# Push-Pull FET Driver with Integrated Oscillator and Clock Output 

## General Description

The MAX5075 is a +4.5 V to +15 V push-pull, current-fed topology driver subsystem with an integrated oscillator for use in telecom module power supplies. The device drives two MOSFETs connected to a center-tapped transformer primary providing secondary-side, isolated, negative or positive voltages. This device features a programmable, accurate, integrated oscillator with a synchronizing clock output that synchronizes an external PWM regulator. A single external resistor programs the internal oscillator frequency from 50 kHz to 1.5 MHz .
The MAX5075 incorporates a dual MOSFET driver with $\pm 3 \mathrm{~A}$ peak drive currents and $50 \%$ duty cycle. The MOSFET driver generates complementary signals to drive external ground-referenced n-channel MOSFETs.
The MAX5075 is available with a clock output frequency to MOSFET driver frequency ratio of $1 x, 2 x$, and $4 x$. The MAX5075 is available in a thermally enhanced 8-pin $\mu \mathrm{MAX}{ }^{\circledR}$ package and is specified over the $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ operating temperature range.

## Applications

Current-Fed, High-Efficiency Power-Supply Modules Power-Supply Building Subsystems Push-Pull Driver Subsystems
$\mu M A X$ is a registered trademark of Maxim Integrated Products, Inc. Pin Configuration appears at end of data sheet.

Features

- Current-Fed, Push-Pull Driver Subsystem
- Programmable, Accurate Internal Oscillator
- Single +4.5V to +15V Supply Voltage Range
- Dual $\pm 3 \mathrm{~A}$ Gate-Drive Outputs
- 1mA Operating Current at 250kHz with No Capacitive Load
- Synchronizing Clock Frequency Generation Options
- Thermally Enhanced 8-Pin $\mu$ MAX Package
$-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ Operating Temperature Range
Ordering Information

| PART | PIN- <br> PACKAGE | TOP <br> MARK | PKG <br> CODE | feLk/fndrv <br> RATIO |
| :---: | :--- | :---: | :---: | :---: |
| MAX5075AAUA | $8 \mu$ MAX-EP* | AAAU | U8E-2 | 1 |
| MAX5075BAUA | $8 \mu$ MAX-EP* | AAAV | U8E-2 | 2 |
| MAX5075CAUA | $8 \mu$ MAX-EP* | AAAW | U8E-2 | 4 |

*EP = Exposed paddle.
Note: All devices specified for $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ operating temperature range.

Typical Operating Circuit


# Push-Pull FET Driver with Integrated Oscillator and Clock Output 

## ABSOLUTE MAXIMUM RATINGS

$V_{C c}$ to DGND, PGND
-0.3 V to +18 V
CLK, RT to DGND. $\qquad$ .-0.3V to +6 V
NDRV1, NDRV2 to PGND $\qquad$ -0.3 V to ( $\left.\mathrm{V}_{\mathrm{CC}}+0.3 \mathrm{~V}\right)$
DGND to PGND $\qquad$ -0.3 V to +0.3 V
CLK Current $\pm 20 \mathrm{~mA}$
NDRV1, NDRV2 Peak Current (200ns).
$\pm 5$ A
NDRV1, NDRV2 Reverse Current (Latchup Current)...... $\pm 500 \mathrm{~mA}$

Continuous Power Dissipation ( $\mathrm{T}_{\mathrm{A}}=+70^{\circ} \mathrm{C}$ ) 8-Pin $\mu$ MAX (derate $10.3 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) ........... 825 mW Operating Temperature Range ......................... $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ Maximum Junction Temperature ..................................... $+150^{\circ} \mathrm{C}$ Storage Temperature Range ............................. $60^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ Lead Temperature (soldering, 10s) ................................. $+300^{\circ} \mathrm{C}$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

