

ISL28108, ISL28208, ISL28408

40V Precision Single Supply Rail-to-Rail Output Low Power Operational Amplifiers

FN6935 Rev.5.01 Feb 6, 2020

The ISL28108, ISL28208, and ISL28408 are single, dual and quad low power precision amplifiers optimized for single supply applications. These devices feature a common mode input voltage range extending to 0.5V below the V- rail, a rail-to-rail differential input voltage range for use as a comparator, and rail-to-rail output voltage swing, which make them ideal for single supply applications where input operation at ground is important.

Added features include low offset voltage, and low temperature drift making them the ideal choice for applications requiring high DC accuracy. The output stage is capable of driving large capacitive loads from rail-to-rail for excellent ADC driving performance. The devices can operate for single or dual supply from 3V (±1.5V) to 40V (±20V) and are fully characterized at ±5V and ±15V. The combination of precision, low power, and small footprint provide the user with outstanding value and flexibility relative to similar competitive parts.

Applications for these amplifiers include precision instrumentation, data acquisition, precision power supply control, and industrial control.

The ISL28108 single is offered in 8 Ld TDFN, MSOP and SOIC packages. The ISL28208 dual amplifier is offered in 8 Ld TDFN, MSOP, and SOIC packages. The ISL28408 is offered in 14 Ld SOIC package. All devices are offered in standard pin configurations and operate across the extended temperature range of -40°C to +125°C.

Features

- Single or dual supply, rail-to-rail output and below ground (V-) input capability
- · Rail-to-rail input differential voltage range for comparator applications
- • Low current consumption $(V_S = \pm 5V) \dots 165\mu A$ • Low input offset voltage (ISL28108)............. 150μV · Superb temperature drift
- Voltage offset TC 0.1µV/°C, Typ • Low input bias current.....-13nA Typ
- · No phase reversal

Applications

- · Precision instruments
- · Medical instrumentation
- · Data acquisition
- · Power supply control
- · Industrial process control

Related Literature

For a full list of related documents, visit our website:

• ISL28108, ISL28208, and ISL28408

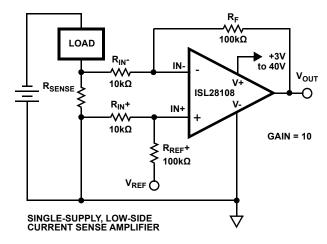


FIGURE 1. TYPICAL APPLICATION CIRCUIT

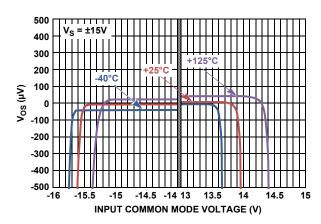


FIGURE 2. INPUT OFFSET VOLTAGE vs INPUT COMMON MODE VOLTAGE, $V_S = \pm 15V$

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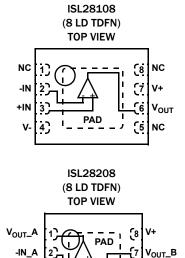
Ordering Information

PART NUMBER	PART	TEMP. RANGE	TAPE AND REEL	PACKAGE	PKG.
(Notes 1, 2, 3)	MARKING	(°C)	(Units) (Note 1)	(RoHS Compliant)	DWG. #
ISL28108FBZ	28108 FBZ	-40 to +125	-	8 Ld SOIC	M8.15E
ISL28108FBZ-T7	28108 FBZ	-40 to +125	1k	8 Ld SOIC	M8.15E
ISL28108FBZ-T7A	28108 FBZ	-40 to +125	250	8 Ld SOIC	M8.15E
ISL28108FBZ-T13	28108 FBZ	-40 to +125	2.5k	8 Ld SOIC	M8.15E
ISL28108FRTZ	108Z	-40 to +125	-	8 Ld TDFN	L8.3x3K
ISL28108FRTZ-T7	108Z	-40 to +125	1k	8 Ld TDFN	L8.3x3K
ISL28108FRTZ-T7A	108Z	-40 to +125	250	8 Ld TDFN	L8.3x3K
ISL28108FRTZ-T13	108Z	-40 to +125	6k	8 Ld TDFN	L8.3x3K
ISL28208FBZ	28208 FBZ	-40 to +125	-	8 Ld SOIC	M8.15E
ISL28208FBZ-T7	28208 FBZ	-40 to +125	1k	8 Ld SOIC	M8.15E
ISL28208FBZ-T7A	28208 FBZ	-40 to +125	250	8 Ld SOIC	M8.15E
ISL28208FBZ-T13	28208 FBZ	-40 to +125	2.5k	8 Ld SOIC	M8.15E
ISL28208FRTZ	208F	-40 to +125	-	8 Ld TDFN	L8.3x3K
ISL28208FRTZ-T7	208F	-40 to +125	1k	8 Ld TDFN	L8.3x3K
ISL28208FRTZ-T7A	208F	-40 to +125	250	8 Ld TDFN	L8.3x3K
ISL28208FRTZ-T13	208F	-40 to +125	2.5k	8 Ld TDFN	L8.3x3K
ISL28208FUZ	8208Z	-40 to +125	-	8 Ld MSOP	M8.118B
ISL28208FUZ-T7	8208Z	-40 to +125	1k	8 Ld MSOP	M8.118B
ISL28208FUZ-T7A	8208Z	-40 to +125	250	8 Ld MSOP	M8.118B
ISL28208FUZ-T13	8208Z	-40 to +125	2.5k	8 Ld MSOP	M8.118B
ISL28408FBZ	28408 FBZ	-40 to +125	-	14 Ld SOIC	M14.15
ISL28408FBZ-T7	28408 FBZ	-40 to +125	1k	14 Ld SOIC	M14.15
ISL28408FBZ-T7A	28408 FBZ	-40 to +125	250	14 Ld SOIC	M14.15
ISL28408FBZ-T13	28408 FBZ	-40 to +125	2.5k	14 Ld SOIC	M14.15
ISL28208SOICEVAL2Z	Evaluation Board	<u>, </u>			

- 1. See $\underline{\text{TB347}}$ for details about reel specifications.
- 2. These Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
- 3. For Moisture Sensitivity Level (MSL), see ISL28108, ISL28208, and ISL28408 device pages. For more information about MSL, see TB363.

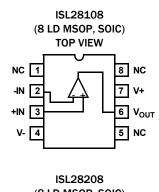
Pin Configurations

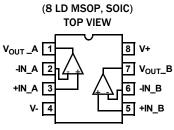
+IN_A

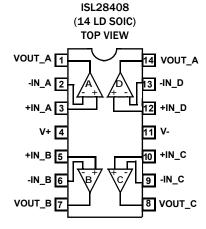


-IN_B

+IN_B

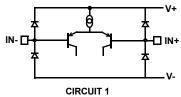


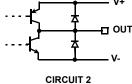


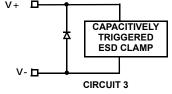


Pin Descriptions

ISL28108 OIC, MSOP, TDFN)	ISL28208 (8 Ld SOIC, MSOP, TDFN)	ISL28408 (14 Ld SOIC)	PIN NAME	EQUIVALENT CIRCUIT	DESCRIPTION
3	-	-	+IN	Circuit 1	Amplifier non-inverting input
-	3	3	+IN_A	_	
-	5	5	+IN_B		
-	-	10	+IN_C		
-	-	12	+IN_D		
4	4	11	V-	Circuit 3	Negative power supply
2	-	-	-IN	Circuit 1	Amplifier inverting input
-	2	2	-IN_A		
-	6	6	-IN_B		
-	-	9	-IN_C	_	
-	-	13	-IN_D	-	
7	8	4	V+	Circuit 3	Positive power supply
6	-	-	V _{OUT}	Circuit 2	Amplifier output
-	1	1	V _{OUT} _A	-	
-	7	7	V _{OUT} B	_	
-	-	8	V _{OUT} _C	-	
-	-	14	V _{OUT} D		
1, 5, 8	-	-	NC	-	No internal connection
PAD	PAD	-	PAD	-	Thermal Pad - TDFN package only. Connect thermal pad to ground or monegative potential.







Absolute Maximum Ratings

Thermal Information

Thermal Resistance (Typical)	θ_{JA} (°C/W	/) θ _{JC} (°C/W)
8 Ld SOIC Package (208, Notes 4, 7)	120	55
8 Ld SOIC Package (108, Notes 4, 7)	120	60
8 Ld TDFN Package (208, Notes 5, 6)	47	6
8 Ld TDFN Package (108, Notes 5, 6)	45	3.5
8 Ld MSOP Package (208, Notes 4, 7)	150	50
14 Ld SOIC Package (408, Notes 4, 7)	71	37
Storage Temperature Range		-65°C to +150°C
Pb-Free Reflow Profile		see <u>TB493</u>

Operating Conditions

Ambient Operating Temperature Range40	0°C to	+125°C
Maximum Operating Junction Temperature		+150°C
Supply Voltage) to 40	V (±20V)

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions can adversely impact product reliability and result in failures not covered by warranty.

NOTES:

- 4. θ_{IA} is measured with the component mounted on a high effective thermal conductivity test board in free air. See TB379 for details.
- 5. θ_{JA} is measured in free air with the component mounted on a high effective thermal conductivity test board with direct attach features. See <u>1B379</u>.
- 6. For θ_{JC} , the case temperature location is the center of the exposed metal pad on the package underside.
- 7. For $\theta_{\text{JC}},$ the case temperature location is taken at the package top center.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

Electrical Specifications, V_S ±15 $V_{CM} = 0$, $V_{O} = 0V$, $R_{L} = 0$ pen, $T_{A} = +25$ °C, unless otherwise noted. **Boldface entries apply over** the operating temperature range, -40 °C to +125 °C. Temperature data established by characterization.

