nA Max I <sub>BIAS</sub>
LT1351	Single 250µA, 3MHz, 200V/µs Op Amp with Shutdown	C-Load <sup>™</sup> Op Amp Stable Driving Any Capacitive Load
LT1352/LT1353	Dual/Quad 250µA, 3MHz, 200V/µs Op Amps	C-Load Op Amps Stable Driving Any Capacitive Load
LTC1440	Micropower Comparator with 1% Reference	1.182V ±1% Reference, ±10mV (Max) Input Offset
LTC1443/LTC1444/ LTC1445	Micropower Quad Comparators with 1% Reference	LTC1443 Has 1.182V Reference, LTC1444/LTC1445 Have 1.221V Reference and Adjustable Hysteresis
LTC1474	Low Quiescent Current High Efficiency Step-Down Converter	10µA Standby Current, 92% Efficiency, Space Saving 8-Pin MSOP Package
LT1495	1.5μΑ Max, Dual Precision Rail-to-Rail Input and Output Op Amp	375μV Max V <sub>OS</sub> , 250pA I <sub>BIAS</sub> , 25pA I <sub>OS</sub>
LT1521	300mA Low Dropout Regulator with Micropower Quiescent Current and Shutdown	0.5V Dropout Voltage, 12µA Quiescent Current, Adjustable Output 3V, 3.3V and 5V Fixed
LT1634	Micropower Precision Shunt Voltage Reference	1.25V, 2.5V, 4.096V, 5V Outputs, 10µA Operating Current, 0.05% Initial Accuracy 25ppm/°C Max Drift, SO-8, MSOP and TO-92 Packages



# **Mouser Electronics**

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Analog Devices Inc.:

LTC1540CDD#TR LTC1540IDD LTC1540IS8#PBF LTC1540CS8 LTC1540IDD#TRPBF LTC1540IMS8#TR LTC1540CS8#PBF LTC1540CMS8#TR LTC1540IS8#TR LTC1540CDD LTC1540IMS8 LTC1540CMS8#TRPBF LTC1540IMS8#TRPBF LTC1540CS8#TR LTC1540CS8#TRPBF LTC1540CDD#PBF LTC1540IS8 LTC1540CMS8 LTC1540IS8#TRPBF LTC1540IDD#TR LTC1540CMS8#PBF LTC1540IMS8#PBF LTC1540IDD#PBF LTC1540IDD#TRPBF