

ISL28177

40V General Purpose Precision Operational Amplifier

FN7859  
Rev 2.00  
April 5, 2012

The ISL28177 is an OP07 replacement featuring low input offset voltage, low input bias current, and competitive noise and AC performance. The ESD ratings are best among competitive parts at 5kV HBM, 300V MM, and 2.2kV CDM. The amplifier operates over the 6V (±3V) to 40V (±20V) range.

Applications include precision active filters, medical and analytical instrumentation, precision power supply controls, and industrial sensors.

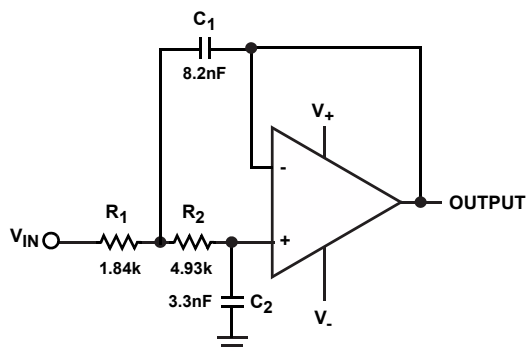
The ISL28177 is available in the SOT23-5 and SOIC-8 packages and operates over the extended temperature range to -40°C to +125°C.

Features

- Wide Supply Range ..... 6V (±3V) to 40V (±20V)
- Low Input Offset Voltage ..... 150µV, Max
- Input Bias Current ..... 1nA, Max
- Low Noise ..... 9.5nV/√Hz @ 1kHz
- Gain Bandwidth ..... 600kHz
- Exceptional ESD Performance ..... 5kV HBM, 300V MM, 2.2kV CDM
- Operating Temperature Range ..... -40°C to +125°C
- Packages
  - ISL28177 (Single) ..... SOT23-5, SOIC-8

Applications

- Precision Active Filters
- Medical and Analytical Instrumentation
- Precision Power Supply Controls
- Industrial Sensors



SALLEN-KEY LOW PASS FILTER (10kHz)

FIGURE 1. TYPICAL APPLICATION

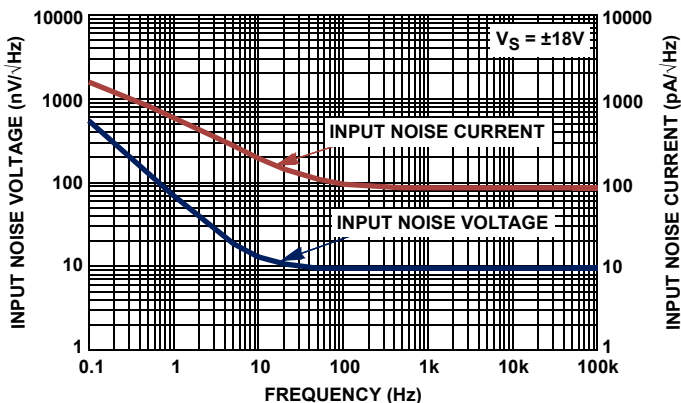


FIGURE 2. INPUT NOISE PERFORMANCE

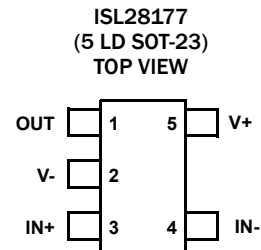
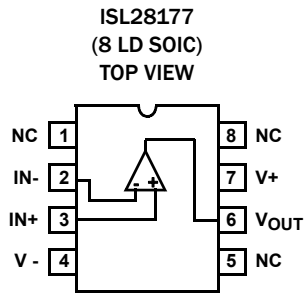
## Ordering Information

PART NUMBER (Note 2, 3)	PART MARKING	TEMP RANGE (°C)	PACKAGE (Pb-free)	PKG. DWG. #
ISL28177FBZ	28177 FBZ	-40 to +125	8 Ld SOIC	M8.15E
ISL28177FBZ-T13 (Note 1)	28177 FBZ	-40 to +125	8 Ld SOIC	M8.15E
ISL28177FBZ-T7 (Note 1)	28177 FBZ	-40 to +125	8 Ld SOIC	M8.15E
ISL28177FBZ-T7A (Note 1)	28177 FBZ	-40 to +125	8 Ld SOIC	M8.15E
Coming Soon ISL28177FHZ	TBD	-40 to +125	SOT23-5	P5.064A

**NOTES:**

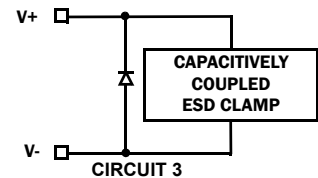
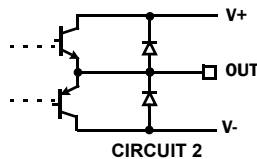
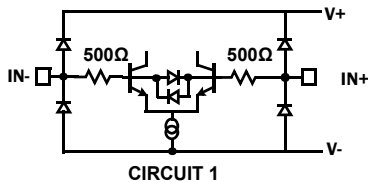
1. Please refer to [TB347](#) for details on reel specifications.
2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
3. For Moisture Sensitivity Level (MSL), please see device information page for [ISL28177](#). For more information on MSL please see techbrief [TB363](#).

## Pin Configurations



## Pin Descriptions

ISL28177 (8 LD SOIC)	ISL28177 (5 LD SOT-23)	PIN NAME	EQUIVALENT CIRCUIT	DESCRIPTION
3	3	IN+	Circuit 1	Amplifier non-inverting input
4	2	V-	Circuit 3	Negative power supply
2	4	IN-	Circuit 1	Amplifier inverting input
7	5	V+	Circuit 3	Positive power supply
6	1	V <sub>OUT</sub>	Circuit 2	Amplifier output
1, 5, 8	-	NC	-	No internal connection



## Absolute Maximum Ratings

Maximum Supply Voltage	44V
Maximum Differential Input Voltage	44V or $V_- - 0.5V$ to $V_+ + 0.5V$
Min/Max Input Voltage	44V or $V_- - 0.5V$ to $V_+ + 0.5V$
Min/Max Input Current	20mA
Output Short-Circuit Duration (1 output at a time)	Indefinite
<b>ESD Ratings</b>	
Human Body Model (Tested per JESD22-A114F)	5kV
Machine Model (Tested per JESD22-A115-A)	300V
Charged Device Model (Tested per CDM-22C10ID)	2.2kV

## Thermal Information

Thermal Resistance (Typical)	$\theta_{JA}$ ( $^{\circ}C/W$ )	$\theta_{JC}$ ( $^{\circ}C/W$ )
5 Ld SOT-23 Package (Notes 4, 5)	TBD	TBD
8 Ld SOIC Package (Notes 4, 5)	125	73
Storage Temperature Range	-65 $^{\circ}C$ to +150 $^{\circ}C$	
Pb-free Reflow Profile	see link below	
	<a href="http://www.intersil.com/pbfree/Pb-FreeReflow.asp">http://www.intersil.com/pbfree/Pb-FreeReflow.asp</a>	

## Operating Conditions

Ambient Operating Temperature Range	-40 $^{\circ}C$ to +125 $^{\circ}C$
Maximum Operating Junction Temperature	+150 $^{\circ}C$
Operating Voltage Range	.6V ( $\pm 3V$ ) to 40V ( $\pm 20V$ )

**CAUTION:** Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

### NOTES:

- $\theta_{JA}$  is measured with the component mounted on a high effective thermal conductivity test board in free air. See Tech Brief [TB379](#) for details.
- For  $\theta_{JC}$ , the "case temp" location is taken at the package top center.