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNIT
V _{OS}	Input Offset Voltage	ISL28208	-230	25	230	μV
		ISL28408	-330		330	μV
		ISL28108 SOIC, TDFN	-150	10	150	μV
			-270		270	μV
TCV _{OS}	Input Offset Voltage Temperature Coefficient	ISL28208 SOIC -40°C to +125°C		0.1	1.1	μV/°C
		ISL28208 MSOP -40°C to +125°C		0.2	1.5	μV/°C
		ISL28208 TDFN ISL28408 -40°C to +125°C		0.2	1.4	μV/°C
		ISL28108 SOIC, TDFN -40°C to +125°C		0.2	1.2	μV/°C
ΔV_{OS}	Input Offset Voltage Match (ISL28208 only)	ISL28208 SOIC, TDFN	-300	5	300	μV
			-400		400	μV
		ISL28208 MSOP	-420		420	μV
I _B	Input Bias Current		-43	-13		nA
			-63			nA
TCIB	Input Bias Current Temperature Coefficient			0.07		nA/°C



Electrical Specifications, V_S \pm 15 $V_{CM} = 0$, $V_O = 0V$, $R_L = 0$ pen, $T_A = +25$ °C, unless otherwise noted. Boldface entries apply over the operating temperature range, -40 °C to +125 °C. Temperature data established by characterization. (Continued)

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNIT
I _{os}	Input Offset Current	ISL28208	-3	0	3	nA
			-4		4	nA
		ISL28108 SOIC, TDFN	-4	0	4	nA
		ISL28408	-5		5	nA
CMRR	Common-Mode Rejection Ratio	V _{CM} = V ₋ -0.5V to V ₊ -1.8V		119		dB
		V _{CM} = V ₋ -0.2V to V ₊ -1.8V		123		dB
				102		dB
		$V_{CM} = V_{-}$ to V_{+} -1.8V	105	123		dB
			102	115		dB
V _{CMIR}	Common Mode Input Voltage	Guaranteed by CMRR test	V ₋ - 0.5		V ₊ - 1.8	V
	Range		V.		V ₊ - 1.8	V
PSRR	Power Supply Rejection Ratio	V _S = 3V to 40V, V _{CMIR} = Valid Input Voltage	110	128		dB
			109	124		dB
A _{VOL}	Open-Loop Gain	V_0 = -13V to +13V, R_L = 10k Ω to ground	117	126		dB
			100			dB
V _{OL}	Output Voltage Low,	$R_L = 10k\Omega$		52	85	mV
	V _{OUT} to V ₋				145	mV
V _{OH}	Output Voltage High,	$R_L = 10k\Omega$		70	110	mV
	V ₊ to V _{OUT}				150	mV
I _S	Supply Current/Amplifier	R _L = Open		185	250	μA
				270	350	μA
I _{SC+}	Output Short Circuit Source Current	$R_L = 10\Omega$ to V		19		mA
I _{sc-}	Output Short Circuit Sink Current	$R_L = 10\Omega$ to V_+		30		mA
V _{SUPPLY}	Supply Voltage Range	Guaranteed by PSRR	3		40	V
AC SPECIFICATI	ONS					1
GBWP	Gain Bandwidth Product	$A_{CL} = 101, V_0 = 100 \text{mV}_{P-P}, R_L = 2 \text{k}\Omega$		1.2		MHz
e _{np-p}	Noise Voltage	0.1Hz to 10Hz; V _S = ±18V		580		nVP-P
e _n	Noise Voltage Density	f = 10Hz; V _S = <u>+</u> 18V		18		nV/√Hz
e _n	Noise Voltage Density	f = 100Hz; V _S = <u>+</u> 18V		16		nV/√Hz
e _n	Noise Voltage Density	f = 1kHz; V _S = ±18V		15.8		nV/√Hz
e _n	Noise Voltage Density	f = 10kHz; V _S = ±18V		15.8		nV/√Hz
i _n	Noise Current Density	f = 10kHz; V _S = ±18V		80		fA/√Hz
THD + N	Total Harmonic Distortion + Noise	1kHz, $A_V = 1$, $V_O = 3.5V_{RMS}$, $R_L = 10k\Omega$		0.00042		%
TRANSIENT RES	SPONSE					
SR	Slew Rate, V _{OUT} 20% to 80%	$A_V = 1$, $R_L = 2k\Omega$, $V_O = 10V_{P-P}$		0.45		V/µs
t _r , t _f , Small Signal	Rise Time, V _{OUT} 10% to 90%	$A_V = 1, V_{OUT} = 100 \text{mV}_{P-P}, R_f = 0\Omega, R_L = 2k\Omega \text{ to}$ V_{CM}		264		ns
	Fall Time, V _{OUT} 90% to 10%	$A_V = 1, V_{OUT} = 100 \text{mV}_{P-P}, R_f = 0\Omega, R_L = 2k\Omega \text{ to } V_{CM}$		254		ns
t _s	Settling Time to 0.01% 10V Step; 10% to V _{OUT}	$A_V = -1$, $V_{OUT} = 10V_{P-P}$, $R_g = R_f = 10k$, $R_L = 2k\Omega$ to V_{CM}		27		μs



Electrical Specifications, V_S \pm 5V $V_{CM} = 0$, $V_0 = 0V$, $T_A = +25\,^{\circ}$ C, unless otherwise noted. Boldface entries apply over the operating temperature range, -40 °C to +125 °C. Temperature data established by characterization.

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	ТҮР	MAX (Note 8)	UNIT
V _{os}	Input Offset Voltage	ISL28208	-230	25	230	μV
		ISL28408	-330		330	μV
		ISL28108 SOIC, TDFN	-150	10	150	μV
			-270		270	μV
TCV _{OS}	Input Offset Voltage Temperature Coefficient	ISL28208 SOIC -40°C to +125°C		0.1	1.1	μV/°C
		ISL28208 MSOP -40°C to +125°C		0.2	1.5	μV/°C
		ISL28208 TDFN ISL28408 -40°C to +125°C		0.2	1.4	μV/°C
		ISL28108 SOIC, TDFN -40°C to +125°C		0.2	1.2	μV/°C
ΔV _{OS}	Input Offset Voltage Match		-300	3	300	μV
	(ISL28208 only)		-400		400	μV
I _B	Input Bias Current		-43	-15		nA
			-63			nA
TCIB	Input Bias Current Temperature Coefficient	-40°C to +125°C		-0.067		nA/°C
I _{OS}	Input Offset Current	ISL28208	-3	0	3	nA
			-4		4	nA
		ISL28108 SOIC, TDFN	-4	0	4	nA
		ISL28408	-5		5	nA
CMRR	Common-Mode Rejection Ratio	$V_{CM} = V_{-} - 0.5V \text{ to } V_{+} - 1.8V$		101		dB
		$V_{CM} = V_{-}-0.2V \text{ to } V_{+}-1.8V$		123		dB
				89		dB
		V _{CM} = V ₋ to V ₊ -1.8V ISL28108, ISL28208 V _{CM} = V ₋ to V ₊ -1.8V	105	123		dB
			100	112		dB
			105	123		dB
		ISL28408	97	112		dB
V _{CMIR}	Common Mode Input Voltage	Guaranteed by CMRR test	V ₋ - 0.5		V ₊ - 1.8	٧
	Range		V.		V ₊ - 1.8	٧
PSRR	Power Supply Rejection Ratio	V _S = 3V to 10V, V _{CMIR} = Valid Input Voltage	110	126		dB
		All except ISL28208 MSOP	109	123		dB
		ISL28208 MSOP	109	126		dB
		V _S = 3V to 10V, V _{CMIR} = Valid Input Voltage	107	123		dB
A _{VOL}	Open-Loop Gain	$V_0 = -3V$ to $+3V$, $R_L = 10k\Omega$ to ground	117	124		dB
			99			dB
V _{OL}	Output Voltage Low,	$R_L = 10k\Omega$		23	38	mV
	V _{OUT} to V ₋				48	mV
V _{OH}	Output Voltage High,	$R_L = 10k\Omega$		30	65	mV
	V ₊ to V _{OUT}	_			70	mV



Electrical Specifications, V_S \pm 5V $V_{CM} = 0$, $V_O = 0V$, $T_A = +25\,^{\circ}$ C, unless otherwise noted. Boldface entries apply over the operating temperature range, -40 °C to +125 °C. Temperature data established by characterization. (Continued)

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNIT
I _S	Supply Current/Amplifier	R _L = Open		165	250	μΑ
				240	350	μΑ
I _{SC+}	Output Short Circuit Source Current	$R_L = 10\Omega$ to V_{-}		14		mA
I _{SC-}	Output Short Circuit Sink Current	$R_L = 10\Omega$ to V_+		22		mA
AC SPECIFICAT	IONS			1	I .	1
GBW	Gain Bandwidth Product	$A_{CL} = 101, V_0 = 100 \text{mV}_{P-P}, R_L = 2k\Omega$		1.2		MHz
e _{np-p}	Noise Voltage	0.1Hz to 10Hz		600		nV _{P-P}
e _n	Noise Voltage Density	f = 10Hz		18		nV/√Hz
e _n	Noise Voltage Density	f = 100Hz		16		nV/√Hz
e _n	Noise Voltage Density	f = 1kHz		15.8		nV/√Hz
e _n	Noise Voltage Density	f = 10kHz		15.8		nV/√Hz
i _n	Noise Current Density	f = 10kHz		90		fA/√Hz
TRANSIENT RE	SPONSE			1		
SR	Slew Rate, V _{OUT} 20% to 80%	$A_V = 1$, $R_L = 2k\Omega$, $V_0 = 4V_{P-P}$		0.4		V/µs
t _r , t _f , Small Signal	Rise Time, V _{OUT} 10% to 90%	$\begin{aligned} &A_V = 1, V_{OUT} = 100 m V_{P-P}, R_f = 0 \Omega, R_L = 2 k \Omega \ to \\ &V_{CM} \end{aligned}$		264		ns
	Fall Time, V _{OUT} 90% to 10%	A_V = 1, V_{OUT} = 100m V_{P-P} , R_f = 0 Ω , R_L = 2k Ω to V_{CM}		254		ns
t _s	Settling Time to 0.01% 4V Step; 10% to V _{OUT}	$A_V = -1, V_{OUT} = 4V_{P-P}, R_g = R_f = 10k, R_L = 2k\Omega to$ V_{CM}		14.4		μs

^{8.} Compliance to data sheet limits is assured by one or more methods: production test, characterization and/or design.