$\left(\mathrm{V}_{C C}=+12 \mathrm{~V}, \mathrm{R}_{\mathrm{RT}}=124 \mathrm{k} \Omega\right.$, NDRV1 $=$ NDRV2 $=$ open, $\mathrm{T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{J}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$, unless otherwise noted. Typical values are measured at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.) (Note 1)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SUPPLY |  |  |  |  |  |  |  |
| Input Voltage Supply Range | VCC |  |  | 4.5 |  | 15.0 | V |
| Switching Supply Current | Iccsw | $\mathrm{fOSC}=250 \mathrm{kHz}$ |  |  | 1 | 3 | mA |
| Undervoltage Lockout | VUVLO | $\mathrm{V}_{\text {CC }}$ rising |  | 3 | 3.5 | 4 | V |
| UVLO Hysteresis |  |  |  |  | 300 |  | mV |
| OSCILLATOR |  |  |  |  |  |  |  |
| Frequency Range | fosc | (Note 2) |  | 50 |  | 1500 | kHz |
| Accuracy |  | $\mathrm{fOSC}=250 \mathrm{kHz}, 6 \mathrm{~V} \leq \mathrm{VCC} \leq 15 \mathrm{~V}$ (Note 3) |  | -8 |  | +10 | \% |
| Oscillator Jitter |  |  |  |  | $\pm 0.6$ |  | \% |
| CLK Output High Voltage |  | $I C L K=1 \mathrm{~mA}$ | $7 \mathrm{~V} \leq \mathrm{V}_{\mathrm{cc}} \leq 15 \mathrm{~V}$ | 3.9 |  | 5.0 | V |
|  |  |  | $4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}} \leq 7 \mathrm{~V}$ | 3.35 |  | 5.0 |  |
| CLK Output Low Voltage |  | ICLK $=-1 \mathrm{~mA}$ |  |  |  | 50 | mV |
| CLK Output Rise Time |  | CCLK $=30 \mathrm{pF}$ |  |  | 35 |  | ns |
| CLK Output Fall Time |  | CCLK $=30 \mathrm{pF}$ |  |  | 10 |  | ns |
| GATE DRIVERS (NDRV1, NDRV2) |  |  |  |  |  |  |  |
| Output High Voltage | VOH | $\mathrm{I}_{\text {NDRV1 }}=\mathrm{I}_{\text {NDRV2 }}=100 \mathrm{~mA}$ |  | $\begin{gathered} \hline \mathrm{V}_{\mathrm{CC}}- \\ 0.3 \end{gathered}$ |  |  | V |
| Output Low Voltage | VOL | $\mathrm{I}_{\text {NDRV1 }}=l_{\text {NDRV2 }}=-100 \mathrm{~mA}$ |  |  |  | 0.3 | V |
| Output Peak Current | IP | Sourcing and sinking |  |  | 3 |  | A |
| Driver Output Impedance |  | NDRV_ sourcing 100mA |  |  | 1.8 | 3 | $\Omega$ |
|  |  | NDRV_ sinking 100mA |  |  | 1.6 | 2.6 |  |
| Latchup Current Protection |  | Reverse current at NDRV1/NDRV2 |  |  | 400 |  | mA |
| Rise Time | tR | CLOAD $=2 \mathrm{nF}$ |  |  | 10 |  | ns |
| Fall Time | $\mathrm{t}_{\mathrm{F}}$ | CLOAD $=2 \mathrm{nF}$ |  |  | 10 |  | ns |

Note 1: The MAX5075 is $100 \%$ tested at $T_{A}=T_{J}=+125^{\circ} \mathrm{C}$. All limits over temperature are guaranteed by design.
Note 2: Use the following formula to calculate the MAX5075 oscillator frequency: fOSC $=10^{12} /\left(32 \times R_{R T}\right)$.
Note 3: The accuracy of the oscillator's frequency is lower at frequencies greater than 1 MHz .