## Electrical Specifications

$V_S = \pm 5V$  to  $\pm 15V$ ,  $R_L = \text{Open}$ ,  $V_{CM} = 0V$ ,  $T_A = +25^{\circ}C$ , unless otherwise specified. **Boldface limits apply over the operating temperature range, -40 $^{\circ}C$  to +125 $^{\circ}C$ .**

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 6)	TYP	MAX (Note 6)	UNIT
$V_{OS}$	Input Offset Voltage				150	$\mu V$
		-40 $^{\circ}C$ to +85 $^{\circ}C$			250	$\mu V$
		-40 $^{\circ}C$ to +125 $^{\circ}C$			<b>350</b>	$\mu V$
$TCV_{OS}$	Input Offset Voltage Temperature Coefficient	-40 $^{\circ}C$ to +125 $^{\circ}C$		<b>0.5</b>	<b>1.4</b>	$\mu V/^{\circ}C$
$\Delta V_{OS}/\text{Time}$	Long Term $V_{OS}$ Stability			0.4		$\mu V/\text{mo}$
$I_B$	Input Bias Current			0.2	1	nA
		-40 $^{\circ}C$ to +125 $^{\circ}C$			<b>1</b>	nA
$I_{OS}$	Input Offset Current			0.2	1	nA
		-40 $^{\circ}C$ to +125 $^{\circ}C$			<b>1</b>	nA
$e_N$	Input Noise Voltage	$f = 0.1\text{Hz}$ to 10Hz		0.38		$\mu V_{P-P}$
	Input Noise Voltage Density	$f = 10\text{Hz}$		13		$nV/\sqrt{\text{Hz}}$
	Input Noise Voltage Density	$f = 100\text{Hz}$		9.6		$nV/\sqrt{\text{Hz}}$
	Input Noise Voltage Density	$f = 1\text{kHz}$		9.5		$nV/\sqrt{\text{Hz}}$
$i_N$	Input Noise Current Density	$f = 1\text{kHz}$		87		$fA/\sqrt{\text{Hz}}$
$V_{CMIR}$	Common Mode Input Voltage Range	Guaranteed by CMRR test	$V_- + 2$		$V_+ - 2$	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = V_- + 2V$ to $V_+ - 2V$	120	140		dB
			<b>120</b>			dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 3V$ to $\pm 20V$	115	130		dB
			<b>115</b>			dB
$V_{OL}$	Output Voltage Low, $V_{OUT}$ to $V_-$	$R_L = 2k\Omega$		1.2	1.25	V
		$R_L = 2k\Omega$ , -40 $^{\circ}C$ to +125 $^{\circ}C$			<b>1.3</b>	V
$V_{OH}$	Output Voltage High, $V_+$ to $V_{OUT}$	$R_L = 2k\Omega$		1.2	1.25	V
		$R_L = 2k\Omega$ , -40 $^{\circ}C$ to +125 $^{\circ}C$			<b>1.3</b>	V
SR	Slew Rate	$R_L = 2k\Omega$ , $C_L = 100\text{pF}$		0.2		V/ $\mu s$
GBWP	Gain Bandwidth Product	$R_L = 100k\Omega$ , $C_L = 60\text{pF}$		600		kHz
AVOL	Large Signal Gain	$V_{OUT} = \pm 3V$ to $\pm 13V$ , $R_L = 10k\Omega$	120	140		dB
			<b>120</b>			dB

**Electrical Specifications**  $V_S = \pm 5V$  to  $\pm 15V$ ,  $R_L = \text{Open}$ ,  $V_{CM} = 0V$ ,  $T_A = +25^\circ\text{C}$ , unless otherwise specified. **Boldface limits apply over the operating temperature range,  $-40^\circ\text{C}$  to  $+125^\circ\text{C}$ .** (Continued)

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 6)	TYP	MAX (Note 6)	UNIT
$I_S$	Supply Current			1.18	1.4	mA
					<b>1.7</b>	mA
$V_S$	Supply Voltage		$\pm 3V$		$\pm 20V$	V
$I_{SC}$	Short Circuit Current			30		mA

NOTE:

- 6. Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or design.

**Typical Performance Curves**  $V_S = \pm 15V$ ,  $V_{CM} = 0V$ ,  $R_L = \text{Open}$ , unless otherwise specified.

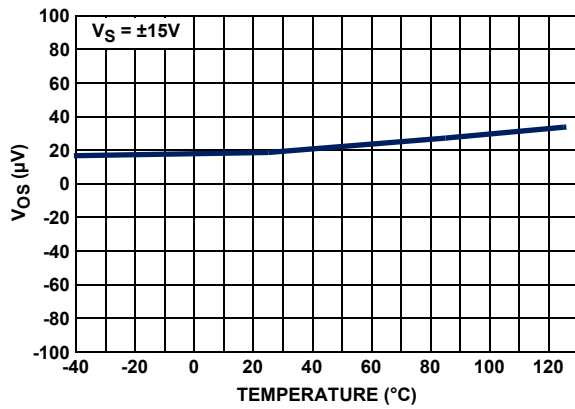


FIGURE 3. INPUT OFFSET VOLTAGE ( $V_{OS}$ ) vs TEMPERATURE

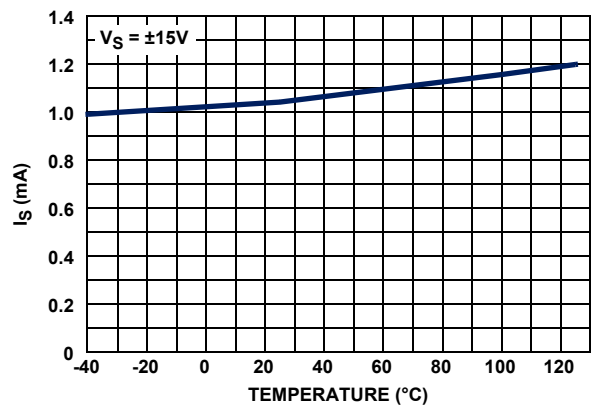


FIGURE 4. POWER SUPPLY CURRENT ( $I_S$ ) vs TEMPERATURE

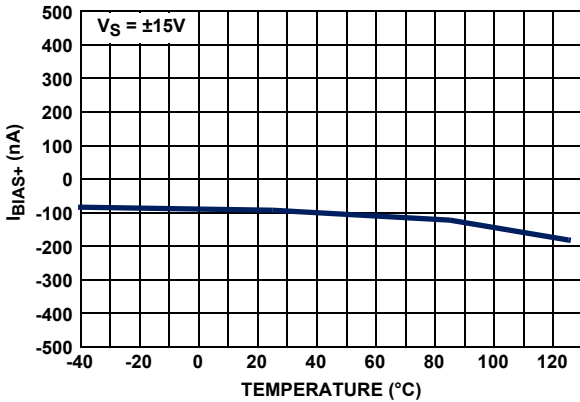


FIGURE 5. POSITIVE INPUT BIAS CURRENT ( $I_{BIAS+}$ ) vs TEMPERATURE

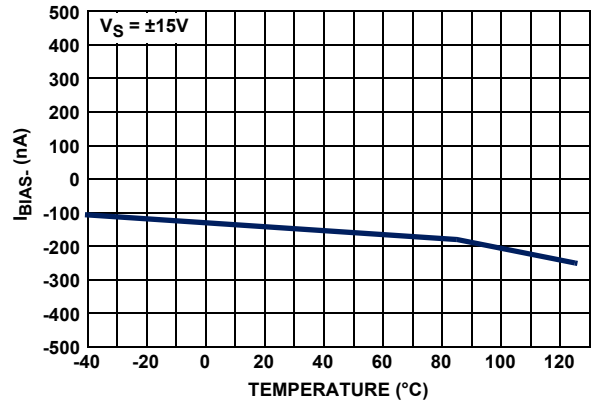


FIGURE 6. NEGATIVE INPUT BIAS CURRENT ( $I_{BIAS-}$ ) vs TEMPERATURE

**Typical Performance Curves**  $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}$ , unless otherwise specified. (Continued)