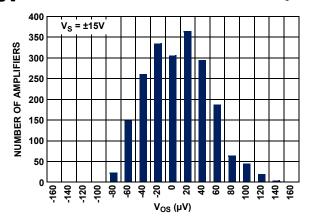


FIGURE 3. ISL28408 SOIC INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 15 V$

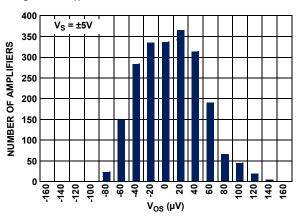


FIGURE 4. ISL28408 SOIC INPUT OFFSET VOLTAGE DISTRIBUTION, $V_c = \pm 5V$

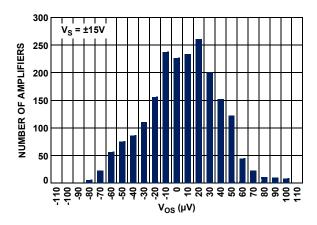


FIGURE 5. ISL28208 INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 15V$

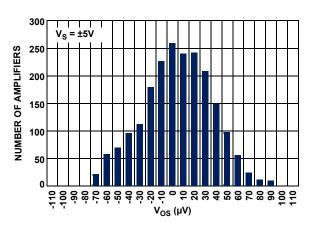


FIGURE 6. ISL28208 INPUT OFFSET VOLTAGE DISTRIBUTION, $V_c = \pm 5V$

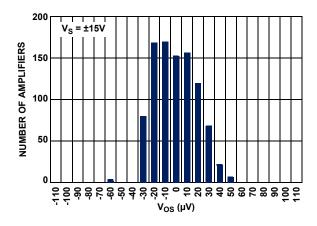


FIGURE 7. ISL28108 SOIC INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 15 V$

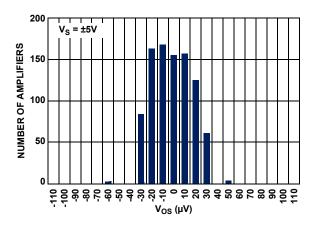


FIGURE 8. ISL28108 SOIC INPUT OFFSET VOLTAGE DISTRIBUTION, $\rm V_S = \pm 5V$



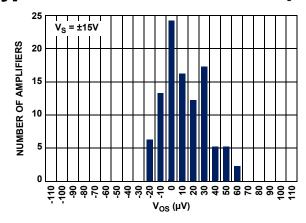


FIGURE 9. ISL28108 TDFN INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 15 V$

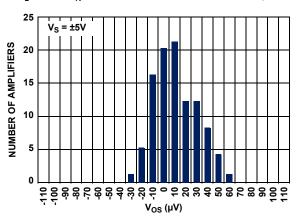


FIGURE 10. ISL28108 TDFN INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 5V$

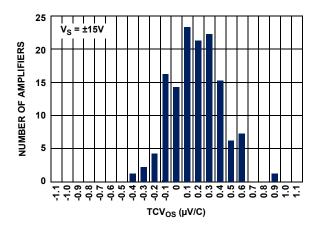


FIGURE 11. ISL28408 SOIC TCV $_{OS}$ vs NUMBER OF AMPLIFIERS, $V_S=\pm 15 \text{V}$

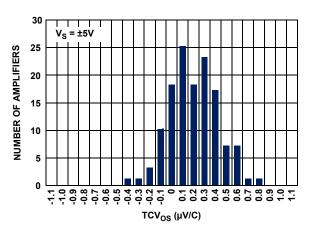


FIGURE 12. ISL28408 SOIC TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 5V$

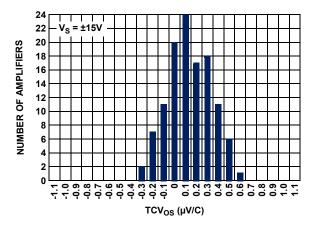


FIGURE 13. ISL28208 SOIC TCV $_{0S}$ vs NUMBER OF AMPLIFIERS, $V_{e}=\pm15V$

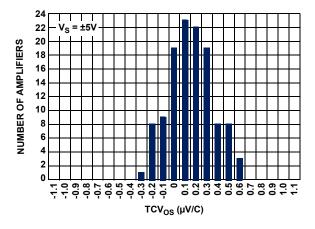


FIGURE 14. ISL28208 SOIC TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_c = \pm 5V$

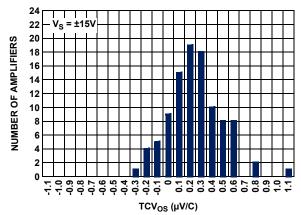


FIGURE 15. ISL28208 TDFN AND MSOP TCV $_{0S}$ vs number of amplifiers, $V_S = \pm 15V$

