



# SY58621L

## Precision 3.2Gbps CML/LVPECL Backplane Transceiver with Integrated Loopback

### General Description

The SY58621L is a low jitter, high-speed transceiver with a variable swing LVPECL transmitter buffer and a CML high-gain receiver optimized for precision telecom and enterprise server transmission line and backplane data management. The SY58621L distributes data to 3.2Gbps guaranteed over temperature and voltage.

The SY58621L transmitter differential input includes Micrel's unique, patented 3-pin input termination architecture that directly interfaces to any (AC- or DC-coupled) differential signal as small as 100mV (200mV<sub>PP</sub>) without any termination resistor network in the signal path. The receiver differential input is optimized to interface directly to AC-coupled signals as small as 10mV (20mV<sub>PP</sub>). The receiver output is 50 $\Omega$  source-terminated CML and the transmitter output is variable swing 80mV to 800mV LVPECL with extremely fast rise/fall time.

To support remote self-testing, the SY58621L features a high-speed loopback test mode. The input control signal LOOPBACK enables an internal loopback path from the transmitter input to the receiver output.

The SY58621L operates from a 3.3V  $\pm$ 10% supply and is guaranteed over the full industrial temperature range of  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ . The SY58621L is part of Micrel's high-speed, Precision Edge<sup>®</sup> product line. For applications that requires a CML receiver and transmitter, consider the SY58620L.

All support documentation can be found on Micrel's web site at: [www.micrel.com](http://www.micrel.com).

### Applications

- Backplane management
- Active cable transceivers
- SONET/SDH data/clock applications
- 4X Fibre Channel applications



Precision Edge<sup>®</sup>

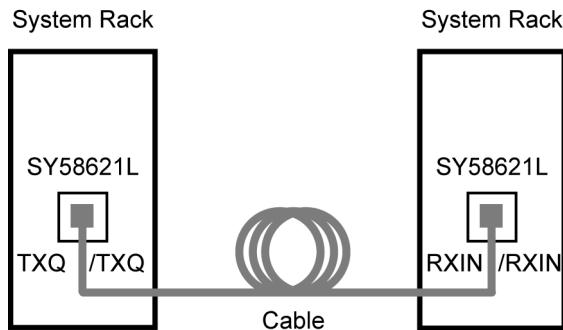
### Features

- Guaranteed AC performance over temperature and voltage:
  - Maximum Throughput 3.2Gbps
  - $<160\text{ps}$  t<sub>r</sub>/t<sub>f</sub> time
- Transmitter
  - Patented input termination directly interfaces to AC- or DC-coupled differential inputs
  - Variable swing LVPECL output
- Receiver
  - 32dB high-gain Input
  - Internal 50 $\Omega$  input termination
  - Accepts AC-coupled input signals as small as 10mV (20mV<sub>PP</sub>)
  - 400mV (800mV<sub>PP</sub>) differential CML output swing
- Loss-of-Signal (LOS)
  - High-gain, TTL-compatible LOS output with internal 4.75k $\Omega$  pull-up
  - Programmable LOS level set
- Ultra-low jitter design
  - $<5\text{ps}_{\text{RMS}}$  random jitter
- Patent-pending MUX isolates the receiver and the transmitter channels minimizing on crosstalk
- Selectable loopback diagnostic mode
- Output enables on transmitter and receiver outputs
- Power supply +3.3V  $\pm$ 10%
- Industrial temperature range  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$
- Available in 24-pin (4mm x 4mm) QFN

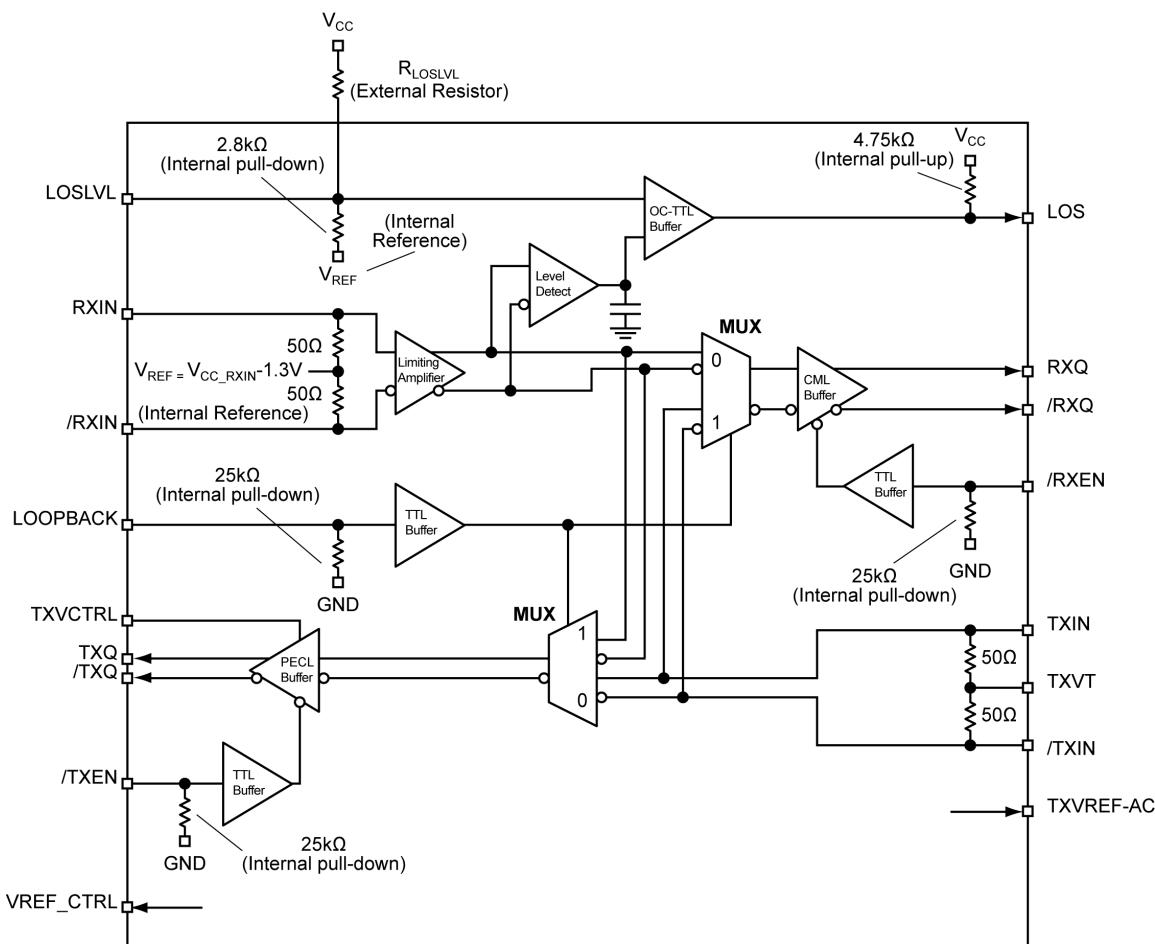
### Markets

- Precision telecom
- Enterprise server
- ATE
- Test and measurement

## Typical Applications



## Functional Block Diagram



**Note:**

It is recommended that  $R_{LOSLVL} \leq 10k\Omega$ . See the "Typical Operating Characteristics" section for more details.