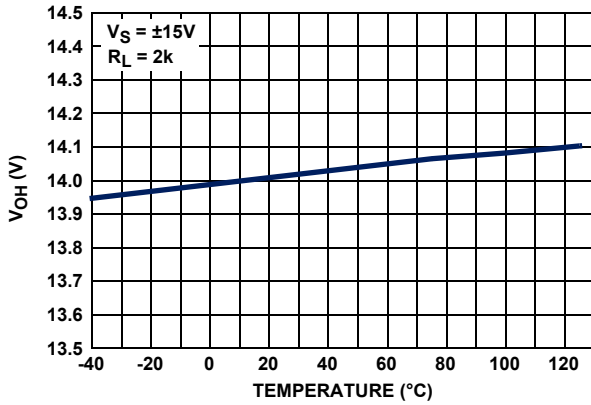


FIGURE 7. POSITIVE OUTPUT VOLTAGE ( $V_{OH}$ ) vs TEMPERATURE

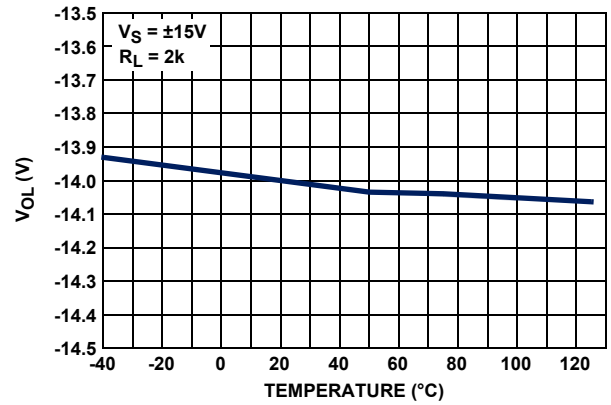


FIGURE 8. POSITIVE OUTPUT VOLTAGE ( $V_{OL}$ ) vs TEMPERATURE

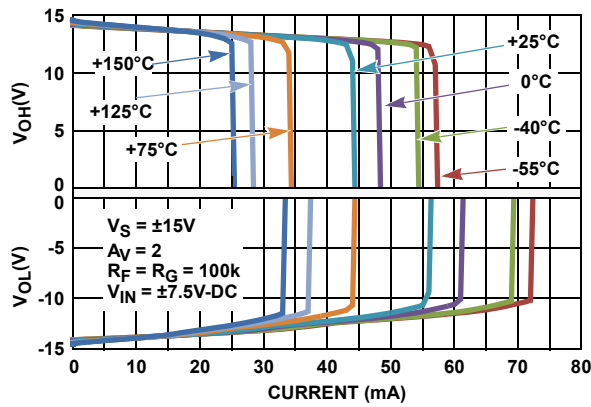


FIGURE 9. POSITIVE OUTPUT VOLTAGE ( $V_{OH}$ ) vs OUTPUT CURRENT ( $I_{OUT}$ ) vs TEMPERATURE

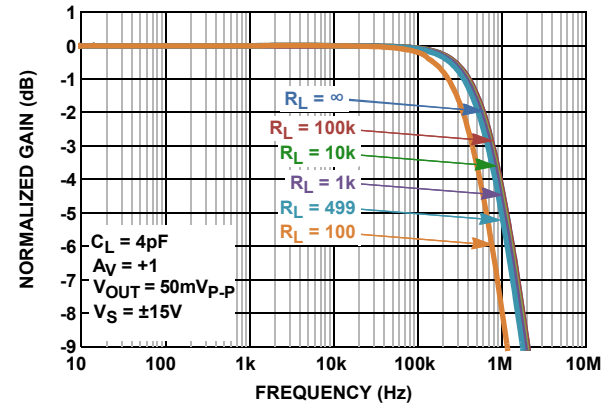


FIGURE 10. UNITY GAIN FREQUENCY RESPONSE vs  $R_L$

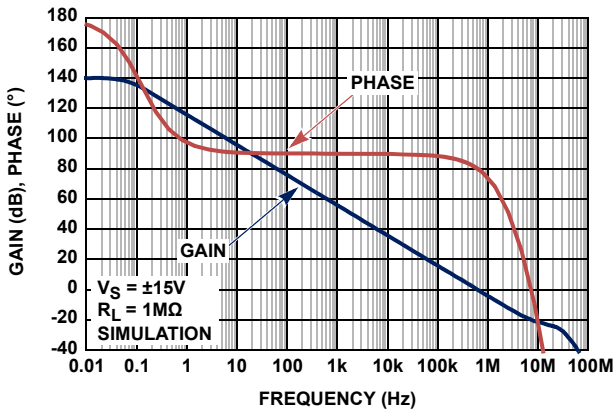


FIGURE 11. OPEN LOOP GAIN-PHASE vs FREQUENCY

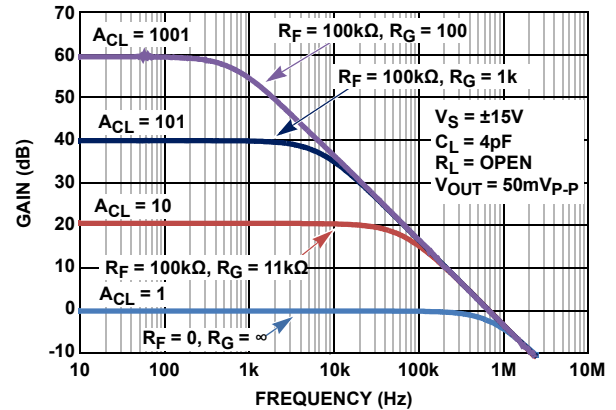


FIGURE 12. FREQUENCY RESPONSE vs CLOSED LOOP GAIN

**Typical Performance Curves**  $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}$ , unless otherwise specified. (Continued)