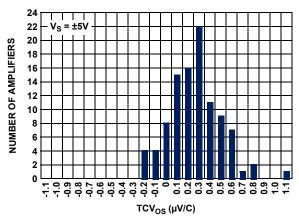


FIGURE 16. ISL28208 TDFN AND MSOP TCV $_{
m OS}$ vs NUMBER OF AMPLIFIERS, V $_{
m S}=\pm5{\rm V}$

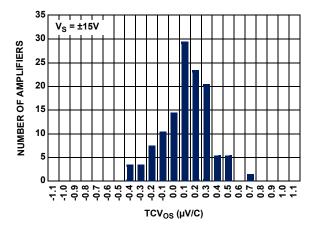


FIGURE 17. ISL28108 SOIC TCV $_{0S}$ vs NUMBER OF AMPLIFIERS, $V_S = \pm 15V$

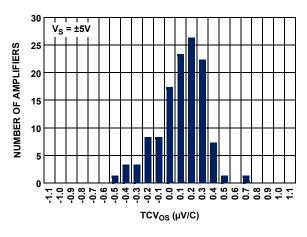


FIGURE 18. ISL28108 SOIC TCV $_{OS}$ vs NUMBER OF AMPLIFIERS, $V_S = \pm 5 V$

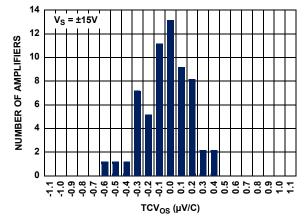


FIGURE 19. ISL28108 TDFN TCV $_{OS}$ vs NUMBER OF AMPLIFIERS, $\rm V_S = \pm 15V$

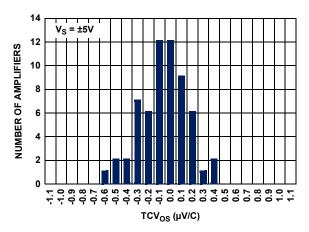


FIGURE 20. ISL28108 TDFN TCV $_{OS}$ vs NUMBER OF AMPLIFIERS, $V_S = \pm 5V$

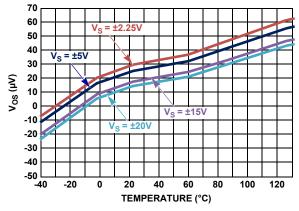


FIGURE 21. V_{OS} vs TEMPERATURE

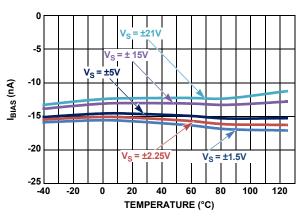


FIGURE 22. IBIAS VS TEMPERATURE VS SUPPLY

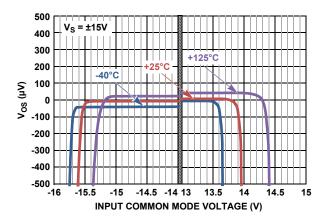


FIGURE 23. INPUT OFFSET VOLTAGE vs INPUT COMMON MODE $VOLTAGE, \ V_S = \pm 15V$

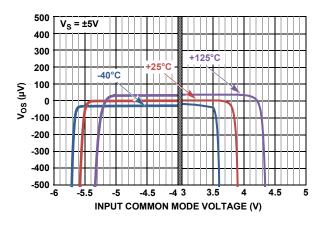


FIGURE 24. INPUT OFFSET VOLTAGE vs INPUT COMMON MODE $VOLTAGE, \ V_S = \pm 5V$

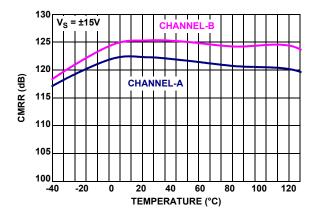


FIGURE 25. CMRR vs TEMPERATURE, $V_S = \pm 15V$

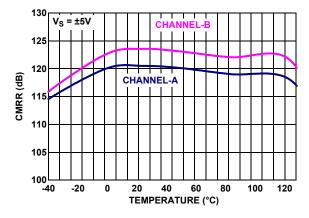


FIGURE 26. CMRR vs TEMPERATURE, $V_S = \pm 5V$

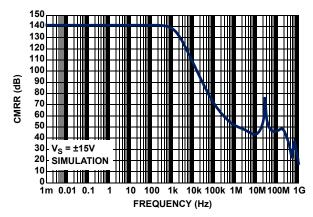


FIGURE 27. CMRR vs FREQUENCY, $V_S = \pm 15V$

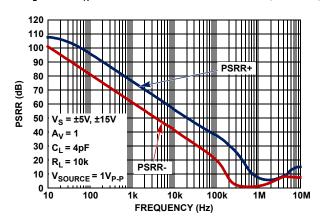


FIGURE 28. PSRR vs FREQUENCY, $V_S = \pm 5V$ AND $\pm 15V$

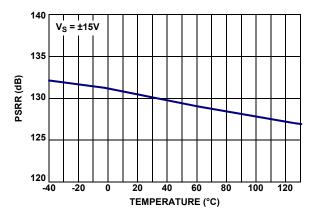


FIGURE 29. PSRR (DC) vs TEMPERATURE, $V_S = \pm 15V$

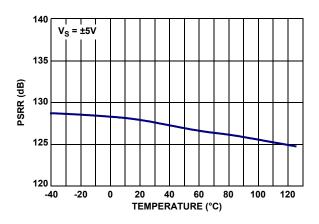


FIGURE 30. PSRR (DC) vs TEMPERATURE, $V_S = \pm 5V$

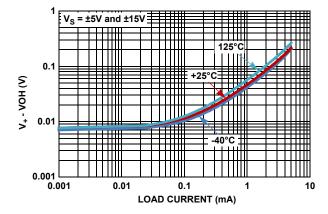


FIGURE 31. OUTPUT OVERHEAD VOLTAGE HIGH vs LOAD CURRENT, $\rm V_S = \pm 5V~AND~\pm 15V$

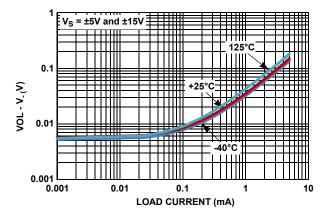


FIGURE 32. OUTPUT OVERHEAD VOLTAGE LOW vs LOAD CURRENT, $V_S = \pm 5 V \; \text{AND} \; \pm 15 V$

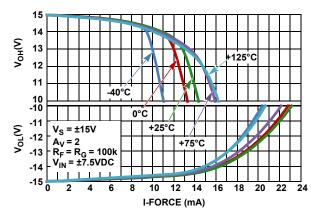


FIGURE 33. ISL28208 OUTPUT VOLTAGE SWING vs LOAD CURRENT $V_S = \pm 15 V$

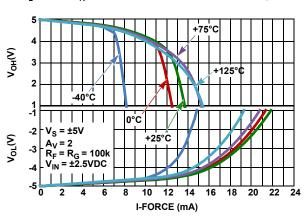


FIGURE 34. ISL28208 OUTPUT VOLTAGE SWING vs LOAD CURRENT $V_S = \pm 5V$

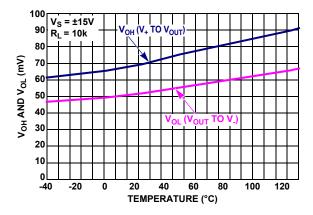


FIGURE 35. V_{OUT} HIGH AND LOW vs TEMPERATURE, $V_S = \pm 15V, R_I = 10k$

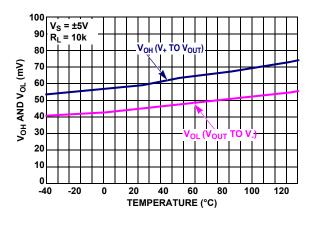


FIGURE 36. V_{OUT} HIGH AND LOW vs TEMPERATURE, $V_S = \pm 5V$, $R_I = 10k$

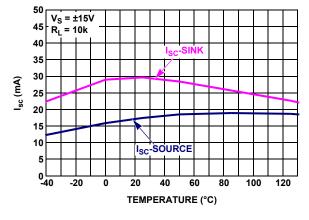


FIGURE 37. SHORT CIRCUIT CURRENT vs TEMPERATURE, $V_S = \pm 15V$

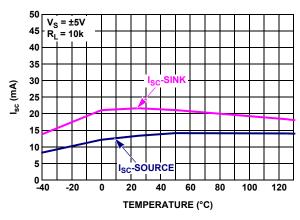


FIGURE 38. SHORT CIRCUIT CURRENT vs TEMPERATURE, $V_S = \pm 5V$

$\textbf{Typical Performance Curves} \quad v_{\text{S}} = \pm 15 \text{V}, \ v_{\text{CM}} = \text{OV}, \ R_{\text{L}} = \text{Open}, \ T_{\text{A}} = +25\,^{\circ}\text{C}, \ \text{unless otherwise specified}. \ \textbf{(Continued)}$

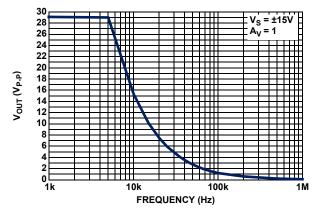


FIGURE 39. MAX OUTPUT VOLTAGE vs FREQUENCY

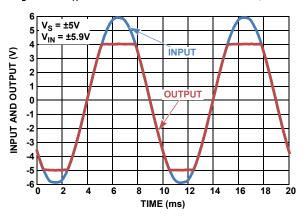


FIGURE 40. NO PHASE REVERSAL

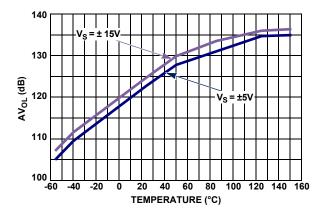


FIGURE 41. AV_{OL} vs TEMPERATURE

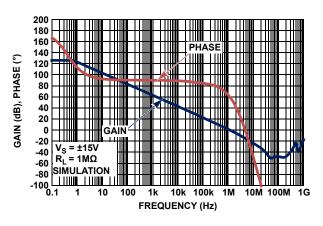


FIGURE 42. OPEN-LOOP GAIN, PHASE vs FREQUENCY, $V_S = \pm 15V$

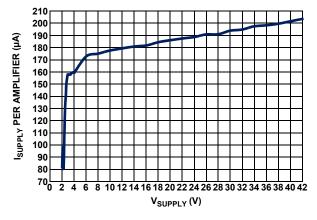


FIGURE 43. SUPPLY CURRENT vs SUPPLY VOLTAGE

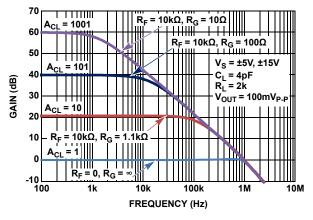


FIGURE 44. FREQUENCY RESPONSE vs CLOSED LOOP GAIN

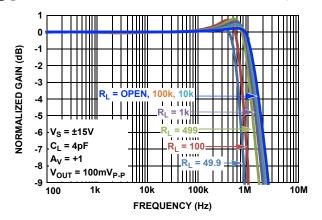


FIGURE 45. GAIN vs FREQUENCY vs R_L , $V_S = \pm 15V$

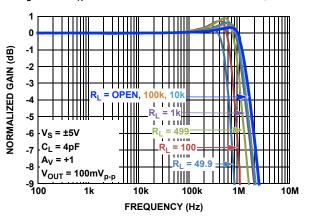


FIGURE 46. GAIN vs FREQUENCY vs R_L , $V_S = \pm 5V$

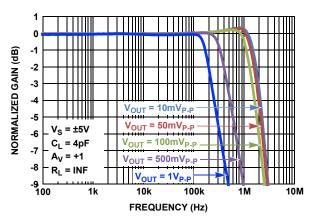


FIGURE 47. GAIN vs FREQUENCY vs OUTPUT VOLTAGE

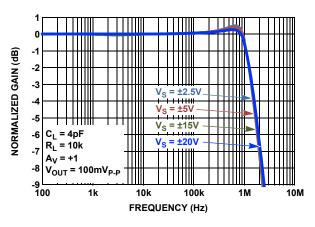


FIGURE 48. GAIN vs FREQUENCY vs SUPPLY VOLTAGE

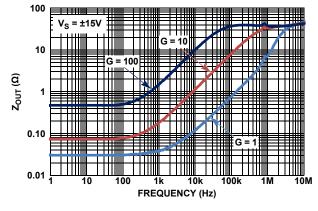


FIGURE 49. OUTPUT IMPEDANCE vs FREQUENCY, $V_S = \pm 15V$

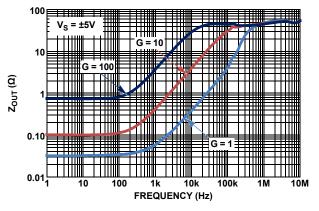
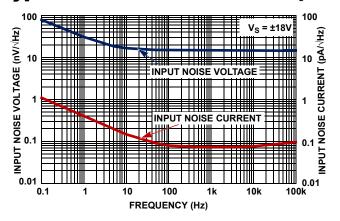


FIGURE 50. OUTPUT IMPEDANCE vs FREQUENCY, $V_S = \pm 5V$

$\textbf{Typical Performance Curves} \quad \textbf{v}_{\text{S}} = \pm 15 \text{V}, \textbf{v}_{\text{CM}} = \text{OV}, \textbf{R}_{\text{L}} = \text{Open}, \textbf{T}_{\text{A}} = +25\,^{\circ}\text{C}, \text{ unless otherwise specified}. \textbf{(Continued)}$

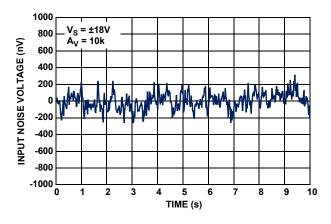


100 V_S = ±5V 100 V

FIGURE 51. INPUT NOISE VOLTAGE (en) AND CURRENT (in) vs FREQUENCY, $V_S = \pm 18V$

FIGURE 52. INPUT NOISE VOLTAGE (en) AND CURRENT (in) vs FREQUENCY, $V_S = \pm 5V$

1000



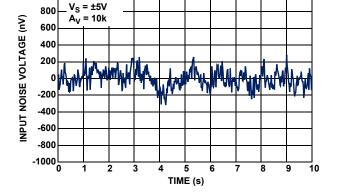
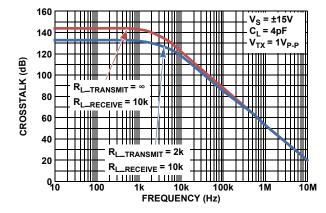


FIGURE 53. INPUT NOISE VOLTAGE 0.1Hz TO 10Hz, $V_S = \pm 18V$

FIGURE 54. INPUT NOISE VOLTAGE 0.1Hz TO 10Hz, $V_S = \pm 5V$



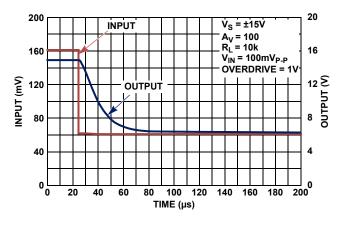
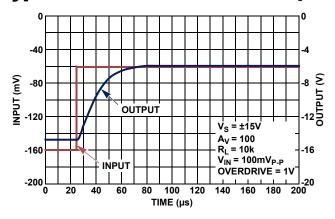


FIGURE 55. ISL28208 CHANNEL SEPARATION vs FREQUENCY, $V_S = \pm 5V, \pm 15V$

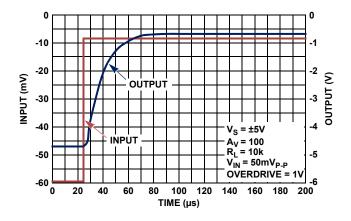
FIGURE 56. POSITIVE OUTPUT OVERLOAD RESPONSE TIME, $V_S = \pm 15V$



A_V = 100 R_L = 10k 50 INPUT 5 $V_{IN}^{\perp} = 50 \text{mV}_{P-P}$ OVERDRIVE = 40 4 (v) TUPTUO INPUT (mV) 30 OUTPUT 20 10 0 20 40 60 100 120 140 160 180 TIME (µs)