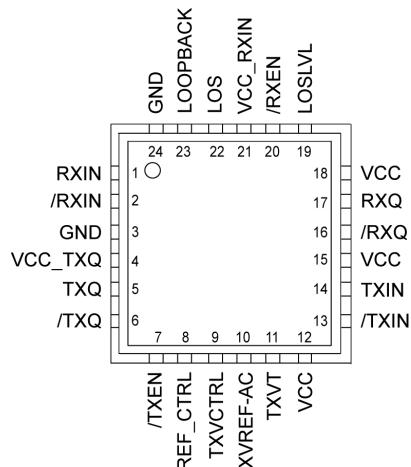
## Ordering Information<sup>(1)</sup>

Part Number	Package Type	Operating Range	Package Marking	Lead Finish
SY58621LMG	QFN-24	Industrial	621L with Pb-Free bar-line indicator	NiPdAu Pb-Free
SY58621LMGTR <sup>(2)</sup>	QFN-24	Industrial	621L with Pb-Free bar-line indicator	NiPdAu Pb-Free

## Notes:

1. Contact factory for die availability. Dice are guaranteed at  $T_A = 25^\circ\text{C}$ , DC Electricals only.
2. Tape and Reel.

## Pin Configuration



## Pin Description

### Inputs

Pin Number	Pin Name	Pin Description
23	LOOPBACK	LOOPBACK Mode Control. TTL/CMOS control input. LOOPBACK is an active HIGH signal used to control the LOOPBACK MUX. LOOPBACK is internally connected to a $25\text{k}\Omega$ pull-down resistor and will default to a LOW state if left open. $V_{TH} = V_{CC}/2$ .
20	/RXEN	Receiver Output Control. TTL/CMOS control input. /RXEN is an active LOW signal used to enable the receiver outputs. /RXEN is internally connected to a $25\text{k}\Omega$ pull-down resistor and will default to a LOW state if left open. $V_{TH} = V_{CC}/2$ .
1, 2	RXIN, /RXIN	Receiver Differential Input. Input accepts AC differential signals as small as 10mV (20mV <sub>PP</sub> ). Each pin internally terminates to $V_{CC\_RXIN}-1.3\text{V}$ (internal voltage reference) through $50\Omega$ . Input will default to an indeterminate state if left open. See figure 6b.
7	/TXEN	Transmitter Output Control. TTL/CMOS control input. /TXEN is an active LOW signal used to enable the transmitter output. /TXEN is internally connected to a $25\text{k}\Omega$ pull-down resistor and will default to a LOW state if left open. $V_{TH} = V_{CC}/2$ .
14, 13	TXIN, /TXIN	Transmitter Differential Input. Input accepts AC- or DC-coupled differential signals as small as 100mV (200mV <sub>PP</sub> ). Each pin terminates to the TXVT pin through $50\Omega$ . Note that this input will default to an indeterminate state if left open. See figure 6a.
9	TXVCTRL	Transmitter Output Swing Control. Input that controls the output amplitude of the transmitter. The operating range of the control input is from $V_{REF\_CTRL}$ (max swing) to $V_{CC}$ (min swing). Control of the output swing can be obtained by using a variable resistor between $V_{REF\_CTRL}$ and $V_{CC\_TXQ}$ through a wiper driving TXVCTRL. Setting TXVCTRL to $V_{CC\_TXQ}$ sets the output swing to min swing. Refer to the "Interface Applications" and "Output Stage" sections for more details.
11	TXVT	Input Termination Center-Tap. Each side of the transmitter differential input pair terminates to the TXVT pin. The TXVT pin provides a center-tap to a termination network for maximum interface flexibility. Refer to the "Input Stage" section for more details.

## Outputs

Pin Number	Pin Name	Pin Description
22	LOS	Loss-of-Signal Output. TTL-compatible output with internal $4.75\text{k}\Omega$ pull-up resistor. Loss-of-Signal asserts to logic HIGH when the receiver input amplitudes fall below the threshold set by LOSLVL.
19	LOSLVL	RX Loss-of-Signal Level Set. A resistor ( $R_{LOSLVL}$ ) connected between LOSLVL and $V_{CC}$ sets the threshold for the data input amplitude at which the LOS output is asserted. Default is max sensitivity. LOSLVL is used to set the Loss-of-Signal (LOS) voltage. It is internally connected to a $2.8\text{k}\Omega$ pull-down resistor to an internal $V_{REF}$ voltage source. See "Typical Operating Characteristics," and "Application Implementation" sections for more details.
17, 16	RXQ, /RXQ	Receiver Differential Output. Output is CML compatible. Refer to the "Truth Table" and "Output Stage" sections for more details. Unused output pair may be left open. The output is designed to drive 400mV (800mV <sub>PP</sub> ) into $50\Omega$ to $V_{CC}$ or $100\Omega$ across the pair.
5, 6	TXQ, /TXQ	Transmitter differential Variable Swing Output. Output is LVPECL-compatible. Please refer to the "Truth Table" section for details. Unused output pair may be left open. Each output is designed to drive 80mV (min) to 800mV (typ) into $50\Omega$ to $V_{CC}$ -2V depending on TXVCTRL.
8	VREF_CTRL	Transmitter Output Reference Voltage. Output biases to $V_{CC\_TXQ}$ -1.3V. Connecting V <sub>REF_CTRL</sub> to TXVCTRL sets the transmitter output swing to max swing.
10	TXVREF-AC	Transmitter Input Reference Voltage. This output biases to $V_{CC}$ -1.3V. It is used when AC coupling the transmitter input. For AC-coupled applications, connect TXVREF-AC to the TXVT pin and bypass with a $0.01\mu\text{F}$ low ESR capacitors to $V_{CC}$ . See "Input Stage" section for more details. Maximum sink/source current is $\pm 1.5\text{mA}$ .

## Power Pins

Pin Number	Pin Name	Pin Description
3, 24	GND, Exposed Pad	Ground. GND pins and exposed pad must be connected to the same ground plane.
12, 15, 18	VCC	$3.3\text{V} \pm 10\%$ Positive Power Supply. Bypass with $0.1\mu\text{F}/0.01\mu\text{F}$ low ESR capacitors and place as close to each $V_{CC}$ pins as possible. Power pins are not connected internally and must be connected to the same power supply externally.
21	VCC_RXIN	$3.3\text{V} \pm 10\%$ Receive Input Power Supply. Bypass with $0.1\mu\text{F}/0.01\mu\text{F}$ low ESR capacitors and place as close to the $V_{CC\_RXIN}$ pin as possible. Power pins are not connected internally and must be connected to the same power supply externally.
4	VCC_TXQ	$3.3\text{V} \pm 10\%$ Output Transmit Power Supply. Bypass with $0.1\mu\text{F}/0.01\mu\text{F}$ low ESR capacitors and place as close to the $V_{CC\_TXQ}$ pin as possible. Power pins are not connected internally and must be connected to the same power supply externally.