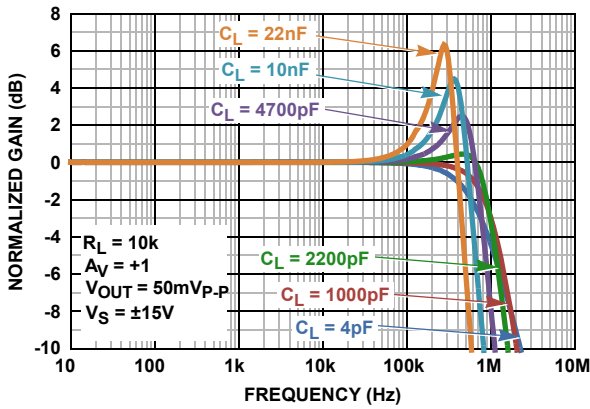


FIGURE 13. UNITY GAIN FREQUENCY RESPONSE vs  $C_L$

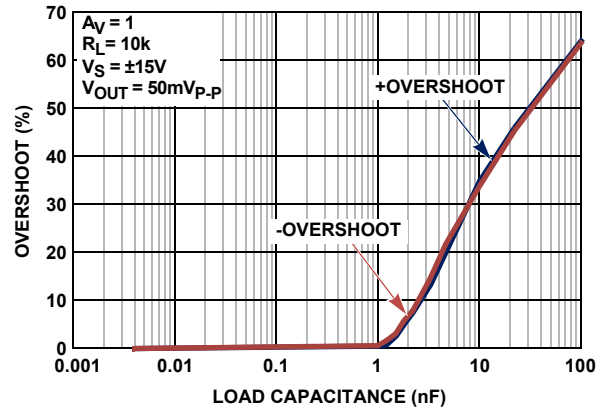


FIGURE 14. OVERSHOOT vs LOAD CAPACITANCE

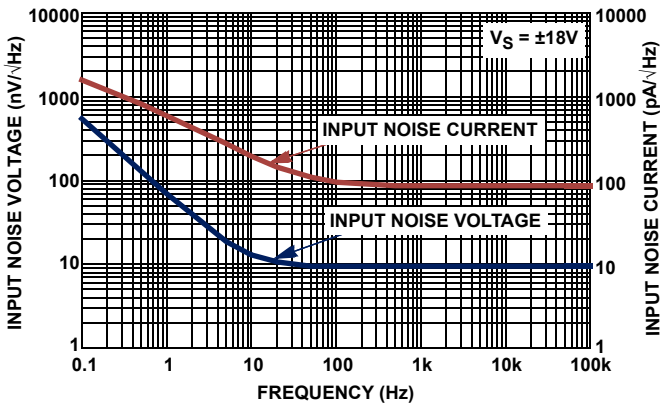


FIGURE 15. INPUT NOISE VOLTAGE AND CURRENT SPECTRAL DENSITY

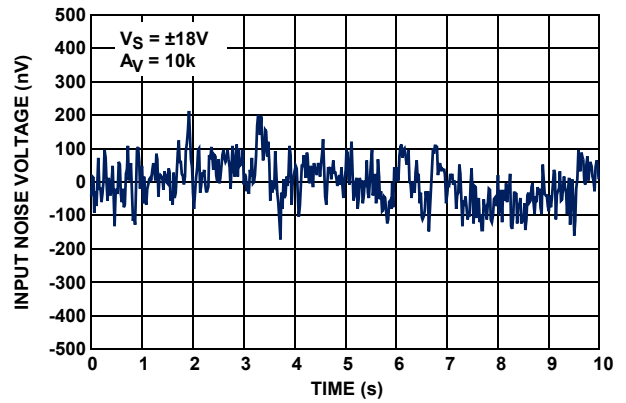


FIGURE 16. INPUT NOISE VOLTAGE 0.1Hz to 10Hz

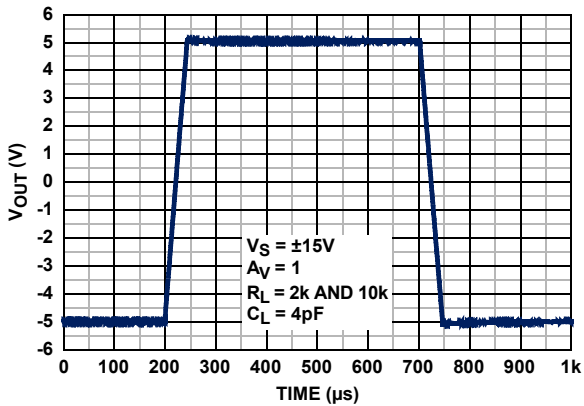


FIGURE 17. LARGE SIGNAL TRANSIENT RESPONSE

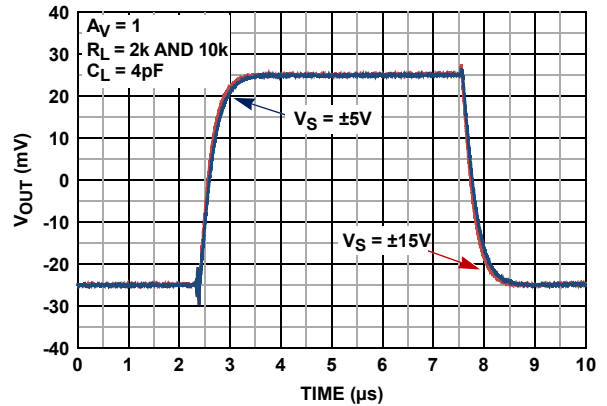


FIGURE 18. SMALL SIGNAL TRANSIENT RESPONSE

## Typical Performance Curves $V_S = \pm 15V$ , $V_{CM} = 0V$ , $R_L = \text{Open}$ , unless otherwise specified. (Continued)

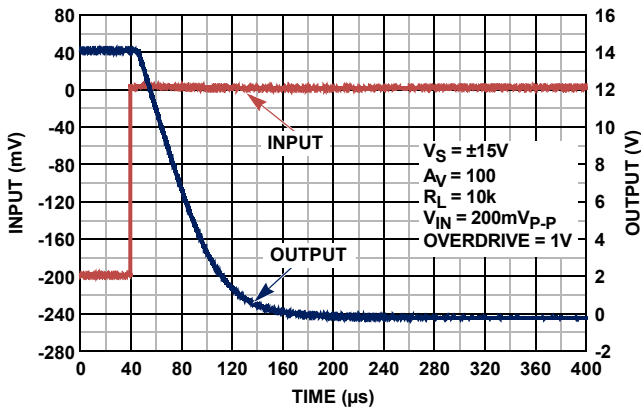


FIGURE 19. POSITIVE OUTPUT OVERLOAD RESPONSE TIME

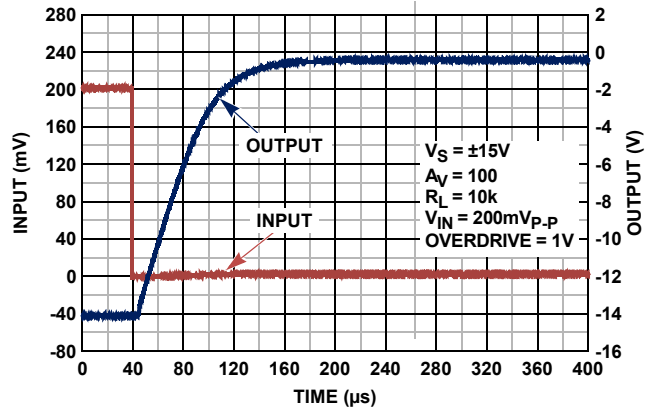


FIGURE 20. NEGATIVE OUTPUT OVERLOAD RESPONSE TIME

## Applications Information

### Functional Description

The ISL28177 is a low noise op amp fabricated in a 40V complementary bipolar DI process designed for general purpose low power applications. It utilizes a super-beta NPN input stage with input bias current cancellation for low input bias current and low input noise voltage. A complimentary bipolar output stage enables high capacitive load drive without external compensation.

### Operating Voltage Range

The ISL28177 is designed to operate over the 6V ( $\pm 3V$ ) to 40V ( $\pm 20V$ ) range. The common mode input voltage range extends to 2V from each rail, and the output voltage swings to 1.3V of each rail.

### Input Performance

The super-beta NPN input pair reduces input bias current while maintaining good frequency response, low input bias current and low noise. Input bias cancellation circuits provide additional bias current reduction to  $<1nA$ , and excellent temperature stabilization and low  $TCV_{OS}$ .

### Input ESD Diode Protection

The input terminals (IN+ and IN-) have internal ESD protection diodes to the positive and negative supply rails, series connected 500 $\Omega$  current limiting resistors and an anti-parallel diode pair across the inputs (Figure 21).

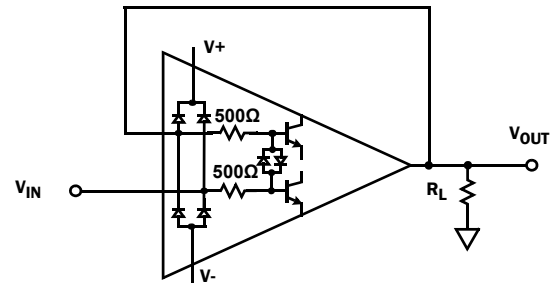


FIGURE 21. INPUT ESD DIODE CURRENT LIMITING

The series resistors limit the high feed-through currents that can occur in pulse applications when the input  $dv/dt$  exceeds the 0.2V/ $\mu s$  slew rate of the amplifier. Without the series resistors, the input can forward-bias the anti-parallel diodes causing current to flow to the output, resulting in severe distortion and possible diode failure. Figure 17 provides an example of distortion free large signal response using a 10V $_{p-p}$  input pulse with an input rise time of  $<1ns$ . The series resistors enable the input differential voltage to be equal to the maximum power supply voltage (40V) without damage.

In applications where one or both amplifier input terminals are at risk of exposure to high voltages beyond the power supply rails, current limiting resistors may be needed at the input terminal to limit the current through the power supply ESD diodes to 20mA max.

### Output Current Limiting

The output current is internally limited to approximately  $\pm 30mA$  at  $+25^\circ C$  and can withstand a short circuit to either rail as long as the power dissipation limits are not exceeded. Continuous operation under these conditions may degrade long term reliability.

### Output Phase Reversal

Output phase reversal is a change of polarity in the amplifier transfer function when the input voltage exceeds the supply voltage. The ISL28177 is immune to output phase reversal.