# Push-Pull FET Driver with Integrated Oscillator and Clock Output 

Typical Operating Characteristics
$\left(\mathrm{V}_{\mathrm{CC}}=+12 \mathrm{~V}, \mathrm{R}_{\mathrm{RT}}=124 \mathrm{k} \Omega\right.$, NDRV $_{-}=$open, $\mathrm{CLK}=$ open.$)$


## Push-Pull FET Driver with Integrated Oscillator and Clock Output

$\left(\mathrm{V}_{C C}=+12 \mathrm{~V}, \mathrm{R}_{\mathrm{RT}}=124 \mathrm{k} \Omega, \mathrm{NDRV}_{-}=\right.$open, CLK $=$open. $)$




# Push-Pull FET Driver with Integrated Oscillator and Clock Output 

Pin Description

| PIN | NAME | FUNCTION |
| :---: | :---: | :--- |
| 1 | CLK | Synchronizing Clock Output. Clock output with a $\pm 10 \mathrm{~mA}$ peak current drive that can be used to <br> synchronize an external PWM regulator. CLK/NDRV1 frequency has a 1x, 2x, or 4x ratio. See the <br> Synchronizing Clock Output section. |
| 2 | I.C. | Internal Connection. Connect to ground. Internal function. |
| 3 | RT | Oscillator Timing Resistor Connection. Bypass RT with a series combination of a 4.7k $\Omega$ resistor and a <br> 1nF capacitor to DGND. Connect a resistor from RT to DGND to set the internal oscillator. |
| 4 | DGND | Digital Ground. Connect DGND to ground plane. |
| 5 | PGND | Power Ground. Connect PGND to ground plane. |
| 6 | NDRV1 | Gate Driver 1. Connect NDRV1 to the gate of the external n-channel FET. |
| 7 | NDRV2 | Gate Driver 2. Connect NDRV2 to the gate of the external n-channel FET. |
| 8 | VCC | Power-Supply Input. Bypass VCC to PGND with 0.14FIl1 1 F ceramic capacitors. |
| EP | EP | Exposed Pad. Internally connected to DGND. Connect exposed pad to ground plane. |



Figure 1. MAX5075 Functional Diagram

# Push-Pull FET Driver with Integrated Oscillator and Clock Output 

## Detailed Description

The MAX5075 is a +4.5 V to +15 V push-pull, current-fed topology driver subsystem with an integrated oscillator for use in 48 V module power supplies.
The MAX5075 features a programmable, accurate integrated oscillator with a synchronizing clock output that can be used to synchronize an external PWM stage. A single external resistor programs the internal oscillator frequency from 50 kHz to 1.5 MHz .
The MAX5075 incorporates a dual MOSFET driver with $\pm 3 \mathrm{~A}$ peak drive currents and a $50 \%$ duty cycle. The MOSFET driver generates complementary signals to drive external ground-referenced n -channel MOSFETs.
The MAX5075 is available with a clock output frequency to MOSFET driver frequency ratios of $1 \mathrm{x}, 2 \mathrm{x}$, and 4 x .

Internal Oscillator
An external resistor at RT programs the MAX5075 internal oscillator frequency from 50 kHz to 1.5 MHz . The MAX5075A/B NDRV1 and NDRV2 switching frequencies are one-half the programmed oscillator frequency with a nominal $50 \%$ duty cycle. The MAX5075C NDRV1 and NDRV2 switching frequencies are one-fourth the oscillator frequency.
Use the following formula to calculate the internal oscillator frequency:

$$
\mathrm{f}_{\mathrm{OSC}}=\frac{10^{12}}{32 \times \mathrm{R}_{\mathrm{RT}}}
$$

where fosc is the oscillator frequency and RRT is a resistor connected from RT to DGND in ohms.
Place a series combination of a $4.7 \mathrm{k} \Omega$ resistor and a 1 nF capacitor from RT to DGND for stability and to filter out noise.

## Synchronizing Clock Output

The MAX5075 provides a buffered clock output that can be used to synchronize the oscillator input of a PWM controller. CLK is powered from an internal 5 V regulator and sources/sinks up to 10 mA . The MAX5075 has internal CLK output frequency to NDRV1 and NDRV2 switching frequency ratios set to $1 \mathrm{x}, 2 \mathrm{x}$, or 4 x (Table 1).
The MAX5075A has a CLK frequency to NDRV_ frequency ratio set to 1x. The MAX5075B has a CLK frequency to NDRV_ frequency ratio set to $2 x$ and the MAX5075C has a CLK frequency to NDRV_ frequency ratio set to 4 x . There is a typical 30 ns delay from CLK to NDRV_ output.