## Truth Table

LOOPBACK	RXQ	TXQ
0	RXIN	TXIN
1	TXIN	RXIN

## Absolute Maximum Ratings<sup>(1)</sup>

### Supply Voltage

( $V_{CC}$ ,  $V_{CC\_TXQ}$ ,  $V_{CC\_RXIN}$ ) ..... -0.5V to +4.0V

### Input Voltage

LOSLVL .....  $V_{REF}$  -1.2V to  $V_{CC}$

LOOPBACK ..... -0.5V to  $V_{CC}$

/TXEN, /RXEN ..... -0.5V to  $V_{CC}$

TXVCTRL ..... VREF\_CTRL -1.2V to  $V_{CC}$

TXIN, /TXIN ..... -0.5V to  $V_{CC}$

### LVPECL Output Current ( $I_{OUT}$ )

TXQ, /TXQ

Continuous .....  $\pm 50$ mA

Surge .....  $\pm 100$ mA

### Source or Sink Current on

TXVT .....  $\pm 100$ mA

LOS .....  $\pm 5$ mA

RXQ, /RXQ .....  $\pm 25$ mA

RXIN, /RXIN .....  $\pm 10$ mA

TXIN, /TXIN .....  $\pm 50$ mA

TXVREF-AC, VREF-CTRL .....  $\pm 2$ mA

Lead Temperature (soldering, 20sec.) ..... 260°C

Storage Temperature ( $T_s$ ) ..... -65°C to +150°C

## Operating Ratings<sup>(2)</sup>

### Supply Voltage

( $V_{CC}$ ,  $V_{CC\_TXQ}$ ,  $V_{CC\_RXIN}$ ) ..... +3.0V to +3.6V

Ambient Temperature ( $T_A$ ) ..... -40°C to +85°C

### Package Thermal Resistance<sup>(3)</sup>

QFN ( $\theta_{JA}$ ) ..... 50°C/W

QFN ( $\psi_{JB}$ ) ..... Junction-to-Board ..... 30°C/W

## DC Electrical Characteristics<sup>(4)</sup>

$T_A$  = -40°C to +85°C, unless otherwise stated.

Symbol	Parameter	Condition	Min	Typ	Max	Units
$V_{CC}$	Power Supply		3	3.3	3.6	V
$V_{CC\_TXQ}$	Transmit Power Supply		3	3.3	3.6	V
$V_{CC\_RXIN}$	Receive Power Supply		3	3.3	3.6	V
$I_{CC}$	Power Supply Current	No load, max. $V_{CC}$		100	150	mA

## Receiver Input DC Electrical Characteristics

$V_{CC\_RXIN}$  = 3.3V  $\pm 10\%$ ;  $T_A$  = -40°C to +85°C, unless otherwise stated.

Symbol	Parameter	Condition	Min	Typ	Max	Units
$R_{IN}$	Input Resistance (RXIN to VREF)		45	50	55	$\Omega$
$R_{DIFF\_IN}$	Input Resistance (RXIN to /RXIN)		90	100	110	$\Omega$
$V_{IN}$	Input Voltage Swing (RXIN, /RXIN)	See Figure 5a AC-coupled	10		900	mV
$V_{DIFF\_IN}$	Differential Input Voltage Swing  RXIN - /RXIN	See Figure 5b AC-coupled	20		1800	mV
$V_{REF}$	Internal Reference Voltage		$V_{CC\_RXIN}$ -1.48	$V_{CC\_RXIN}$ -1.32	$V_{CC\_RXIN}$ -1.16	V

### Notes:

1. Permanent device damage may occur if absolute maximum ratings are exceeded. This is a stress rating only and functional operation is not implied at conditions other than those detailed in the operational sections of this data sheet. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
2. The data sheet limits are not guaranteed if the device is operated beyond the operating ratings.
3. Package thermal resistance assumes exposed pad is soldered (or equivalent) to the devices most negative potential on the PCB.  $\theta_{JA}$  and  $\psi_{JB}$  values are determined for a 4-layer board in still-air, unless otherwise stated.
4. The circuit is designed to meet the DC specifications shown in the above table after thermal equilibrium has been established.

## Receiver Output DC Electrical Characteristics

$V_{CC} = 3.3V \pm 10\%$ ,  $R_L = 100\Omega$  across the outputs;  $T_A = -40^\circ C$  to  $+85^\circ C$ , unless otherwise stated.

Symbol	Parameter	Condition	Min	Typ	Max	Units
$V_{OH}$	Output HIGH Voltage (RXQ, /RXQ)	$R_L = 50\Omega$ to $V_{CC}$	$V_{CC} - 0.020$	$V_{CC} - 0.010$	$V_{CC}$	V
$V_{OUT}$	Output Voltage Swing (RXQ, /RXQ)	See Figure 5a	325	400	500	mV
$V_{DIFF\_OUT}$	Differential Output Voltage Swing (RXQ, /RXQ)	See Figure 5b	650	800	1000	mV
$R_{OUT}$	Single-Ended Output Impedance		45	50	55	$\Omega$
$R_{DIFF\_OUT}$	Differential Output Impedance		90	100	110	$\Omega$
$V_{OFFSET}$	Differential Output Offset	$R_L = 50\Omega$ to $V_{CC}$ , limiting mode	-140		+140	mV

## Transmitter Input DC Electrical Characteristics

$V_{CC} = 3.3V \pm 10\%$ ;  $T_A = -40^\circ C$  to  $+85^\circ C$ , unless otherwise stated.

Symbol	Parameter	Condition	Min	Typ	Max	Units
$R_{IN}$	Input Resistance (TXIN to TXVT)		45	50	55	$\Omega$
$R_{DIFF\_IN}$	Differential Input Resistance (TXIN to /TXIN)		90	100	110	$\Omega$
$V_{IH}$	Input HIGH Voltage (TXIN, /TXIN)		1.2		$V_{CC}$	V
$V_{IL}$	Input LOW Voltage (TXIN, /TXIN)		0		$V_{IH} - 0.1$	V
$V_{IN}$	Input Voltage Swing (TXIN, /TXIN)	See Figure 5a	0.1		$V_{CC}$	V
$V_{DIFF\_IN}$	Differential Input Voltage Swing $ TXIN - /TXIN $	See Figure 5b	0.2			V
$V_{T\_IN}$	TXIN, /TXIN to VT				1.28	V
$V_{TXVREF-AC}$	Output Reference Voltage		$V_{CC} - 1.4$	$V_{CC} - 1.3$	$V_{CC} - 1.3$	V
$V_{REF\_CTRL}$	Output Reference Voltage		$V_{CC} - 1.4$	$V_{CC} - 1.3$	$V_{CC} - 1.3$	V
$V_{TXVCTRL}$	Input Voltage (TXVCTRL)		$V_{REF\_CTRL}$		$V_{CC}$	V

## Transmitter Output DC Electrical Characteristics

$V_{CC\_TXQ} = 3.3V \pm 10\%$ ,  $R_L = 50\Omega$  to  $V_{CC\_TXQ} - 2V$ ;  $T_A = -40^\circ C$  to  $+85^\circ C$ , unless otherwise stated.