## Power Dissipation

It is possible to exceed the +150°C maximum junction temperature under certain load and power supply conditions. It is therefore important to calculate the maximum junction temperature ( $T_{JMAX}$ ) for all applications to determine if power supply voltages, load conditions, or package type need to be modified to remain in the safe operating area. These parameters are related using Equation 1:

$$T_{JMAX} = T_{MAX} + \theta_{JA} \times PD_{MAXTOTAL} \quad (EQ. 1)$$

where:

- $PD_{MAXTOTAL}$  is the sum of the maximum power dissipation of each amplifier in the package ( $PD_{MAX}$ )
- $PD_{MAX}$  for each amplifier can be calculated using Equation 2:

$$PD_{MAX} = V_S \times I_{qMAX} + (V_S - V_{OUTMAX}) \times \frac{V_{OUTMAX}}{R_L} \quad (EQ. 2)$$

where:

- $T_{MAX}$  = Maximum ambient temperature
- $\theta_{JA}$  = Thermal resistance of the package
- $PD_{MAX}$  = Maximum power dissipation of 1 amplifier
- $V_S$  = Total supply voltage
- $I_{qMAX}$  = Maximum quiescent supply current of 1 amplifier
- $V_{OUTMAX}$  = Maximum output voltage swing of the application

## ISL28177 SPICE Model

Figure 22 shows the SPICE model schematic and Figure 23 shows the net list for the SPICE model. The model is a simplified version of the actual device and simulates important AC and DC parameters. AC parameters incorporated into the model are: 1/f and flatband noise voltage, Slew Rate, CMRR, Gain and Phase. The DC parameters are, VOS,  $I_{OS}$ , total supply current and output voltage swing. The model uses typical parameters given in the “Electrical Specifications” table beginning on page 3. The AVOL is adjusted for 140dB with the dominant pole at 0.075Hz. The CMRR is set 145dB,  $f_{cm} = 500kHz$ . The input stage models the actual device to present an accurate AC representation. The model is configured for ambient temperature of +25°C.

Figures 24 through 37 show the characterization vs simulation results for the Noise Voltage, Closed Loop Gain vs Frequency, Small Signal 0.1V Step, Large Signal 5V Step Response, Open Loop Gain Phase, CMRR, Unity Gain Frequency Response vs  $C_L$  and Output Voltage Swing for  $\pm 15V$  supplies.

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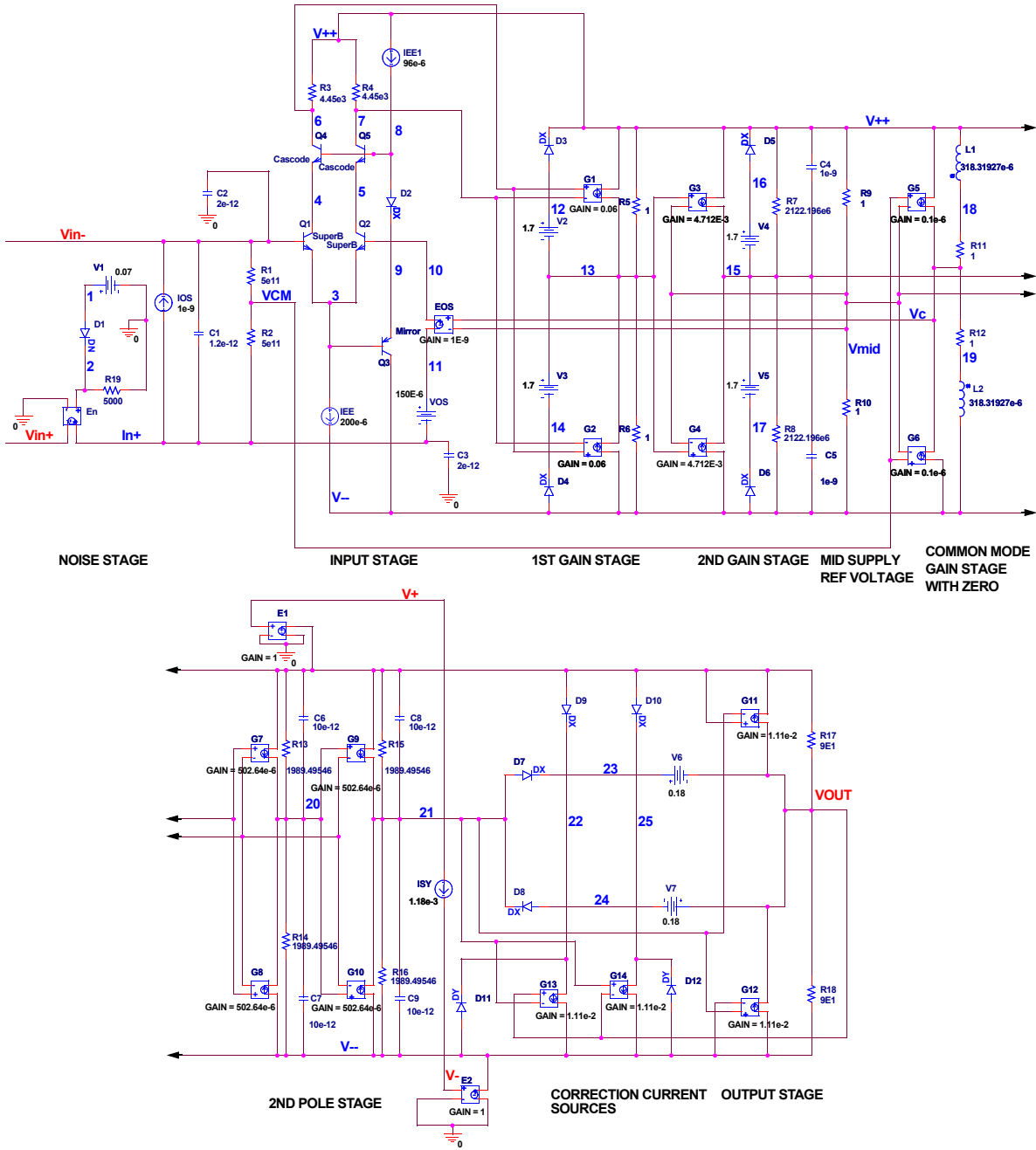


FIGURE 22. SPICE MODEL SCHEMATIC

\*ISL28177 Macromodel  
 \*\*Revision History:  
 \*Revision A, LaFontaine December 14, 2011  
 \*Model for Noise, quiescent supply currents,  
 \*CMRR 145dB, fcm=500kHz, AVOL 140dB  
 \*f=0.075Hz SR = 0.2V/us, GBWP 600kHz,  
 \*2nd pole 8Mhz, output voltage clamp  
 \*and short ckt current limit.  
 \*  
 \*Copyright 2011 by Intersil Corporation  
 \*Refer to data sheet "LICENSE  
 \*STATEMENT", Use of this model indicates  
 \*your acceptance with the terms and  
 \*provisions in the License Statement.  
 \*  
 \*Intended use:  
 \*This Pspice Macromodel is intended to give  
 \*typical DC and AC performance  
 \*characteristics under a wide range of  
 \*external circuit configurations using  
 \*compatible simulation platforms - such as  
 \*iSim PE.  
 \*  
 \*Device performance features supported by  
 \*this model  
 \*Typical, room temp., nominal power supply  
 \*voltages used to produce the following  
 \*characteristics:  
 \*Open and closed loop I/O impedances  
 \*Open loop gain and phase  
 \*Closed loop bandwidth and frequency  
 \*response  
 \*Loading effects on closed loop frequency  
 \*response  
 \*Input noise terms including 1/f effects  
 \*Slew rate  
 \*Input and Output Headroom limits to I/O  
 \*voltage swing  
 \*Supply current at nominal specified supply  
 \*voltages  
 \*\*  
 \*Device performance features NOT  
 \*supported by this model:  
 \*Harmonic distortion effects  
 \*Disable operation (if any)  
 \*Thermal effects and/or over temperature  
 \*parameter variation  
 \*Limited performance variation vs. supply  
 \*voltage is modeled  
 \*Part to part performance variation due to  
 \*normal process parameter spread  
 \*Any performance difference arising from  
 \*different packaging  
 \* source  
 :  
 \*  

```

    +input  |  -input
    |       |  |
    |       |  | +Vsupply
    |       |  | |
    |       |  | | -Vsupply
    |       |  | | |
    |       |  | | | output
    |       |  | | | |
    +-----+--+--+--+--+
    
```