FIGURE 57. NEGATIVE OUTPUT OVERLOAD RESPONSE TIME, $V_S = \pm 15 \text{V} \label{eq:VS}$

FIGURE 58. POSITIVE OUTPUT OVERLOAD RESPONSE TIME, $V_S = \pm 5V$



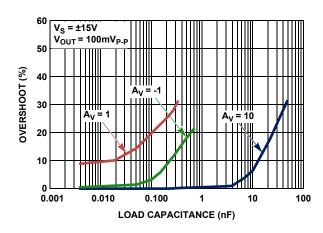
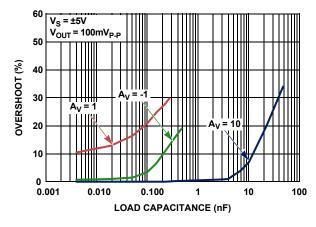


FIGURE 59. NEGATIVE OUTPUT OVERLOAD RESPONSE TIME, $\label{eq:VS} V_S = \pm 5 V$

FIGURE 60. OVERSHOOT vs CAPACITIVE LOAD, $V_S = \pm 15V$



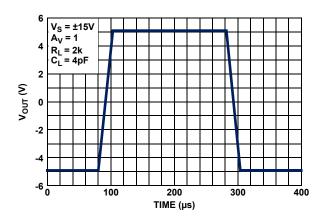
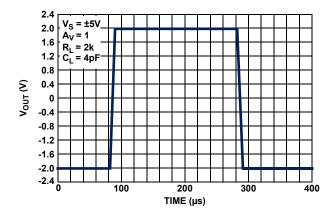


FIGURE 61. OVERSHOOT vs CAPACITIVE LOAD, $V_S = \pm 5V$

FIGURE 62. LARGE SIGNAL 10V STEP RESPONSE, $V_S = \pm 15V$

$\textbf{Typical Performance Curves} \quad v_{\text{S}} = \pm 15 \text{V}, \ v_{\text{CM}} = \text{OV}, \ R_{\text{L}} = \text{Open}, \ T_{\text{A}} = +25\,^{\circ}\text{C}, \ \text{unless otherwise specified}. \ \textbf{(Continued)}$



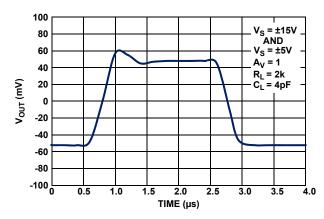


FIGURE 63. LARGE SIGNAL 4V STEP RESPONSE, $V_S = \pm 5V$

FIGURE 64. SMALL SIGNAL TRANSIENT RESPONSE $V_S = \pm 5V, \pm 15V$

Applications Information

Functional Description

The ISL28108, ISL28208, and ISL28408 are single, dual and quad, 1.2MHz, single supply rail-to-rail output amplifiers with a common mode input voltage range extending to a range of 0.5V below the V- rail. Their input stages are optimized for precision sensing of ground referenced signals in low voltage, single supply applications. The input stage has the capability of handling large input differential voltages without phase inversion making them suitable for high voltage comparator applications. Their bipolar design features high open loop gain and excellent DC input and output temperature stability. These op amps feature low quiescent current of 165µA, and a maximum temperature drift ranging from $1.1\mu V/^{\circ}C$ for the ISL28208 and ISL28408 in the SOIC package to $1.4\mu V/^{\circ}C$ for the ISL28208 in the TDFN package and the ISL28408 in the SOIC package (see Figures 11 through 20). All devices are fabricated in a new precision 40V complementary bipolar DI process and immune from latch-up.

Operating Voltage Range

The devices are designed to operate over the 3V ($\pm 1.5V$) to 40V ($\pm 20V$) range and are fully characterized at $\pm 5V$ and $\pm 15V$. Both DC and AC performance remain virtually unchanged over the $\pm 5V$ to $\pm 1.5V$ operating voltage range. Parameter variation with operating voltage is shown in the "Typical Performance Curves" beginning on page 10.

Input Stage Performance

The PNP input stage has a common mode input range extending up to 0.5V below ground at +25°C (see Figures 23 and 24). Full amplifier performance is guaranteed down to ground (V-) over the -40°C to +125°C temperature range. For common mode voltages down to -0.5V the amplifiers are fully functional, but performance degrades slightly over the full temperature range. This feature provides excellent CMRR, AC performance and DC accuracy when amplifying low level ground referenced signals.

The input stage has a maximum input differential voltage equal to a diode drop greater than the supply voltage (max 42V) and does not contain the back-to-back input protection diodes found on many similar amplifiers. This feature enables the device to function as a precision comparator by maintaining very high input impedance for high voltage differential input comparator voltages. The high differential input impedance also enables the device to operate reliably in large signal pulse applications without the need for anti-parallel clamp diodes required on MOSFET and most bipolar input stage op amps. Thus, input signal distortion caused by nonlinear clamps under high slew rate conditions are avoided.

In applications where one or both amplifier input terminals are at risk of exposure to voltages beyond the supply rails, current limiting resistors may be needed at each input terminal (see Figure 65, R_{IN} +, R_{IN} -) to limit current through the power supply ESD diodes to 20mA.

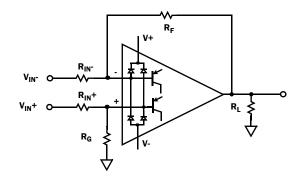


FIGURE 65. INPUT ESD DIODE CURRENT LIMITING

Output Drive Capability

The bipolar rail-to-rail output stage features low saturation levels that enable an output voltage swing to less than 10mV when the total output load (including feedback resistance) is held below $50\mu\text{A}$ (Figures 31 and 32). With $\pm15\text{V}$ supplies this can be achieved by using feedback resistor values >300k Ω . The low input bias and offset currents (-43nA and $\pm3\text{nA}$ +25°C max respectively) minimize DC offset errors at these high resistance values. For example, a balanced 4 resistor gain circuit (Figure 65) with $1\text{M}\Omega$ feedback resistors (RF, RG) generates a worst case input offset error of only $\pm3\text{mV}$. Furthermore, the low noise current reduces the added noise associated with high feedback resistance.

The output stage is internally current limited. Output current limit over-temperature is shown in Figures 37 and 38. The amplifiers can withstand a short circuit to either rail as long as the power dissipation limits are not exceeded. This applies to only one amplifier at a time for the dual op amp. Continuous operation under these conditions may degrade long-term reliability.

The amplifiers perform well driving capacitive loads (Figures 60 and 61). The unity gain, voltage follower (buffer) configuration provides the highest bandwidth, but is also the most sensitive to ringing produced by load capacitance found in BNC cables. Unity gain overshoot is limited to 30% at capacitance values to 0.33nF. At gains of 10 and higher, the device is capable of driving more than 10nF without significant overshoot.

Output Phase Reversal

Output phase reversal is a change of polarity in the amplifier transfer function when the input voltage exceeds the supply voltage. These devices are immune to output phase reversal, out to 0.5V beyond the rail ($V_{ABS\ MAX}$) limit (see Figure 40).



Unused Channels

If the application requires only one channel, the user must configure any unused channel to prevent it from oscillating. Unused channels can oscillate if the input and output pins are floating. This will result in higher-than-expected supply currents and possible noise injection into the channel being used. The proper way to prevent oscillation is to short the output to the inverting input, and ground the positive input (Figure 66).



FIGURE 66. PREVENTING OSCILLATIONS IN UNUSED CHANNELS

Power Dissipation

It is possible to exceed the $+150\,^{\circ}$ C maximum junction temperatures under certain load and power supply conditions. It is therefore important to calculate the maximum junction temperature (T_{JMAX}) for all applications to determine if power supply voltages, load conditions, or package type need to be modified to remain in the safe operating area. These parameters are related using Equation 1:

$$T_{JMAX} = T_{MAX} + \theta_{JA} x PD_{MAXTOTAL}$$
 (EQ. 1)

where:

- PD_{MAXTOTAL} is the sum of the maximum power dissipation of each amplifier in the package (PD_{MAX})
- PD_{MAX} for each amplifier can be calculated using Equation 2:

$$PD_{MAX} = V_S \times I_{qMAX} + (V_S - V_{OUTMAX}) \times \frac{V_{OUTMAX}}{R_I}$$
 (EQ. 2)

where:

- T_{MAX} = Maximum ambient temperature
- θ_{IA} = Thermal resistance of the package
- PD_{MAX} = Maximum power dissipation of 1 amplifier
- V_S = Total supply voltage
- IqMAX = Maximum quiescent supply current of 1 amplifier
- V_{OUTMAX} = Maximum output voltage swing of the application
- R_L = Load resistance

ISL28108, ISL28208, ISL28408 SPICE Model

Figure 67 shows the SPICE model schematic and Figure 68 shows the net list for the SPICE model. The model is a simplified version of the actual device and simulates important AC and DC parameters. AC parameters incorporated into the model are: 1/f and flat band noise voltage, Slew Rate, CMRR, Gain and Phase. The DC parameters are I_{OS} , total supply current and output voltage swing. The model uses typical parameters given in the "Electrical Specifications" Table beginning on page 6. The AVOL is adjusted for 122dB with the dominant pole at 1Hz. The CMRR is set 128dB, f=6kHz. The input stage models the actual device to present an accurate AC representation. The model is configured for ambient temperature of +25 °C.

Figures 69 through 83 show the characterization vs simulation results for the Noise Voltage, Open Loop Gain Phase, Closed Loop Gain vs Frequency, Gain vs Frequency vs R_L , CMRR, Large Signal 10V Step Response, Small Signal 0.05V Step and Output Voltage Swing ± 15 V supplies.