Symbol	Parameter	Condition	Min	Typ	Max	Units
$V_{OH}$	Output HIGH Voltage (TXQ, /TXQ)		$V_{CC\_TXQ} - 1.145$	$V_{CC\_TXQ} - 1.020$	$V_{CC\_TXQ} - 0.895$	V
$V_{OL}$	Output LOW Voltage (TXQ, /TXQ)	$TXVCTRL = V_{REF\_CTRL}$	$V_{CC\_TXQ} - 1.945$	$V_{CC\_TXQ} - 1.820$	$V_{CC\_TXQ} - 1.695$	V
		$TXVCTRL = V_{CC\_TXQ}$		$V_{CC\_TXQ} - 1.100$		V
$V_{OUT}$	Output Voltage Swing (TXQ, /TXQ)	$TXVCTRL = V_{REF\_CTRL}$ See Figure 5a	550	800		mV
		$TXVCTRL = V_{CC\_TXQ}$ See Figure 5a		80		mV
$V_{DIFF\_OUT}$	Differential Output Voltage Swing (TXQ, /TXQ)	$TXVCTRL = V_{REF\_CTRL}$ See Figure 5b	1100	1600		mV
		$TXVCTRL = V_{CC\_TXQ}$ See Figure 5b		160		mV

## LVTTL/CMOS INPUT DC Control Electrical Characteristics<sup>(5)</sup>

$V_{CC} = 3.3V \pm 10\%$ ;  $T_A = -40^\circ C$  to  $+85^\circ C$ , unless otherwise stated.

Symbol	Parameter	Condition	Min	Typ	Max	Units
$V_{IL}$	/TXEN, /RXEN, LOOPBACK				0.8	V
$V_{IH}$	/TXEN, /RXEN, LOOPBACK		2			V
$I_{IL}$	/TXEN, /RXEN, LOOPBACK	$I_{IL}@V_{IN} = 0.5V$	0		50	$\mu A$
$I_{IH}$	/TXEN, /RXEN, LOOPBACK	$I_{IH}@V_{IN} = V_{CC}$			300	$\mu A$

**Note:**

5. /TXEN, /RXEN, and LOOPBACK have an internal pull-down  $25k\Omega$  resistor.

## LOS DC Electrical Characteristics

$V_{CC} = 3.3V \pm 10\%$ ;  $T_A = -40^\circ C$  to  $+85^\circ C$ .

Symbol	Parameter	Condition	Min	Typ	Max	Units
$V_{LOSLVL}$	LOSLVL Voltage Range		$V_{REF}$		$V_{CC}$	V
$V_{OH}$	Output HIGH Voltage	Source $100\mu A$ ; $V_{CC} \geq 3.3V$	2.4			V
$V_{OL}$	Output LOW Voltage	Sink $2mA$			0.5	V
VSR	LOS Sensitivity Range		7		35	$mV_{PP}$
LOS <sub>AL</sub>	Low LOS Assert Level	$R_{LOSLVL} = 10k\Omega$ $2^7-1$ Data Pattern, Note 7				
			622Mbps	15		mV
LOS <sub>DL</sub>	Low LOS De-assert Level	$R_{LOSLVL} = 10k\Omega$ $2^7-1$ Data Pattern, Note 7				
			622Mbps	20		mV
HYS <sub>L</sub>	Low LOS Hysteresis	$R_{LOSLVL} = 10k\Omega$ , limiting mode $2^7-1$ Data Pattern, Note 6 and 7				
			622Mbps	3		dB
LOS <sub>AM</sub>	Medium LOS Assert Level	$R_{LOSLVL} = 5k\Omega$ $2^7-1$ Data Pattern, Note 7				
			622Mbps	20		mV
LOS <sub>DM</sub>	Medium LOS De-assert Level	$R_{LOSLVL} = 5k\Omega$ $2^7-1$ Data Pattern, Note 7				
			622Mbps	30		mV
HYS <sub>M</sub>	Medium LOS Hysteresis	$R_{LOSLVL} = 5k\Omega$ , limiting mode $2^7-1$ Data Pattern, Note 6 and 7				
			622Mbps	4		dB
LOS <sub>AH</sub>	High LOS Assert Level	$R_{LOSLVL} = 1k\Omega$ $2^7-1$ Data Pattern, Note 7				
			622Mbps	35		mV
LOS <sub>DH</sub>	High LOS De-assert Level	$R_{LOSLVL} = 1k\Omega$ $2^7-1$ Data Pattern, Note 7				
			622Mbps	60		mV
HYS <sub>H</sub>	High LOS Hysteresis	$R_{LOSLVL} = 1k\Omega$ , limiting mode $2^7-1$ Data Pattern, Note 6 and 7				
			622Mbps	5		dB
			3.2Gbps	5.5		dB

**Notes:**

6. Hysteresis is defined as:  $20\log_{10} \left( \frac{SD\_AssertVoltage}{SD\_De-assertVoltage} \right) dB$ .
7. See the "Typical Operating Characteristics" section for more details on  $R_{LOSLVL}$  and its associated LOS assert and de-assert amplitudes for a  $2^7-1$  PRBS data pattern. See the "PRBS Discussion" section for more details on the  $2^7-1$  PRBS data pattern.

## AC Electrical Characteristics<sup>(8)</sup>

$V_{CC} = V_{CC\_TXQ} = V_{CC\_RXIN} = 3.3V \pm 10\%$ , Receiver Load:  $R_L = 100\Omega$  across the outputs. Transmitter Load:  $R_L = 50\Omega$  to  $V_{CC\_TXQ} - 2V$ ;  $T_A = -40^\circ C$  to  $+85^\circ C$ , unless otherwise stated.

### Receiver and Transmitter

Symbol	Parameter	Condition	Min	Typ	Max	Units
$t_{JITTER}$	Deterministic Jitter (DJ)	Note 9			Note 13	$\text{ps}_{\text{PP}}$
	Random Jitter (RJ)	Note 10		0.7	5	$\text{ps}_{\text{RMS}}$
	Crosstalk-Induced Jitter	Note 11			1.2	$\text{ps}_{\text{RMS}}$

### Receiver

Symbol	Parameter	Condition	Min	Typ	Max	Units
$f_{MAX}$	Maximum Operating Frequency	$V_{RXIN} \geq 10\text{mV}$ (20mV <sub>PP</sub> )	3.2			Gbps
BW	-3dB	$V_{RXIN} \geq 10\text{mV}$ (20mV <sub>PP</sub> )		2.5		GHz
$S_{21}$	Single-Ended Gain	Linear mode		32		dB
$A_{V(DIFF)}$	Differential Voltage Gain	Linear mode		38		dB
$t_r, t_f$	Output Rise/Fall Time (20% to 80%)	Limiting mode		60	120	ps
LOS Frequency Range	LOS Operating Frequency Range	Note 12	0.622		3.2	Gbps
$t_{OFF}$	LOS De-assert Time			0.1	0.5	$\mu\text{s}$
$t_{ON}$	LOS Assert Time			0.2	0.5	$\mu\text{s}$

### Transmitter

Symbol	Parameter	Condition	Min	Typ	Max	Units
$f_{MAX}$	Maximum Operating Frequency	$V_{TXIN} \geq 100\text{mV}$ (200mV <sub>PP</sub> )	3.2			Gbps
BW	-3dB	$V_{REF\_CTRL} \leq TXCTRL \leq V_{CC\_TXQ}$		2		GHz
$t_r, t_f$	Output Rise/Fall Time (20% to 80%)	$V_{TXVCTRL} = V_{REF\_CTRL}$		100	160	ps

#### Notes:

8. High-frequency AC-parameters are guaranteed by design and characterization.
9. Deterministic jitter is measured with both K28.5 and  $2^{23}-1$  PRBS data-pattern, measured at  $<f_{MAX}$ .  $V_{IN} = 10\text{mV}$  (20mV<sub>PP</sub>) RX,  $100\text{mV}$  (200mV<sub>PP</sub>) TX. See the "PRBS Discussion" section for more details on the K28.5 and  $2^{23}-1$  PRBS data pattern.
10. Random jitter is measured with a K28.7 character pattern, measured at  $<f_{MAX}$ .  $V_{IN} = 10\text{mV}$  (20mV<sub>PP</sub>) RX,  $100\text{mV}$  (200mV<sub>PP</sub>) TX. See the "PRBS Discussion" section for more details on the K28.7 PRBS data pattern.
11. Crosstalk is measured at the output while applying two similar differential clock frequencies that are asynchronous with respect to each other at the inputs.
12. LOS is guaranteed to be chatter-free at  $f_{MAX} \geq 622\text{Mbps}$  or  $f_{MAX} \geq 311\text{MHz}$  with  $V_{RXIN} \geq 10\text{mV}$  (20mV<sub>PP</sub>) with a  $2^7-1$  PRBS data pattern.
13. Contact factory for limits.