 .subckt ISL28177 Vin+ Vin- V+ V- VOUT  
 \* source ISL28177\_SPICEMODEL  
 \*  
 \*Voltage Noise  
 E\_En IN+ VIN+ 2 0 1  
 D\_D1 1 2 DN  
 V\_V1 1 0 0.07  
 R\_R19 2 0 5000  
 \*

```

*Input Stage
I_IOS IN+ VIN- DC 1e-9
C_C1 IN+ VIN- 1.2e-12
C_C2 0 VIN- 2e-12
C_C3 0 IN+ 2e-12
R_R1 VCM VIN- 5e11
R_R2 IN+ VCM 5e11
R_R3 6 V++ 4.45e3
R_R4 7 V++ 4.45e3
Q_Q1 4 VIN- 3 SuperB
Q_Q2 5 10 3 SuperB
Q_Q3 V-- 3 9 Mirror
Q_Q4 6 8 4 Cascode
Q_Q5 7 8 5 Cascode
I_IEE 3 V-- DC 200e-6
I_IEE1 V++ 8 DC 96e-6
D_D2 8 9 DX
E_EOS 10 11 VC VMID 1E-9
V_VOS 11 IN+ 30E-6
*
*1st Gain Stage
G_G1 V++ 13 6 7 0.06
G_G2 V-- 13 6 7 0.06
R_R5 13 V++ 1
R_R6 V-- 13 1
V_V2 12 13 1.7
V_V3 13 14 1.7
D_D3 12 V++ DX
D_D4 V-- 14 DX
*
*2nd Gain Stage
G_G3 V++ 15 13 VMID 4.712E-3
G_G4 V-- 15 13 VMID 4.712E-3
R_R7 15 V++ 2122.196e6
R_R8 V-- 15 2122.196e6
V_V4 16 15 1.7
V_V5 15 17 1.7
D_D5 16 V++ DX
D_D6 V-- 17 DX
C_C4 15 V++ 1e-9
C_C5 V-- 15 1e-9
*
*Mid supply Ref
R_R9 VMID V++ 1
R_R10 V-- VMID 1
E_E1 V++ 0 V+ 0 1
E_E2 V-- 0 V- 0 1
I_ISY V+ V- DC 1.18e-3
*
*Common Mode Gain Stage with Zero
G_G5 V++ VC VCM VMID 0.1e-6
G_G6 V-- VC VCM VMID 0.1e-6
R_R11 VC 18 1
R_R12 19 VC 1
L_L1 18 V++ 318.31927e-6
L_L2 19 V-- 318.31927e-6
*
*2nd Pole Stage
G_G7 V++ 20 15 VMID 502.64e-6
G_G8 V-- 20 15 VMID 502.64e-6
G_G9 V++ 21 20 VMID 502.64e-6
G_G10 V-- 21 20 VMID 502.64e-6
R_R13 20 V++ 1989.49546
R_R14 V-- 20 1989.49546
R_R15 21 V++ 1989.49546
R_R16 V-- 21 1989.49546
C_C6 20 V++ 10e-12
C_C7 V-- 20 10e-12
C_C8 21 V++ 10e-12
C_C9 V-- 21 10e-12

```

```

*
*Output Stage with Correction Current Sources
G_G11 VOUT V++ V++ 21 1.11e-2
G_G12 V-- VOUT 21 V-- 1.11e-2
G_G13 22 V-- VOUT 21 1.11e-2
G_G14 25 V-- 21 VOUT 1.11e-2
D_D7 21 23 DX
D_D8 24 21 DX
D_D9 V++ 22 DX
D_D10 V++ 25 DX
D_D11 V-- 22 DY
D_D12 V-- 25 DY
V_V6 23 VOUT 0.18
V_V7 VOUT 24 0.18
R_R17 VOUT V++ 9E1
R_R18 V-- VOUT 9E1
*
.model SuperB npn
+ is=184E-15 bf=30e3 va=15 ik=70E-3 rb=50
+ re=0.065 rc=35 cje=1.5E-12 cjc=2E-12
+ kf=0 af=0
.model Cascode npn
+ is=502E-18 bf=150 va=300 ik=17E-3
+rb=140 re=0.011 rc=900 cje=0.2E-12
+cjc=0.16E-12f kf=0 af=0
.model Mirror prnp
+ is=4E-15 bf=150 va=50 ik=138E-3 rb=185
+ re=0.101 rc=180 cje=1.34E-12
+ cjc=0.44E-12
+ kf=0 af=0
.model DN D(KF=6.69e-9 AF=1)
.MODEL DX D(IS=1E-12 Rs=0.1)
.MODEL DY D(IS=1E-15 BV=50 Rs=1)
.ends subckt ISL28177

