

NCV4275C

PIN FUNCTION DESCRIPTION

Pin No.	Symbol	Description
1	I	Input; Battery Supply Input Voltage. Bypass to ground with a ceramic capacitor.
2	RO	Reset Output; Open Collector Active Reset (accurate when I > 1.0 V).
3, TAB	GND	Ground; Pin 3 internally connected to tab.
4	D	Reset Delay; timing capacitor to GND for Reset Delay function.
5	Q	Output; $\pm 2.0\%$, 450 mA output. Bypass with 22 μF capacitor, ESR < 4.5 Ω (5.0 V Version), 3.5 Ω (3.3 V Version).

MAXIMUM RATINGS

Rating	Symbol	Min	Max	Unit	
Input Voltage	V_I	-42	45	V	
Input Peak Transient Voltage	V_I	-	45	V	
Output Voltage	V_Q	-1.0	16	V	
Reset Output Voltage	V_{RO}	-0.3	25	V	
Reset Output Current	I_{RO}	-5.0	5.0	mA	
Reset Delay Voltage	V_D	-0.3	7.0	V	
Reset Delay Current	I_D	-2.0	2.0	mA	
ESD Susceptibility (Note 1)	- Human Body Model	ESD_{HBM}	4.0	-	kV
	- Machine Model	ESD_{MM}	200	-	V
	- Charge Device Model	ESD_{CDM}	1000	-	V
Junction Temperature	T_J	-40	150	$^{\circ}\text{C}$	
Storage Temperature	T_{stg}	-55	150	$^{\circ}\text{C}$	

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. This device incorporates ESD protection and is tested by the following methods: ESD Human Body Model tested per AEC-Q100-002, ESD Machine Model tested per AEC-Q100-003, ESD Charged Device Model tested per AEC-Q100-011, Latch-up tested per AEC-Q100-004.

OPERATING RANGE

Rating	Symbol	Min	Max	Unit
Input Voltage Operating Range, 5.0 V Output	V_I	5.5	42	V
Input Voltage Operating Range, 3.3 V Output	V_I	4.4	42	V
Junction Temperature	T_J	-40	150	$^{\circ}\text{C}$

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

LEAD TEMPERATURE SOLDERING REFLOW AND MSL (Note 2)

Rating	Symbol	Min	Max	Unit
Lead Free, 60 sec-150 sec above 217 $^{\circ}\text{C}$	T_{SLD}	-	265 Peak	$^{\circ}\text{C}$
Moisture Sensitivity Level	MSL	1		

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THERMAL CHARACTERISTICS

Characteristic	Test Conditions (Typical Value)		Unit
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DPAK 5-PIN PACKAGE

	Min Pad Board (Note 3)	1 in Pad Board (Note 4)	
Junction-to-Tab ($R_{\theta JT}$)	5.1	5.5	°C/W
Junction-to-Ambient ($R_{\theta JA}$)	82.4	58.1	°C/W

D²PAK 5-PIN PACKAGE

	0.4 sq. in. Spreader Board (Note 5)	1.2 sq. in. Spreader Board (Note 6)	
Junction-to-Tab ($R_{\theta JT}$)	4.5	4.8	°C/W
Junction-to-Ambient ($R_{\theta JA}$)	66.0	49.0	°C/W

2. PR_R IPC / JEDEC J-STD-020C
3. 1 oz. copper, 0.26 inch² (168 mm²) copper area, 0.062" thick FR4.
4. 1 oz. copper, 1.14 inch² (736 mm²) copper area, 0.062" thick FR4.
5. 1 oz. copper, 0.373 inch² (241 mm²) copper area, 0.062" thick FR4.
6. 1 oz. copper, 1.222 inch² (788 mm²) copper area, 0.062" thick FR4.

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ELECTRICAL CHARACTERISTICS ($V_I = 13.5\text{ V}$; $-40^\circ\text{C} < T_J < 150^\circ\text{C}$; unless otherwise noted.)