## Detailed Description

## Receiver

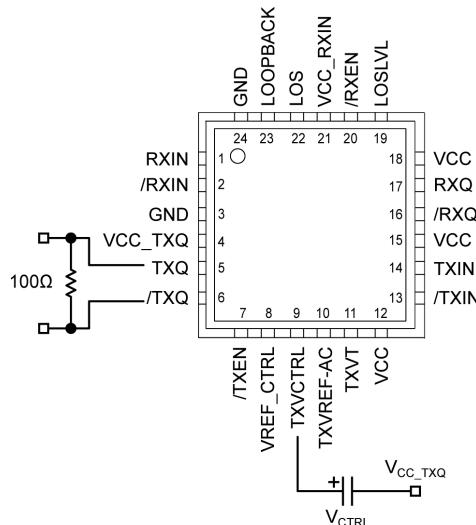
The receiver AC-coupled differential input distributes data to 3.2Gbps with signals as small as 10mV (20mV<sub>PP</sub>) or as large as 900mV (1.8V<sub>PP</sub>). The receiver input features an internal 50Ω input termination connected to an internal reference which optimizes the inputs for AC-coupled signals. Input signals are linearly amplified with 38dB of differential gain and the output signal is limited to 400mV (800mV<sub>PP</sub>).

The receiver output buffer features  $50\Omega$  source termination resistors and a current source that provides  $400\text{mV}$  ( $800\text{mV}_{\text{PP}}$ ) swing into  $50\Omega$  termination. The output buffers terminates to standard CML loads ( $100\Omega$  across the output pair or equivalent). See the “Output Stage Receiver” section for more details.

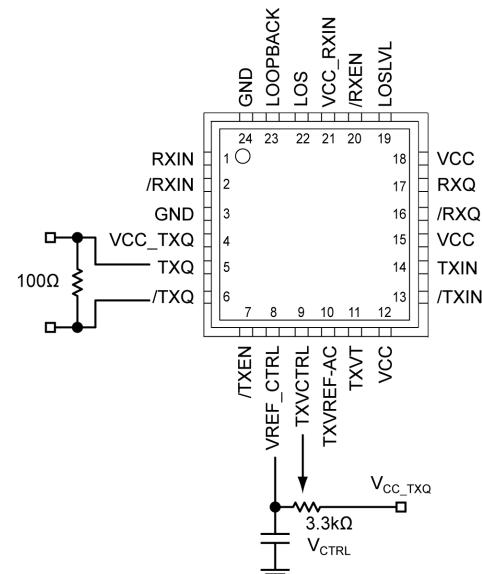
## Transmitter

The transmitter differential input includes Micrel's unique, patented 3-pin input termination architecture that directly interfaces to any (AC- or DC-coupled) differential signal as small as 100mV (200mV<sub>PP</sub>) without any termination resistor network in the signal path.

The transmitter output buffer terminates to standard LVPECL loads ( $R_L = 50$  to  $V_{CC\_TXQ} - 2V$ ). The output buffer is a special variable swing LVPECL buffer controlled by TXVCTRL. The output buffer features emitter follower output that provides 80mV (160mV<sub>PP</sub>) to 800mV (1.6V<sub>PP</sub>) swing into 50<sub>Ω</sub> transmission lines. See the next section and Figures 1a and 1b for more details on how to control the variable output swing feature.



**Figure 1a. Voltage Source Implementation**



### Figure 1b. Alternative Implementation

## Transmitter PECL Variable-Swing Output Buffer

- Connecting  $V_{REF\_CTRL}$  to TXVCTRL sets the transmitter output buffer to maximum swing
- Setting TXVCTRL to  $V_{CC\_TXQ}$ , sets the transmitter output buffer to minimum swing
- Control of the transmitter output swing buffers can be obtained by using a variable resistor connected between  $V_{REF\_CTRL}$  and  $V_{CC\_TXQ}$  with a wiper connected to TXVCTRL as shown in Figure 1b

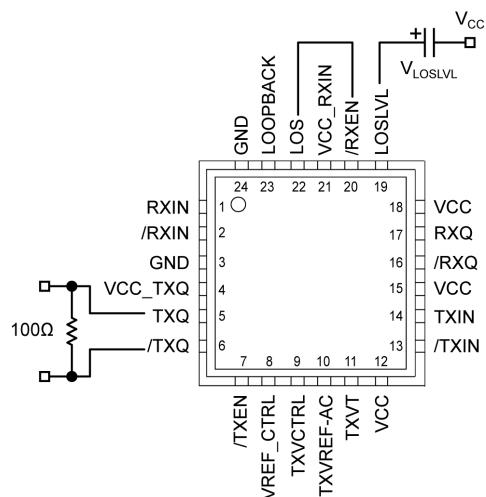
## Receiver LOS

The SY58621L features a chatter-free Loss-of-Signal (LOS) TTL compatible output with an internal 4.75k $\Omega$  pull-up resistor. LOS circuitry monitors the input receiver signal and asserts a signal when the input signal falls below the threshold set by the programmable LOS level set pin (LOSLVL). When the amplitude of the receiver input signal falls below the threshold, LOS is asserted HIGH with a response time of  $\sim$ 0.2uS. LOS can be fed into /RXEN to maintain output stability by disabling the output during a Loss-of-Signal condition. Figure 2a and 2b shows the LOS connection to /RXEN. When /RXEN is HIGH, the output signal RXQ is held LOW and /RXQ is held HIGH. Typically, 2dB of LOS hysteresis is adequate to prevent the receiver output from chattering. LOS operation is optimized for data rates  $\geq$ 622Mbps with an input receiver amplitude of at least 10mV (20mV<sub>PP</sub>). Due to the long time constant in slower data rates below 622Mbps, the SY58621L LOS function does not guarantee chatter-free operation for low amplitude signals.