```

FIGURE 23. SPICE NET LIST

## Characterization vs Simulation Results

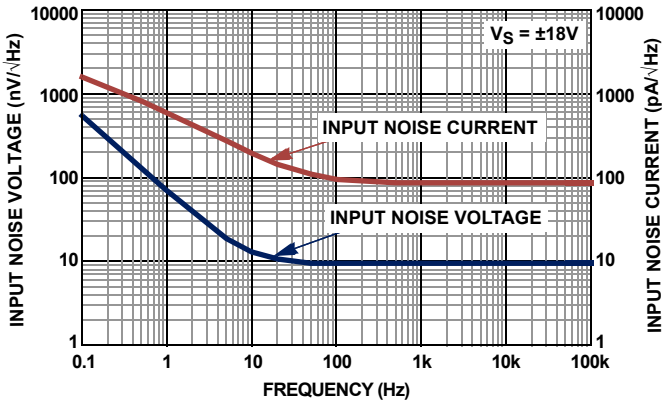


FIGURE 24. CHARACTERIZED INPUT NOISE VOLTAGE

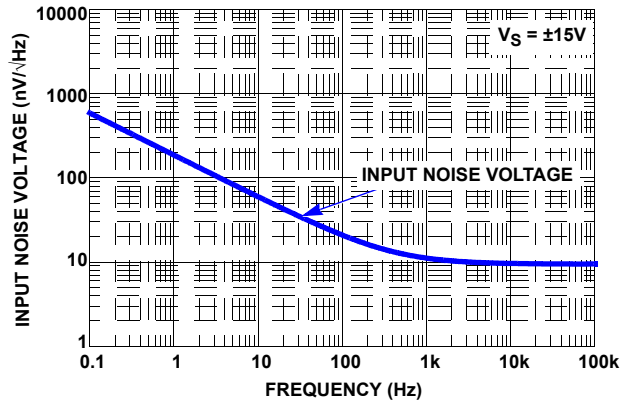


FIGURE 25. SIMULATED INPUT NOISE VOLTAGE

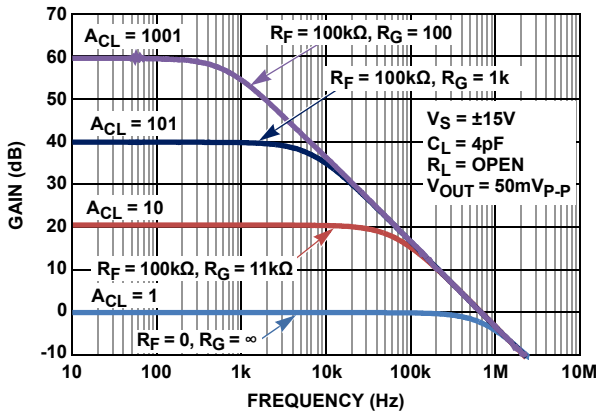


FIGURE 26. CHARACTERIZED CLOSED LOOP GAIN vs FREQUENCY

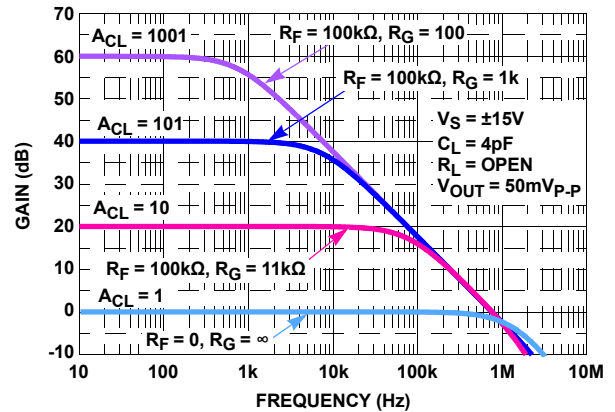


FIGURE 27. SIMULATED CLOSED LOOP GAIN vs FREQUENCY

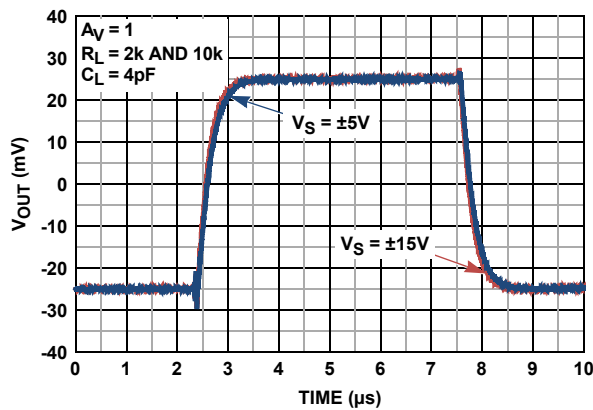


FIGURE 28. CHARACTERIZED SMALL SIGNAL TRANSIENT RESPONSE vs  $R_L$ ,  $V_S = \pm 0.9V, \pm 2.5V$

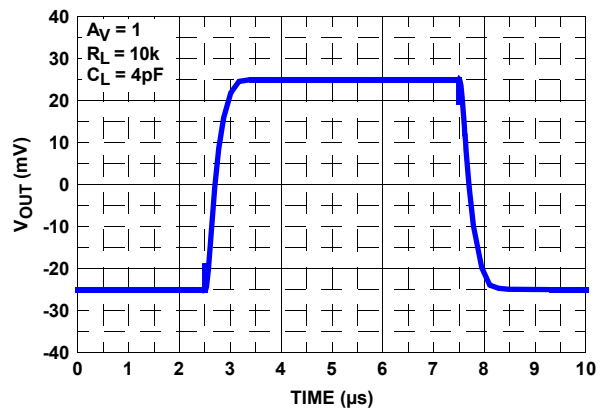


FIGURE 29. SIMULATED SMALL SIGNAL TRANSIENT RESPONSE  $V_S = \pm 15V$

## Characterization vs Simulation Results (Continued)

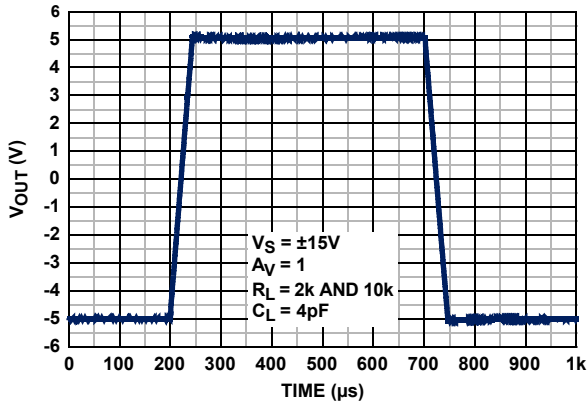


FIGURE 30. CHARACTERIZED LARGE SIGNAL TRANSIENT RESPONSE vs  $R_L$ ,  $V_S = \pm 15V$

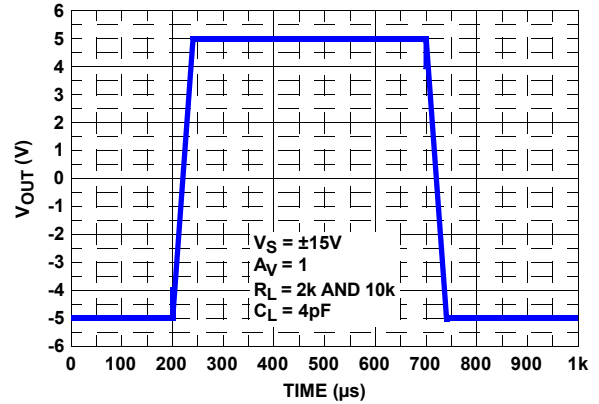


FIGURE 31. SIMULATED LARGE SIGNAL TRANSIENT RESPONSE,  $V_S = \pm 14V$

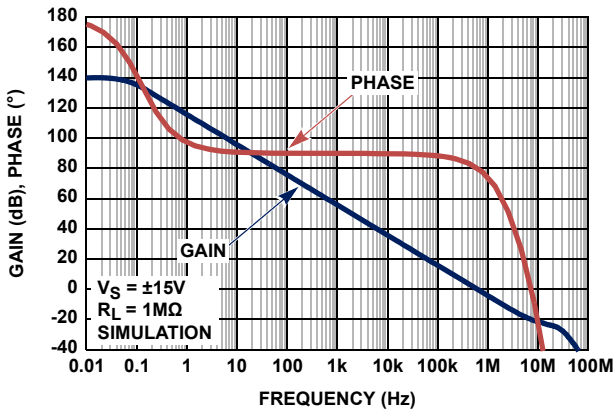


FIGURE 32. SIMULATED (DESIGN) OPEN-LOOP GAIN, PHASE vs FREQUENCY

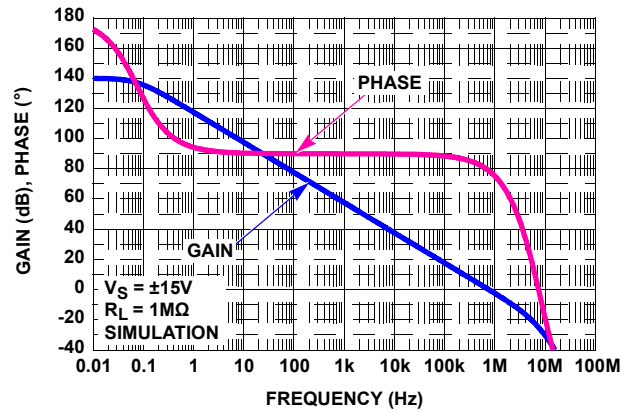


FIGURE 33. SIMULATED (SPICE) OPEN-LOOP GAIN, PHASE vs FREQUENCY

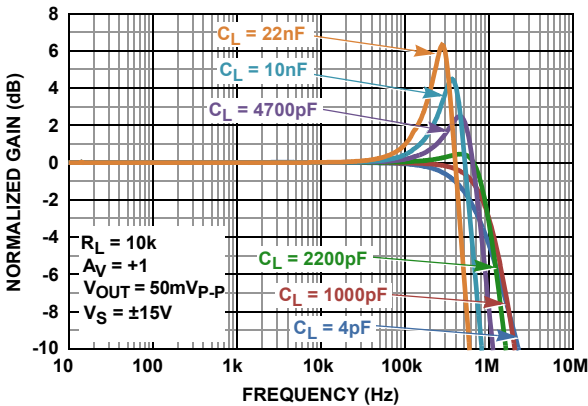


FIGURE 34. CHARACTERIZED UNITY GAIN FREQUENCY RESPONSE vs  $C_L$

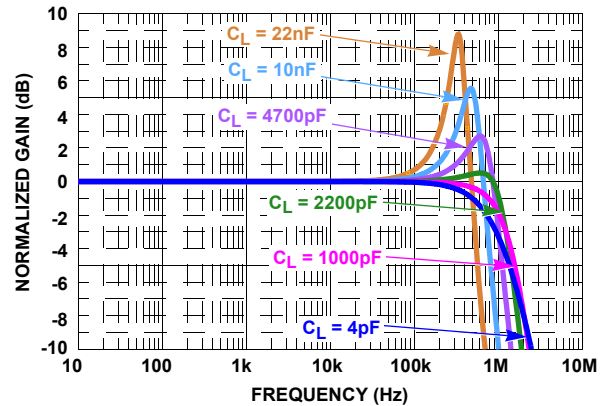


FIGURE 35. SIMULATED UNITY GAIN FREQUENCY RESPONSE vs  $C_L$

## Characterization vs Simulation Results (Continued)

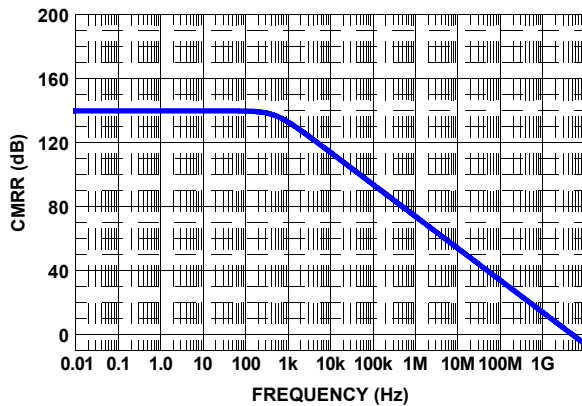
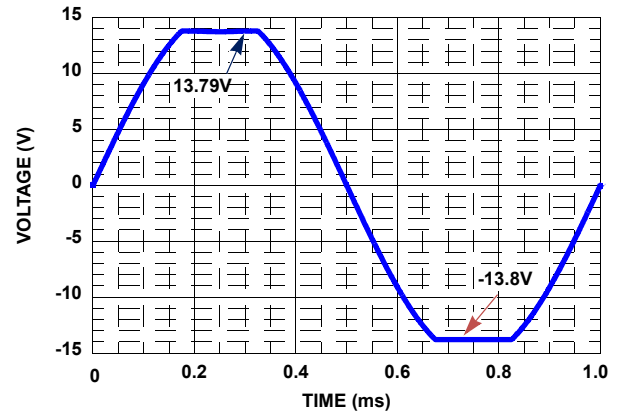


FIGURE 36. SIMULATED (SPICE) CMRR

FIGURE 37. SIMULATED OUTPUT VOLTAGE SWING  $\pm 15V$ 

## Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest Rev.

DATE	REVISION	CHANGE
March 29, 2012	FN7859.2	Changed Note 1 in "Ordering Information" on page 2 from: "Add "-T*" suffix for tape and reel. Please refer to TB347 for details on reel specifications." to: "Please refer to TB347 for details on reel specifications." Listed out tape and reel parts individually in "Ordering Information" on page 2 (ISL28177FBZ-T13, ISL28177FBZ-T7, ISL28177FBZ-T7A)
January 5, 2012	FN7859.1	Added SPICE model to data sheet. Added ESD Ratings to description on page 1.