LICENSE STATEMENT

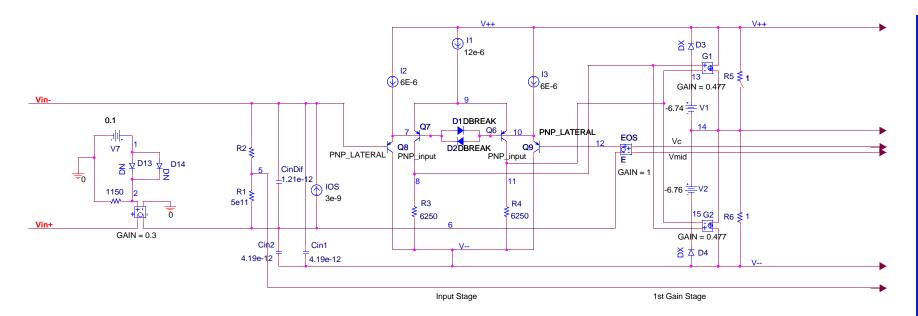
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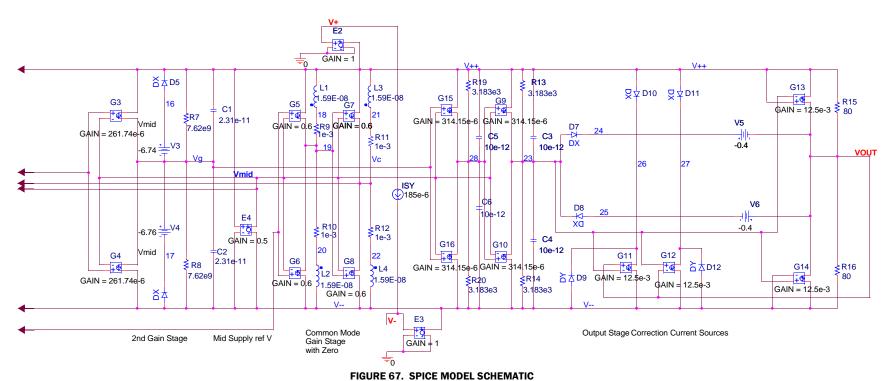
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V-- 28 10e-12
*ISL28108_208 Macromodel - covers following
                                                 R_R17
                                                             2 0 1150
                                                                                                   C_C6
*products
*ISL28108
                                                 *Input Stage
                                                                                                   G_G9
                                                                                                             V++ 23 28 VMID 314.15e-6
*ISL28208
                                                 Q_Q6
                                                            11 10 9 PNP_input
                                                                                                   G_G10
                                                                                                              V-- 23 28 VMID 314.15e-6
*ISL28408
                                                 Q_Q7
                                                            879 PNP_input
                                                                                                   R R13
                                                                                                              23 V++ 3.18319e3
                                                 Q Q8
                                                            V-- VIN- 7 PNP LATERAL
                                                                                                   R R14
                                                                                                              V-- 23 3.18319e3
*Revision History:
                                                 Q_Q9
                                                            V-- 12 10 PNP_LATERAL
                                                                                                  C_C3
                                                                                                             23 V++ 10e-12
* Revision B, LaFontaine January 22 2014
                                                                                                             V-- 23 10e-12
                                                 I_I1
                                                          V++ 9 DC 12e-6
                                                                                                   C_C4
* Model for Noise, supply currents, CMRR
*128dB f=6kHz ,AVOL 122dB f=1Hz
                                                 I_I2
                                                          V++ 7 DC 6E-6
                                                 I 13
                                                          V++ 10 DC 6E-6
                                                                                                   *Output Stage with Correction Current Sources
* SR = 0.45V/us, GBWP 1.2MHz.
                                                 I_IOS
                                                            6 VIN- DC 3e-9
                                                                                                   G_G11
                                                                                                              26 V-- VOUT 23 12.5e-3
*Copyright 2011 by Intersil Corporation
                                                 D D1
                                                            7 10 DBREAK
                                                                                                   G G12
                                                                                                              27 V-- 23 VOUT 12.5e-3
*Refer to data sheet "LICENSE STATEMENT"
                                                 D_D2
                                                            107 DBREAK
                                                                                                  G_G13
                                                                                                              VOUT V++ V++ 23 12.5e-3
*Use of this model indicates your acceptance
                                                 R_R1
                                                            5 6 5e11
                                                                                                  G_G14
                                                                                                              V-- VOUT 23 V-- 12.5e-3
*with the terms and provisions in the License
                                                 R_R2
                                                            VIN-5 5e11
                                                                                                   D_D7
                                                                                                             23 24 DX
*Statement.
                                                 R R3
                                                            V-- 8 6250
                                                                                                   D D8
                                                                                                             25 23 DX
                                                 R_R4
                                                            V-- 11 6250
                                                                                                   D_D9
                                                                                                             V-- 26 DY
*Intended use:
                                                 C Cin1
                                                             V-- VIN- 4.19e-12
                                                                                                  D D10
                                                                                                              V++ 26 DX
*This Pspice Macromodel is intended to give
*typical DC and AC performance characteristics
                                                 C_Cin2
                                                             V-- 6 4.19e-12
                                                                                                   D_D11
                                                                                                              V++ 27 DX
*under a wide range of external circuit
                                                 C_CinDif
                                                              6 VIN- 1.21E-12
                                                                                                   D_D12
                                                                                                              V-- 27 DY
*configurations using compatible simulation
                                                                                                   V_V5
                                                                                                             24 VOUT -0.4
*platforms – such as iSim PE.
                                                 *1st Gain Stage
                                                                                                   V V6
                                                                                                             VOUT 25 -0.4
                                                 G_G1
                                                            V++ 14 8 11 0.4779867
                                                                                                   R_R15
                                                                                                              VOUT V++ 80
*Device performance features supported by this
                                                 G G2
                                                            V-- 14 8 11 0.4779867
                                                                                                   R R16
                                                                                                              V-- VOUT 80
*model
                                                 V_V1
                                                           13 14 -6.74
                                                                                                   .model PNP_LATERAL pnp(is=1e-016 bf=250
*Typical, room temp., nominal power supply
                                                                                                  va=80
                                                 V_V2
                                                           14 15 -6.76
*voltages used to produce the following
                                                                                                   + ik=0.138 rb=0.01 re=0.101 rc=180 kf=0 af=1)
*characteristics:
                                                 D_D3
                                                            13 V++ DX
                                                                                                   .model PNP_input pnp(is=1e-016 bf=100
                                                 D D4