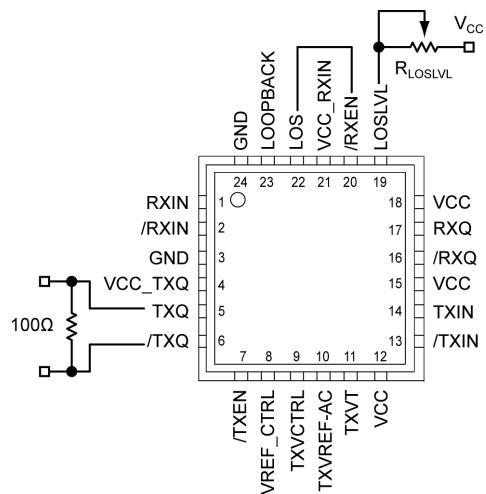
LOSLVL sets the threshold of the LOS input amplitude detection. Connecting an external resistor,  $R_{LOSLVL}$ ,

between VCC and LOSLVL sets the input amplitude LOS detection trip-point by setting up a voltage divider between VCC and  $V_{REF}$  (an internal voltage source set at  $V_{CC}-1.3V$ ), since there is a  $2.8k\Omega$  internal resistor connected between LOSLVL and  $V_{REF}$ . The input voltage range of LOSLVL ranges from  $V_{CC}$  to  $V_{REF}$ . See the “Functional Block Diagram” section and Figures 2a and 2b, to see how  $R_{LOSLVL}$  sets up a voltage divider between VCC and  $V_{REF}$ . See the “LOS Output DC Electrical Characteristics” table and “Typical Operating Characteristics” section to see how different  $R_{LOSLVL}$  values affect LOS sensitivity.

connected to VCC with a wiper connected to LOSLVL, as shown in Figure 2b



**Figure 2a. Voltage Source Implementation**



### Figure 2b. Alternative Implementation

## LOS Output

- Connecting the input /RXEN to the LOS output as shown in Figures 2a and 2b, maintains receiver output stability under a Loss-of-Signal condition
- Sensitivity of the LOS signal can be programmed using the LOSLVL input by using a variable resistor

- $\geq 2\text{dB}$  hysteresis is insured if  $R_{\text{LOSLVL}} \leq 10\text{k}\Omega$
- LOS is guaranteed chatter-free at  $f \geq 622\text{Mbps}$  (311MHz)

### Hysteresis

The SY58621L provides a minimum of 2dB of LOS hysteresis, see the Figure 3 for more details.

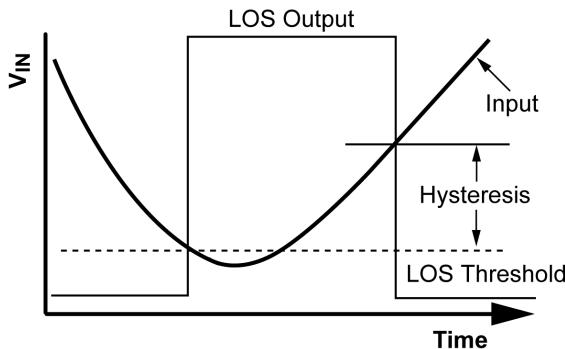


Figure 3. LOS Hysteresis Assert/De-assert

Hysteresis is defined as:  $20\log_{10}\left(\frac{SD\_AssertVoltage}{SD\_De-assertVoltage}\right)\text{dB}$ .

### Loopback

To support diagnostic system testing, the SY58621L features a loopback test mode, activated by setting LOOPBACK to logic HIGH. Loopback mode enables an internal loopback path from the transmitter input to the receiver output and supports the full 3.2Gbps data rate throughput.

### Crosstalk

The SY58621L features a patent-pending isolation between the receiver and transmitter channels. The following guide lines can be used to minimize on layout induced crosstalk:

#### 1. *Ground Stripping*

Ground stripping is an effective method to reduce crosstalk. Ground stripping involves running a ground trace between the receiver and transmitter channels.

#### 2. *Vertical and Horizontal Traces*

Another way to reduce crosstalk is to route the receiver and transmitter channels on separate layers with an embedded ground or power supply layer between the layers. When routing the traces on different layers, run the receiver traces horizontal to the transmitter traces and route the transmitter traces vertical to the receiver traces.

## PRBS Discussion

### LOS Testing

The LOS function is tested with a  $2^7\text{-}1$  PRBS (Pseudo Random Bit Stream) data-pattern. A PRBS data-pattern of  $2^7\text{-}1$  is used because it is a good approximation to an 8b10b-encoded NRZ data stream. 8b10b encodes 8 bits of data and replaces it with 10 bits of symbol. The extra bits are added to improve transition density and the BER (Bit Error Rate) of the system.

### Deterministic Jitter Testing and the K28.5 Pattern

The K28.5 (11000001010011111010) and  $2^{23}\text{-}1$  PRBS data-patterns are used to characterize DJ because both data patterns have lower spectral frequency content which provides a best approximation to scrambled NRZ data streams.

### Random Jitter Testing and the K28.7 Pattern

The K28.7 (1111100000...) data pattern is used to measure RJ since the pattern is free of DJ. In addition, because the K28.7 data-pattern can be used to compare the  $T_N$  ( $N^{\text{TH}}$  period) to the  $T_0$  (1<sup>st</sup> period), low frequency jitter components can be accumulated.

## Power Supply Filtering

Although the SY58621L is fully differential, it is recommended that the power supplies are filtered as shown in Figure 4.

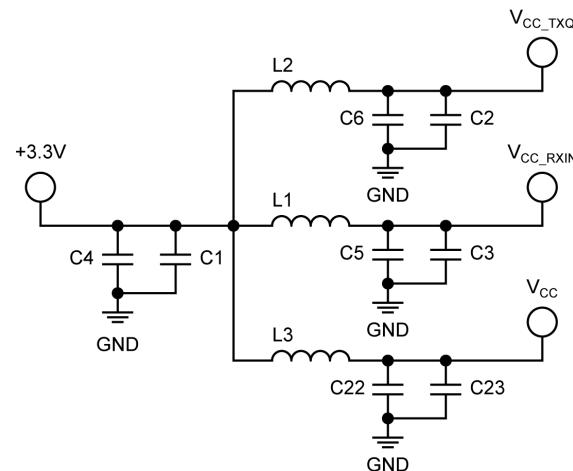


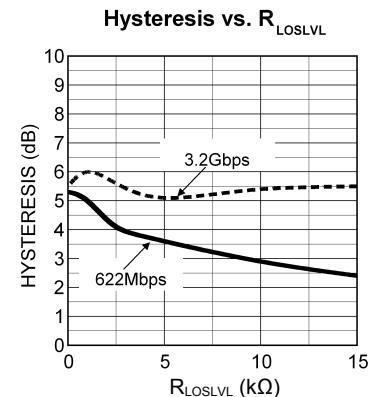
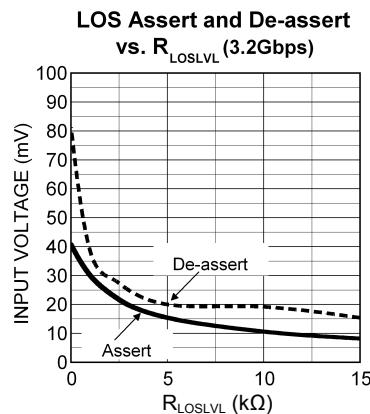
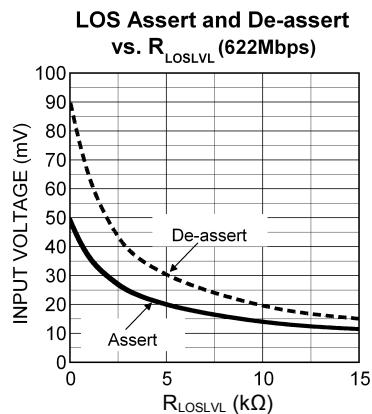
Figure 4. Power Supply Filtering Scheme

Item	Description
C1, C2, C3, C23	0.1 $\mu\text{F}$ Capacitor
C4, C5, C6, C22	0.01 $\mu\text{F}$ Capacitor
L1, L2, L3	1.2 $\mu\text{H}$ Ferrite Bead Inductor

Table 1. Bill of Materials

## Typical Operating Characteristics

$V_{CC} = V_{CC\_TXQ} = V_{CC\_RXIN} = 3.3V \pm 10\%$ , Receiver:  $R_L = 100\Omega$  across the outputs. Transmitter:  $R_L = 50\Omega$  to  $V_{CC\_TXQ}$   $-2V$ ;  $T_A = 25^\circ C$ , unless otherwise stated.