October 31, 2011	FN7859.0	Initial Release

## Products

Intersil Corporation is a leader in the design and manufacture of high-performance analog semiconductors. The Company's products address some of the industry's fastest growing markets, such as, flat panel displays, cell phones, handheld products, and notebooks. Intersil's product families address power management and analog signal processing functions. Go to [www.intersil.com/products](http://www.intersil.com/products) for a complete list of Intersil product families.

For a complete listing of Applications, Related Documentation and Related Parts, please see the respective device information page on [intersil.com](http://intersil.com): [ISL28177](http://intersil.com/ISL28177)

To report errors or suggestions for this datasheet, please go to: [www.intersil.com/askourstaff](http://www.intersil.com/askourstaff)

FITs are available from our website at: <http://rel.intersil.com/reports/sear>

# Package Outline Drawing (M8.15E)

## M8.15E

8 LEAD NARROW BODY SMALL OUTLINE PLASTIC PACKAGE

Rev 0, 08/09



**NOTES:**

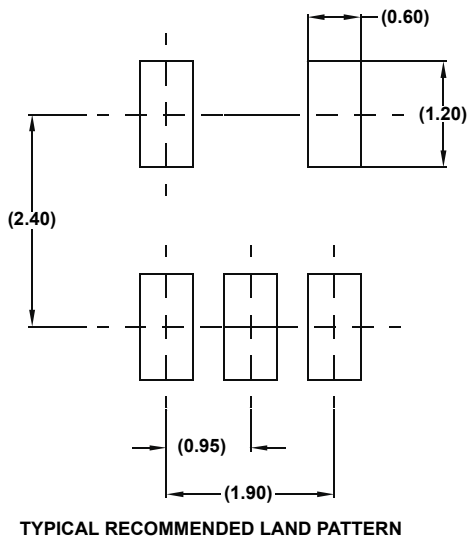
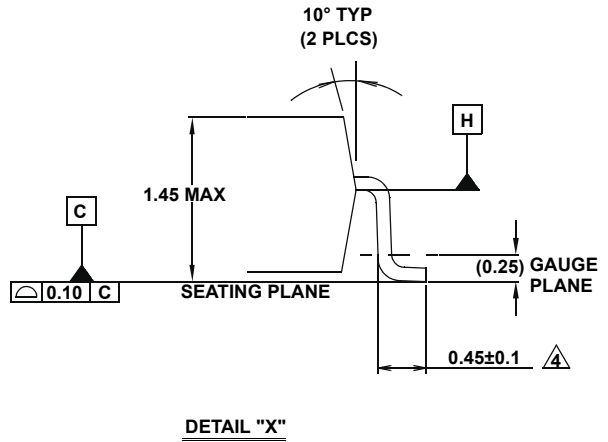
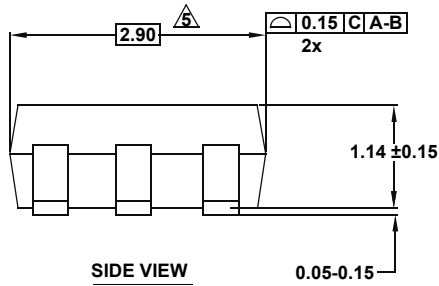
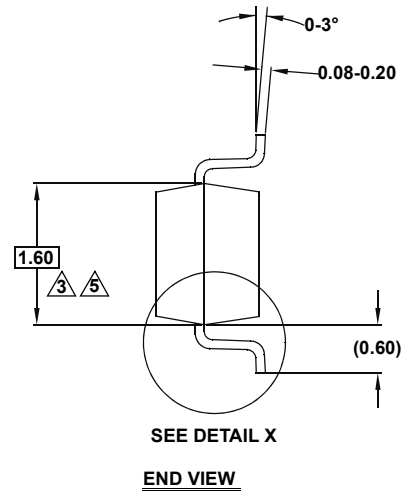
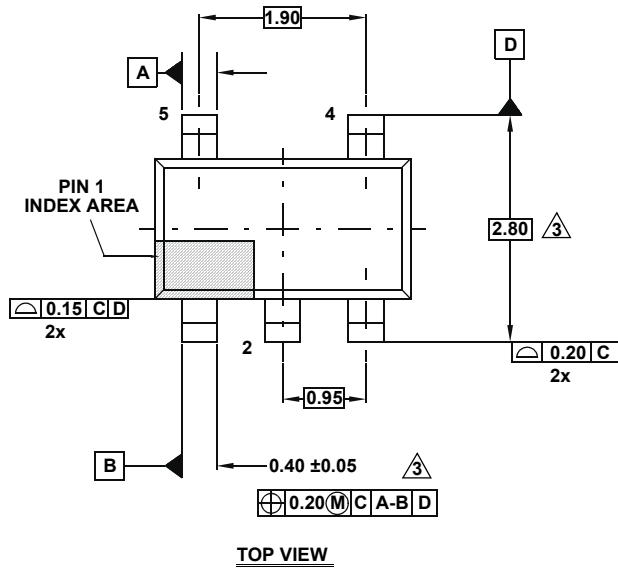
1. Dimensions are in millimeters.  
Dimensions in ( ) for Reference Only.
2. Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
3. Unless otherwise specified, tolerance : Decimal ± 0.05
4. Dimension does not include interlead flash or protrusions.  
Interlead flash or protrusions shall not exceed 0.25mm per side.
5. The pin #1 identifier may be either a mold or mark feature.
6. Reference to JEDEC MS-012.

# Package Outline Drawing

## P5.064A

### 5 LEAD SMALL OUTLINE TRANSISTOR PLASTIC PACKAGE

Rev 0, 2/10



**NOTES:**

1. Dimensions are in millimeters.  
Dimensions in ( ) for Reference Only.
2. Dimensioning and tolerancing conform to ASME Y14.5M-1994.
3. Dimension is exclusive of mold flash, protrusions or gate burrs.
4. Foot length is measured at reference to gauge plane.
5. This dimension is measured at Datum "H".
6. Package conforms to JEDEC MO-178AA.