                                                            V-- 15 DX
*Open and closed loop I/O impedances,
                                                                                                  va=80
*Open loop gain and phase,
                                                 R_R5
                                                            14 V++ 1
                                                                                                   + ik=0.138 rb=0.01 re=0.101 rc=180 kf=0 af=1)
                                                 R_R6
                                                            V-- 14 1
*Closed loop bandwidth and frequency
                                                                                                   .model DBREAK D(bv=43 rs=1)
*response.
                                                                                                   .model DN D(KF=6.69e-9 AF=1)
*Loading effects on closed loop frequency
                                                 *2nd Gain Stage
                                                                                                   .MODEL DX D(IS=1E-12 Rs=0.1)
                                                 G_G3
                                                            V++ VG 14 VMID 261.748e-6
                                                                                                   .MODEL DY D(IS=1E-15 BV=50 Rs=1)
*Input noise terms including 1/f effects,
                                                 G G4
                                                            V-- VG 14 VMID 261.748e-6
*Slew rate,
                                                                                                   .ends ISL28108_208
                                                 V_V3
                                                           16 VG -6.74
*Input and Output Headroom limits to I/O
                                                 V_V4
                                                           VG 17 -6.76
*voltage swing,
                                                 D_D5
                                                            16 V++ DX
*Supply current at nominal specified supply
                                                 D D6
                                                            V-- 17 DX
                                                 R_R7
                                                            VG V++ 7.62283e9
                                                 R R8
                                                            V-- VG 7.62283e9
*Device performance features NOT supported
                                                 C_C1
                                                            VG V++ 2.31e-11
*by this model:
                                                 C_C2
                                                            V-- VG 2.31e-11
*Harmonic distortion effects,
*Output current limiting (current will limit at
                                                 *Mid supply Ref
*40mA),
                                                 E_E2
                                                           V++ 0 V+ 0 1
*Disable operation (if any),
                                                 E E3
                                                           V-- 0 V- 0 1
*Thermal effects and/or over temperature
                                                           VMID V-- V++ V-- 0.5
*parameter variation.
                                                 E_E4
*Limited performance variation vs. supply
                                                 I_ISY
                                                           V+ V- DC 185E-6
*voltage is modeled,
*Part to part performance variation due to
                                                 *Common Mode Gain Stage with Zero
*normal process parameter spread,
                                                 G G5
                                                            V++ 19 5 VMID 0.6
*Any performance difference arising from
                                                            V-- 19 5 VMID 0.6
                                                 G_G6
*different packaging source,
                                                 G_G7
                                                            V++ VC 19 VMID 0.6
*Load current reflected into the power supply
                                                 G G8
                                                            V-- VC 19 VMID 0.6
*current.
                                                 E_EOS
                                                              126 VC VMID 1
                                                 L_L1
                                                           18 V++ 1.59159F-08
* Connections:
                    +input
                                                 L L2
                                                           20 V-- 1.59159E-08
                           -input
                                                 L_L3
                                                           21 V++ 1.59159E-08
                                                 L_L4
                                                           22 V-- 1.59159E-08
                              +Vsupply
                                  -Vsupply
                                                 R R9
                                                            19 18 1e-3
                                                 R_R10
                                                             20 19 1e-3
                                     output
                                                 R_R11
                                                             VC 21 1e-3
.subckt ISL28108_208 Vin+ Vin-V+ V- VOUT
                                                 R_R12
                                                             22 VC 1e-3
* source ISL28118_218_subckt_check_0
                                                 *Pole Stage
*Voltage Noise
                                                 G_G15
                                                             V++ 28 VG VMID 314.15e-6
E_En
          VIN+6200.3
                                                 G_G16
                                                             V-- 28 VG VMID 314.15e-6
```

FIGURE 68. SPICE NET LIST

28 V++ 3.18319e3

V-- 28 3.18319e3

28 V++ 10e-12



R_R19

R_R20

 C_C5

12 DN

12 DN

100.1

D_D13

D_D14

V_V7

Characterization vs Simulation Results

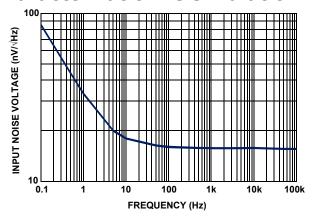


FIGURE 69. CHARACTERIZED INPUT NOISE VOLTAGE

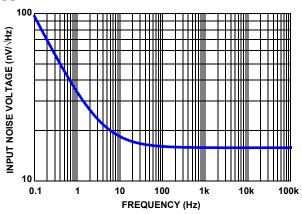


FIGURE 70. SIMULATED INPUT NOISE VOLTAGE

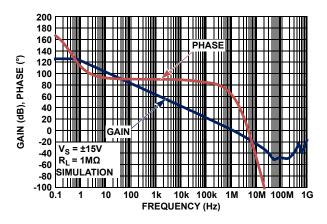


FIGURE 71. CHARACTERIZED OPEN-LOOP GAIN, PHASE vs FREQUENCY

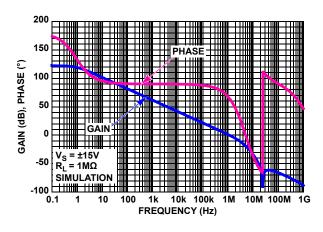


FIGURE 72. SIMULATED OPEN-LOOP GAIN, PHASE vs FREQUENCY

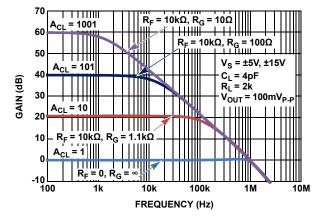


FIGURE 73. CHARACTERIZED CLOSED LOOP GAIN vs FREQUENCY

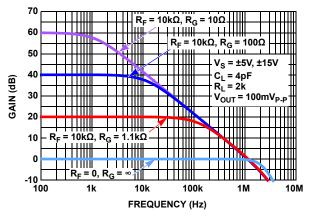


FIGURE 74. SIMULATED CLOSED LOOP GAIN vs FREQUENCY

Characterization vs Simulation Results (Continued)

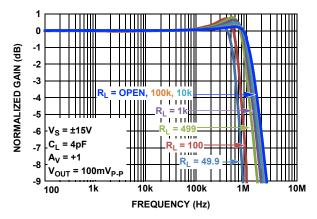


FIGURE 75. CHARACTERIZED GAIN vs FREQUENCY vs RL

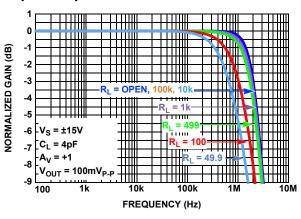


FIGURE 76. SIMULATED GAIN vs FREQUENCY vs RL

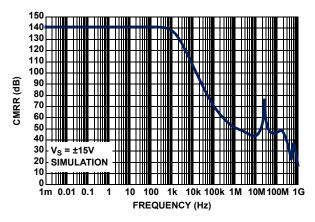


FIGURE 77. CHARACTERIZED CMRR vs FREQUENCY

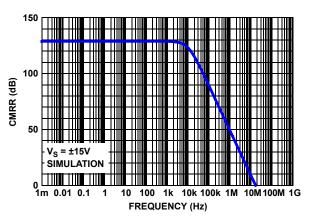


FIGURE 78. SIMULATED CMRR vs FREQUENCY

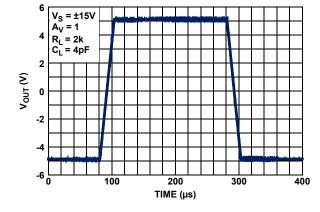


FIGURE 79. CHARACTERIZED LARGE SIGNAL 10V STEP RESPONSE

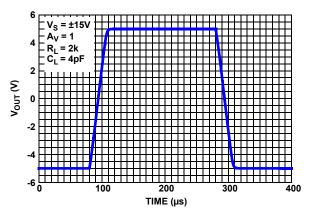
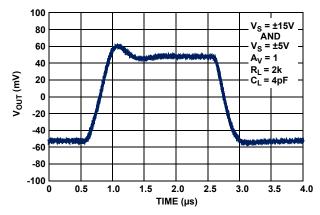


FIGURE 80. SIMULATED LARGE SIGNAL 10V STEP RESPONSE

Characterization vs Simulation Results (Continued)



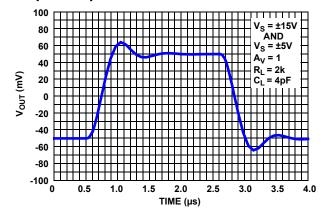


FIGURE 81. CHARACTERIZED SMALL SIGNAL TRANSIENT RESPONSE

FIGURE 82. SIMULATED SMALL SIGNAL TRANSIENT RESPONSE

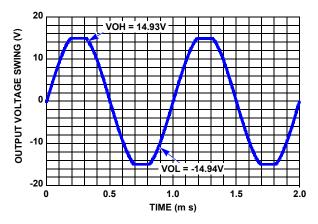


FIGURE 83. SIMULATED OUTPUT VOLTAGE SWING

Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest Rev.

DATE	REVISION	CHANGE
Feb 6, 2020	5.01	Updated links throughout document. Updated related literature section. Updated ordering information table by adding tape and reel information, updating note 1, and removing ISL28108FUZ. Remove About Intersil section. Updated POD L8.3X3K to the latest revision, changes are as follows: Tiebar Note 5 updated From: Tiebar shown (if present) is a non-functional feature. To: Tiebar shown (if present) is a non-functional feature and may be located on any of the 4 sides (or ends). Updated disclaimer.
Mar 4, 2014	5.00	Updated absolute Maximum Ratings table on page 6, by adding ISL28108 MSOP value to ESD tolerance. Updated Spice model netlisl on page 24.
Jan 22, 2013	4.00	 Added ISL28208 MSOP specifications. Moved Table of Contents to page 2. Removed "Coming Soon" from ISL28208FUZ entry in "Ordering Information" table on page 3. page 6, removed package indicators on Vos entries for ISL28208 and ISL28408; added ISL28208 MSOP entries for TCVos and deltaVos. page 7, removed package indicators on Ios entries for ISL28208 and ISL28408. page 8, removed package indicators on Vos and Ios entries for ISL28208 and ISL28408; added ISL28208 MSOP entries for TCVos, and added "ISL28108, ISL28208" to 4th and 5th rows of CMRR parameter. page 8, added ISL28208 MSOP entries for PSRR. page 10, added "+25°C" to "default" info at top of page. page 12, added "MSOP" to titles of Figures 15 and 16.