## Single-Ended and Differential Swings

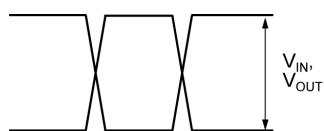


Figure 5a. Single-Ended Voltage Swing

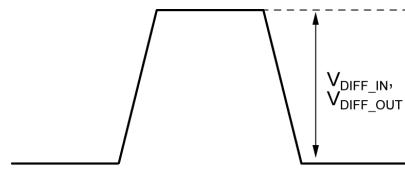


Figure 5b. Differential Voltage Swing

## Input Stage

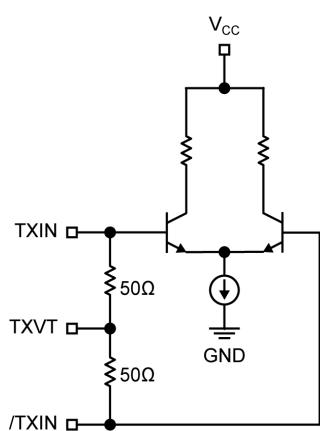


Figure 6a. TX Simplified Differential Input Stage

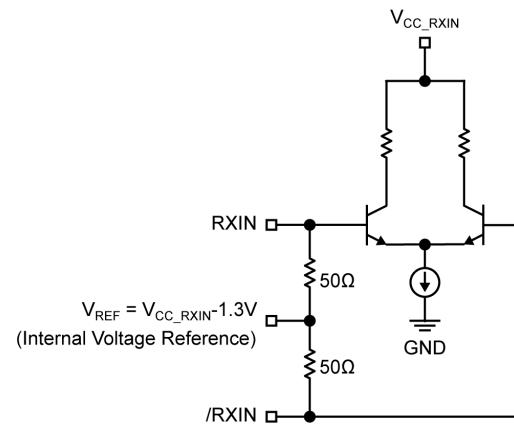


Figure 6b. RX Simplified Differential Input Stage

## Output Stage

### Receiver

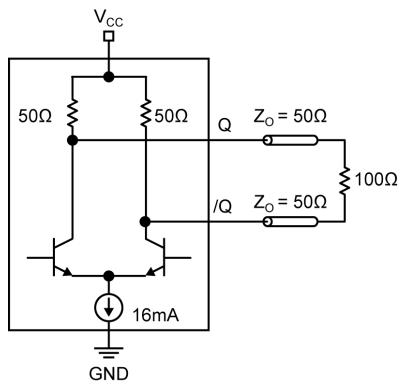


Figure 7a. Receiver CML DC-Coupled Output

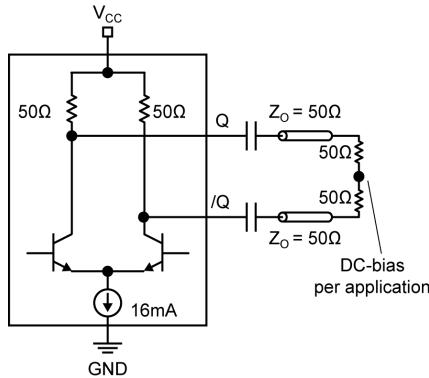


Figure 7b. Receiver CML AC-Coupled Output

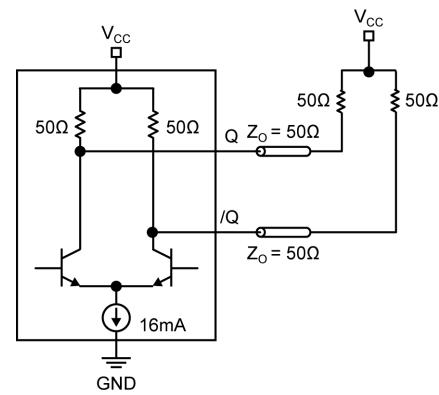


Figure 7c. Receiver CML DC-Coupled Output (50Ω to Vcc)

## Transmitter

The transmitters output is a variable swing LVPECL open emitter driver. LVPECL has very low output (open emitter) impedance, and small signal swing which result in low EMI.

LVPECL is ideal for driving  $50\Omega$  and  $100\Omega$ -controlled impedance transmission lines. There are several techniques for terminating the LVPECL output: Parallel Termination-Thevenin Equivalent, Parallel Termination (3-Resistor), and AC-Coupled Termination. Unused output pairs may be left floating. However, the unused half of a single-ended output must be terminated, or balanced.