Table 1. MAX5075 CLK Output Frequency

| PART | fcLK | fNDRV1 | fcLk to fsw <br> RATIO |
| :--- | :---: | :---: | :---: |
| MAX5075A | fosc $/ 2$ | fosc $/ 2$ | 1 |
| MAX5075B | fosC | fosc $/ 2$ | 2 |
| MAX5075C | fosc | fosc $/ 4$ | 4 |



Figure 2. MAX5075 CLK Timing Diagrams

## Applications Information

Supply Bypassing
Pay careful attention to bypassing and grounding the MAX5075. Peak supply and output currents may exceed 3A when driving large MOSFETs. Ground shifts due to insufficient device grounding may also disturb other circuits sharing the same ground-return path. Any series inductance in the Vcc, NDRV1, NDRV2, and/or GND paths can cause noise due to the very high di/dt when switching the MAX5075 with any capacitive load. Place one or more $0.1 \mu \mathrm{~F}$ ceramic capacitors in parallel as close to the device as possible to bypass Vcc to PGND. Use a ground plane to minimize ground-return resistance and inductance. Place the external MOSFETs as close as possible to the MAX5075 to further minimize board inductance and AC path impedance.

# Push-Pull FET Driver with Integrated Oscillator and Clock Output 

Power Dissipation
The power dissipation of the MAX5075 is a function of the sum of the quiescent current and the output current (either capacitive or resistive load). Maintain the sum of the currents so the maximum power dissipation limit is not exceeded. The power dissipation (PDISS) due to the quiescent switching supply current (ICCSW) can be calculated as:

$$
\text { PDISS }=\mathrm{V}_{C C} \times \mathrm{Iccsw}
$$

For capacitive loads, use the following equation to estimate the power dissipation:

$$
\text { PLOAD }=2 \times \text { CLOAD } \times \mathrm{V}_{\mathrm{CC}}{ }^{2} \times \mathrm{f}_{\mathrm{NDRV}}
$$

where CLOAD is the capacitive load at NDRV1 and NDRV2, $\mathrm{V}_{\mathrm{CC}}$ is the supply voltage, and $\mathrm{f}_{\mathrm{NDRV}}$ _ is the MAX5075 NDRV_ switching frequency.
Calculate the total power dissipation ( $\mathrm{PT}_{\mathrm{T}}$ ) as follows:
PT = PDISS + PLOAD

## Layout Recommendations

The MAX5075 sources and sinks large currents that can create very fast rise and fall edges at the gate of the switching MOSFETs. The high di/dt can cause unacceptable ringing if the trace lengths and impedances are not well controlled. Use the following PC board layout guidelines when designing with the MAX5075:

- Place one or more $0.1 \mu \mathrm{~F}$ decoupling ceramic capacitors from Vcc to PGND as close to the device as possible. Connect $\mathrm{V}_{C C}$ and all ground pins to large copper areas. Place one bulk capacitor of $10 \mu \mathrm{~F}$ on the PC board with a low-impedance path to the VCC input and PGND of the MAX5075.
- Two AC current loops form between the device and the gate of the driven MOSFETs. The MOSFETs look like a large capacitance from gate to source when the gate pulls low. The current loop is from the MOSFET gate to NDRV1 and NDRV2 of the MAX5075, to PGND, and to the source of the MOSFET. When the gate of the MOSFET pulls high, the current is from the Vcc terminal of the decoupling capacitor, to $\mathrm{V}_{\mathrm{CC}}$ of the MAX5075, to NDRV1 and NDRV2, and to the MOSFET gate and source. Both charging current and discharging current loops are important. Minimize the physical distance and the impedance in these AC current paths.
- Keep the device as close to the MOSFET as possible.


## Chip Information

TRANSISTOR COUNT: 1335
PROCESS: BiCMOS


## Push-Pull FET Driver with Integrated Oscillator and Clock Output

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to www.maxim-ic.com/packages.)


Revision History
Pages changed at Rev 1: 1, 2, 5, 6, 8

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