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest Rev.

DATE	REVISION	CHANGE
DATE Oct 19, 2011	3.00	 On page 6, ±15V Electrical Specifications table, added the following parameters: VOS ISL28108 SOIC, TDFN TCVOS ISL28108 SOIC, TDFN IOS ISL28108 SOIC, TDFN On page 6 and 6, ±15V Electrical Specifications table, changed the following in Conditions column: VOS: changed "ISL28208 SOIC, TDFN" to "ISL28208 SOIC, TDFN; ISL28408 SOIC" TCVOS: changed "ISL28208 TDFN" to "ISL28208 TDFN, ISL28408 SOIC" IOS: changed "ISL28108 SOIC, TDFN" to "ISL28108 SOIC, TDFN; ISL28408 SOIC" On page 8, ±5V Electrical Specifications table, added the following: VOS ISL28108 SOIC, TDFN TCVOS ISL28108 SOIC, TDFN IOS ISL28108 SOIC, TDFN On page 8, ±5V Electrical Specifications table, changed the following in Conditions column: VOS: changed "ISL28208 SOIC, TDFN" to "ISL28208 SOIC, TDFN; ISL28408 SOIC" TCVOS: changed "ISL28208 TDFN" to "ISL28208 SOIC, TDFN; ISL28408 SOIC" IOS: changed "ISL28208 TDFN" to "ISL28208 TDFN, ISL28408 SOIC" On page 10 through page 12, added the following to Typical Performance Curves: Figures 3, 4: ISL28408 SOIC ±15V VOS distribution Voltage, ±15V and ±5V Figures 9, 10: ISL28108 TDFN ±15V VOS distribution, and ±5V VOS distribution Figures 17, 18: ISL28108 SOIC ±15V TCVOS vs. number of Amplifiers ±15V and ±5V Figures 17, 18: ISL28108 SOIC ±15V TCVOS distribution, and ±5V TCVOS distribution Figures 17, 18: ISL28108 SOIC ±15V TCVOS distribution, and ±5V TCVOS distribution On page 22. changed Package Outline Drawing M3.118 On page 33: changed Package Outline Drawing M8.118 On page 34: added Package Outline Drawing M8.118 On page 37: changed Package Outline Drawing M8.118 On page 38: changed Package Outline Drawing M8.118 On page 39: changed Package Outline Drawing M8.118 On page 30: changed Package Outline Drawing M8.118 On page 30: changed Package Outline D
		 Dwg. # for MSOP package from M8.118 to M8.118B. On page 6, Absolute Maximum Ratings, changed "ESD Tolerance (ISL28208)" to "ESD Tolerance (ISL28208, ISL28408)". Added ESD information for ISL28108 as follows: ESD Tolerance (ISL28108) Human Body Model (Tested per JESD22-A114F)5.5kV Machine Model (Tested per JESD22-A115-C)300V Charged Device Model (Tested per JESD22-C110D)2kV
		• On page 6, Thermal Information, changed package temperatures from : 8 Ld SOIC Package (108, 208, Notes 4, 7), θ_{JA} = 120, θ_{JC} = 55 8 Ld TDFN Package (108, 208, Notes 5, 6), θ_{JA} = 47, θ_{JC} = 6 8 Ld MSOP Package (108, 208, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 45 14 Ld SOIC Package (408, Notes 4, 7), θ_{JA} = -, θ_{JA} = - To: 8 Ld SOIC Package (208, Notes 4, 7), θ_{JA} = 120, θ_{JC} = 55 and 8 Ld SOIC Package (108, Notes 4, 7), θ_{JA} = 120, θ_{JC} = 60 8 Ld TDFN Package (208, Notes 5, 6), θ_{JA} = 47, θ_{JC} = 6 and 8 Ld TDFN Package (108, Notes 5, 6), θ_{JA} = 45 θ_{JC} = 3.5 8 Ld MSOP Package (208, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JA} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JC} = 150, θ_{JC} = 50 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JC} = 150, θ_{JC} = 150 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JC} = 150 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JC} = 150 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JC} = 150 and 8 Ld MSOP Package (108, Notes 4, 7), θ_{JC} = 150 and
		165, θ_{JC} = 57 14 Ld SOIC Package (408, Notes 4, 7), θ_{JA} = 71, θ_{JC} = 37



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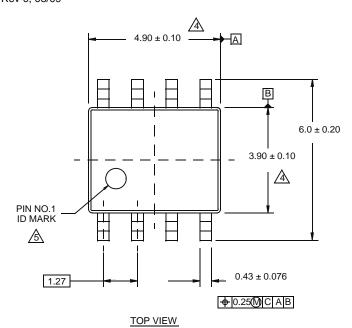
DATE	REVISION	CHANGE
Apr 20, 2011	2.00	Added discussion of ISL28408 throughout datasheet.
		On page 3 in "Ordering Information": Added new part, "ISL28408FBZ". Corrected part marking for ISL28208FRTZ from 208Z to 208F. Added "ISL28408" to Note 3. Under" Pin Configurations," added ISL28408 (14 Ld SOIC) pin configuration diagram.
		On page 5: in Pin Descriptions table, added column for ISL28408 14Ld SOIC. Corrected schematic for Circuit 2.
		• On page 6: under "Thermal Information" added "14 Ld SOIC Package (408, Notes 4, 7)" and added ISL28108 to 8 Ld TDFN and 8 Ld MSOP. Changed θ_{JA} and θ_{JC} for 8 Ld TDFN Package from 48 and 5.5 to 4 and 6. Added Note 6 regarding θ_{JC} "case temp" measurement, and applied it to 8 Ld TDFN Package.
		 On page 6: in Electrical Specifications table, changed TYP spec for TCl_B from 70 pA/° C to 0.07nA/° C. C page 8, changed TYP spec for TCl_B from -67 pA/° C to -0.067nA/° C. These are not spec changes since the values are the same.
		 On page 14, Figs. 31 and 32: changed y axis units label from (mV) to (V); changed x axis units label from (μA) to (mA).
		 On page 21, under "Output Drive Capability," para 2, changed "The output stage can swing at moderate levels of output current (Figures 21 and 22) and the output stage is internally current limited. Output current limit over-termperature" to "The output stage is internally current limited. Output current limit over-temperature"
Mar 11, 2011	1.00	On page 1, in the first paragraph - added the following after V-rail: "a rail-to-rail differential input voltage range for use as a comparator,"
		On page 1 in "Features:
		- Added bullet - "Rail-to-rail input differential voltage range for comparator applications"
		- Changed Low Noise Current from "100fA/sq.root Hz" to "80fA/sq.root Hz"
		On page 3 in "Ordering Information" - Removed "coming soon" from ISL28208FRTZ part since it is releasing.
		On page 6, changed "ESD Tolerance (ISL28208, ISL28408)" as follows:
		- Human Body Model changed from "3kV" to "6kV"
		- Machine Model changed from "300V" to "400V"
		- Added JEDEC Test information for all ESD ratings
		On page 6 and page 8, added test conditions for SOIC TCVos specs. Added TCVos specs for TDFN.
		On page 7 changed "Noise Current Density" Typical from "100" to "80"
		On page 21, updated Applications Information Functional Description
		On page 21 Updated Input Stage Performance Section
		On page 21 Updated Output Drive Capability Section
		On page 22 Added ISL28108 AND ISL28208 SPICE MODEL and License Agreement section
		On page 23 Added SPICE NET LIST
		On page 25 Added Characterization vs Simulation Results curves
Feb 16, 2011	0.00	Initial Release

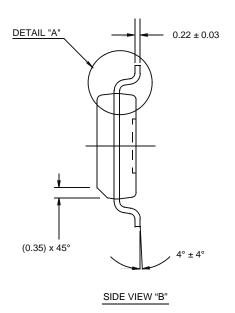


Package Outline Drawings

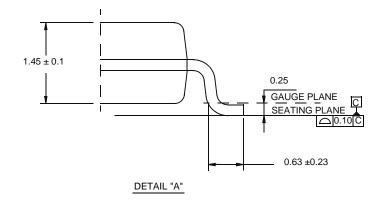
For the most recent package outline drawing, see M8.15E.

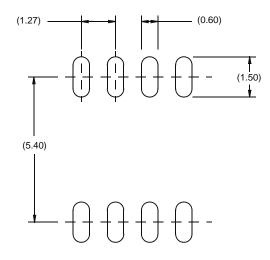
8 LEAD NARROW BODY SMALL OUTLINE PLASTIC PACKAGE Rev 0, 08/09





1.75 MAX 0.175 ± 0.075 SIDE VIEW "A





TYPICAL RECOMMENDED LAND PATTERN

- 1. Dimensions are in millimeters. Dimensions in () for Reference Only.
- Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
- Unless otherwise specified, tolerance : Decimal ± 0.05 3.
- Dimension does not include interlead flash or protrusions. Interlead flash or protrusions shall not exceed 0.25mm per side.
- 5. The pin #1 identifier may be either a mold or mark feature.
- Reference to JEDEC MS-012.

L8.3x3K

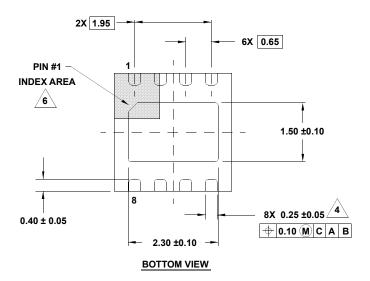
8 LEAD THIN DUAL FLAT NO-LEAD PLASTIC PACKAGE Rev 2, 5/15

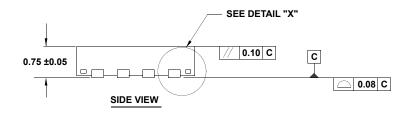
3.00
A
B
B
INDEX AREA

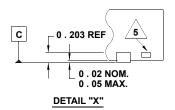
(4X) 0.15

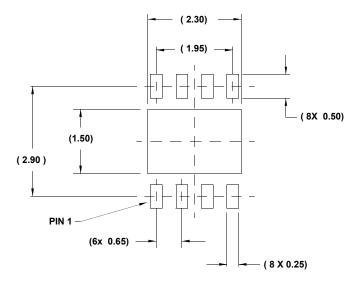
TOP VIEW

For the most recent package outline drawing, see <u>L8.3x3K</u>.









TYPICAL RECOMMENDED LAND PATTERN

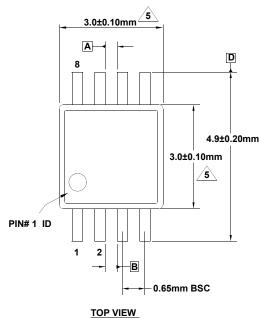
- Dimensions are in millimeters.
 Dimensions in () for Reference Only.
- 2. Dimensioning and tolerancing conform to ASME Y14.5m-1994.
- 3. Unless otherwise specified, tolerance : Decimal ± 0.05
- 4. Dimension applies to the metallized terminal and is measured between 0.15mm and 0.20mm from the terminal tip.
- 5. Tiebar shown (if present) is a non-functional feature and may be located on any of the 4 sides (or ends).
 - The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.
 - 7. Compliant to JEDEC MO-229 WEEC-2 except for the foot length.

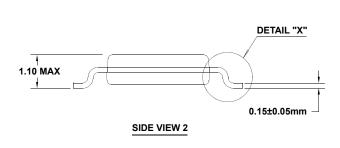
M8.118B

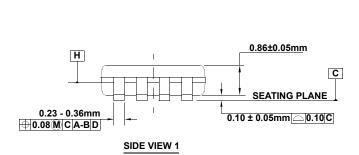
8 LEAD MINI SMALL OUTLINE PLASTIC PACKAGE

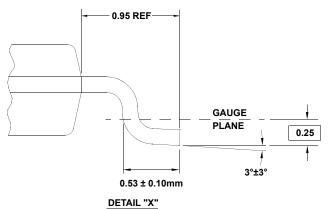
Rev 1, 3/12

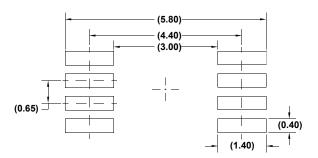
For the most recent package outline drawing, see M8.118B.









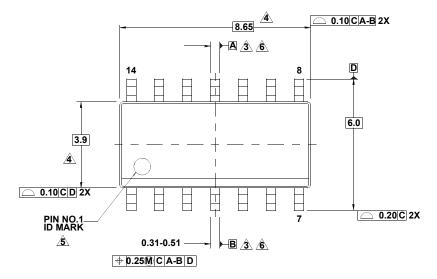


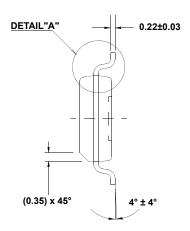
TYPICAL RECOMMENDED LAND PATTERN

- 1. Dimensions are in millimeters.
- Dimensioning and tolerancing conform to JEDEC MO-187-AA and AMSEY14.5m-1994.
- Plastic or metal protrusions of 0.15mm max per side are not included.
- 4. Plastic interlead protrusions of 0.15mm max per side are not included.
- 5. Dimensions are measured at Datum Plane "H".
- 6. Dimensions in () are for reference only.

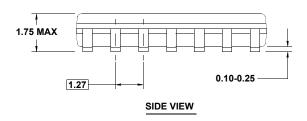
M14.15 14 LEAD NARROW BODY SMALL OUTLINE PLASTIC PACKAGE Rev 1, 10/09

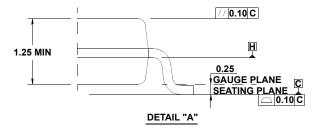
For the most recent package outline drawing, see M14.15.

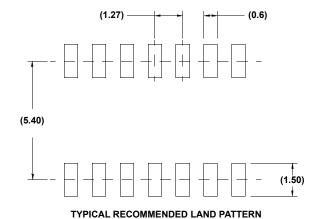




TOP VIEW







- 1. Dimensions are in millimeters. Dimensions in () for Reference Only.
- 2. Dimensioning and tolerancing conform to AMSEY14.5m-1994.
- 3. Datums A and B to be determined at Datum H.
- 4. Dimension does not include interlead flash or protrusions. Interlead flash or protrusions shall not exceed 0.25mm per side.
- 5. The pin #1 indentifier may be either a mold or mark feature.
- 6. Does not include dambar protrusion. Allowable dambar protrusion shall be 0.10mm total in excess of lead width at maximum condition.
- 7. Reference to JEDEC MS-012-AB.

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