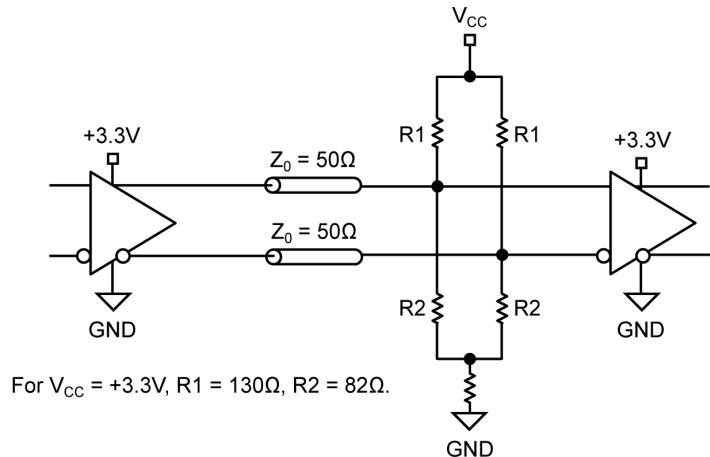


Figure 8a. Parallel Thevenin-Equivalent Termination

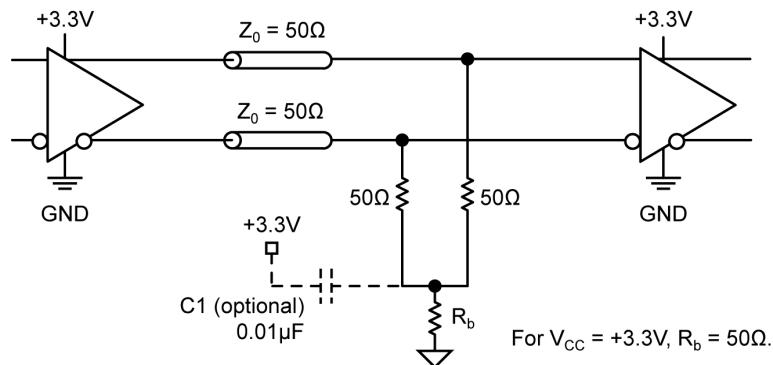


Figure 8b. Parallel Termination – 3-Resistors

## Interface Applications

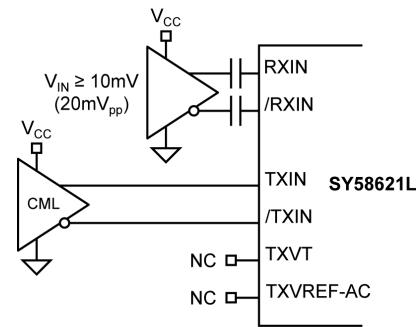
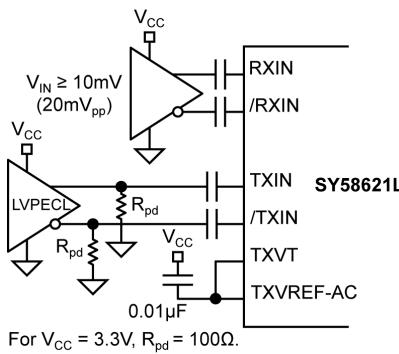
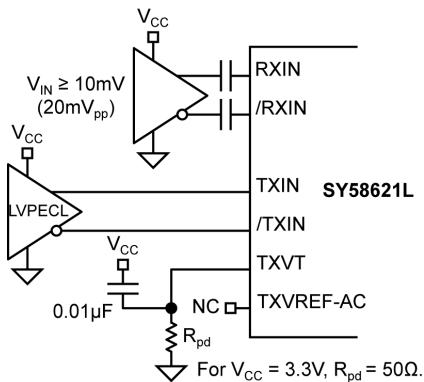


Figure 9c. CML Interface (TX DC-Coupled/RX AC-Coupled)

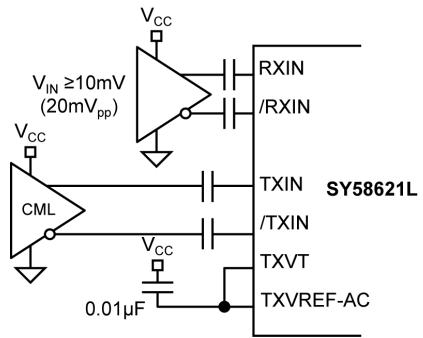


Figure 9d. CML Interface (TX AC-Coupled/RX AC-Coupled)

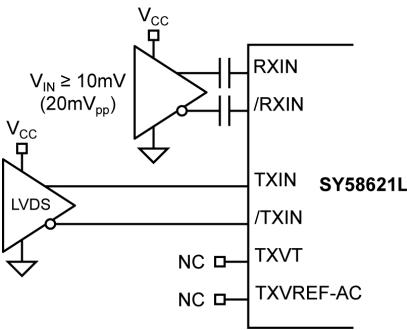
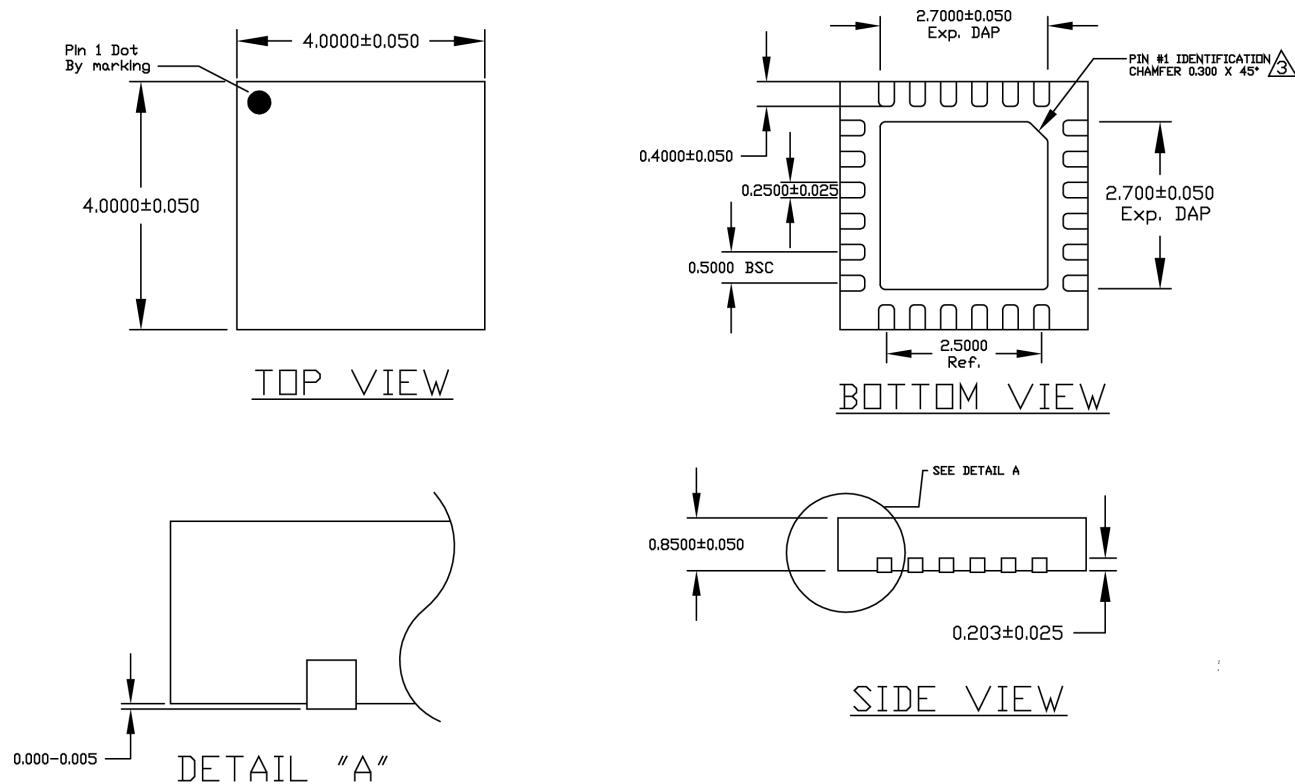


Figure 9e. LVDS Interface (TX DC-Coupled/RX AC-Coupled)

## Related Product and Support Documentation

Part Number	Function	Data Sheet Link
SY58620L	Precision 4.25Gbps CML Transceiver with Integrated Loopback	<a href="http://www.micrel.com/product-info/products/sy58620l.shtml">www.micrel.com/product-info/products/sy58620l.shtml</a>
HBW Solutions	New Products and Applications	<a href="http://www.micrel.com/product-info/products/solutions.shtml">www.micrel.com/product-info/products/solutions.shtml</a>

## Package Information



## NOTE:

1. ALL DIMENSIONS ARE IN MILLIMETERS (mm).
2. THE PIN#1 IDENTIFIER MUST EXIST ON THE TOP SURFACE OF PACKAGE BY USING IDENTIFICATION MARK OR OTHER FEATURE OF PACKAGE BODY.

**△** CHAMFER STYLE PIN 1 IDENTIFIER ON BOTTOM SIDE

## 24-Pin QFN

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