Characteristic	Symbol	Test Conditions	Min	Typ	Max	Unit
Output						
Output Voltage	V_Q	$100\ \mu\text{A} \leq I_Q \leq 400\ \text{mA}$ $6.0\ \text{V} \leq V_I \leq 28\ \text{V}$ (5.0 V Version) $4.4\ \text{V} \leq V_I \leq 28\ \text{V}$ (3.3 V version)	4.9 3.23 (2%)	5.0 3.3	5.1 3.37 (2%)	V
Output Voltage	V_Q	$100\ \mu\text{A} \leq I_Q \leq 200\ \text{mA}$ $6.0\ \text{V} \leq V_I \leq 40\ \text{V}$ (5.0 V Version) $4.4\ \text{V} \leq V_I \leq 40\ \text{V}$ (3.3 V version)	4.9 3.23 (2%)	5.0 3.3	5.1 3.37 (2%)	V
Output Current Limitation	I_Q	$V_Q = 0.9 \times V_{Q,\text{typ}}$	450	650	–	mA
Quiescent Current $I_q = I_I - I_Q$	I_q	$I_Q = 1.0\ \text{mA}$, $T_J = 25^\circ\text{C}$	–	135	150	μA
		$I_Q = 1.0\ \text{mA}$	–	150	200	μA
		$I_Q = 250\ \text{mA}$	–	10	15	mA
		$I_Q = 400\ \text{mA}$	–	23	35	mA
Dropout Voltage (Note 7)	V_{dr}	$I_Q = 300\ \text{mA}$ $V_{\text{dr}} = V_I - V_Q$	–	250	500	mV
Load Regulation	ΔV_Q	$I_Q = 5.0\ \text{mA}$ to $400\ \text{mA}$	–30	15	30	mV
Line Regulation	ΔV_Q	$\Delta V_I = 8.0\ \text{V}$ to $32\ \text{V}$, $I_Q = 5.0\ \text{mA}$	–15	5.0	15	mV
Power Supply Ripple Rejection	PSRR	$f_r = 100\ \text{Hz}$, $V_r = 0.5\ V_{\text{pp}}$	–	60	–	dB
Temperature Output Voltage Drift	dV_Q/dT	–	–	0.5	–	mV/K

Reset Timing D and Output RO

Reset Switching Threshold 5.0 V Version 3.3 V Version	$V_{Q,\text{rt}}$	V_{out} decreasing $V_{\text{in}} > 5.5\ \text{V}$ $V_{\text{in}} > 4.4\ \text{V}$	90 90	93 93	96 96	% V_{out}
Reset Output Low Voltage	V_{ROL}	$R_{\text{ext}} \geq 5.0\ \text{k}\Omega$, $V_Q \geq 1.0\ \text{V}$	–	0.2	0.4	V
Reset Output Leakage Current	I_{ROH}	$V_{\text{ROH}} = 5.0\ \text{V}$	–	0	10	μA
Reset Charging Current	$I_{\text{D,C}}$	$V_D = 1.0\ \text{V}$	3.0	5.5	9.0	μA
Upper Timing Threshold	V_{DU}	–	1.5	1.8	2.2	V
Lower Timing Threshold	V_{DL}	–	0.2	0.4	0.7	V
Reset Delay Time	t_{rd}	$C_D = 47\ \text{nF}$	10	16	22	ms
Reset Reaction Time	t_{rr}	$C_D = 47\ \text{nF}$	–	1.5	4.0	μs

Thermal Shutdown

Shutdown Temperature (Note 8)	T_{SD}	–	150	–	210	$^\circ\text{C}$
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Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

7. Only for 5 V Version. Measured when the output voltage V_Q has dropped 100 mV from the nominal value obtained at $V_I = 13.5\ \text{V}$.

8. Guaranteed by design, not tested in production.

TYPICAL PERFORMANCE CHARACTERISTICS

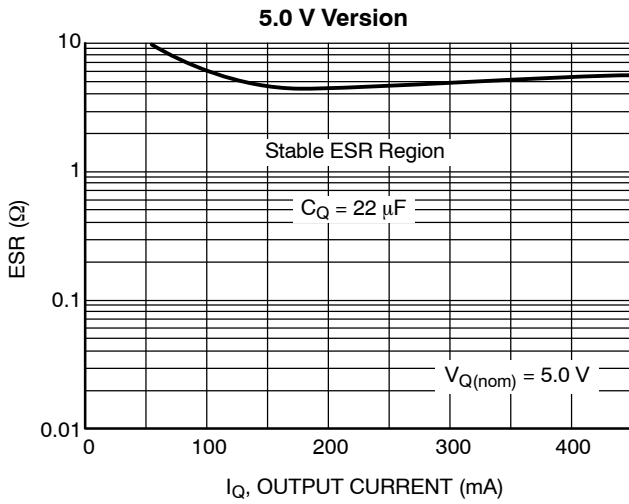


Figure 2. Output Stability with Output Capacitor ESR

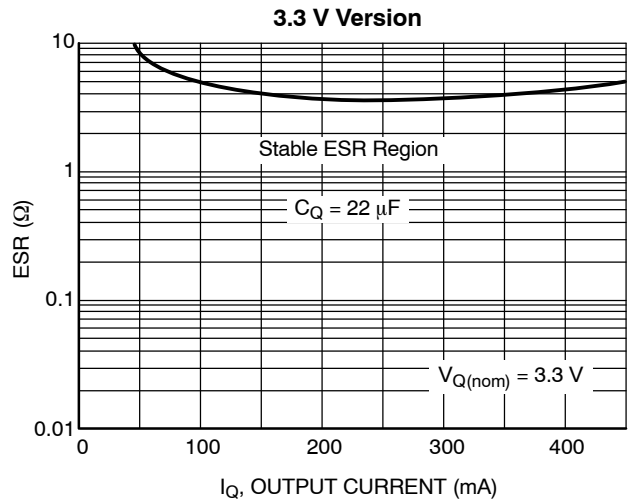


Figure 3. Output Stability with Output Capacitor ESR

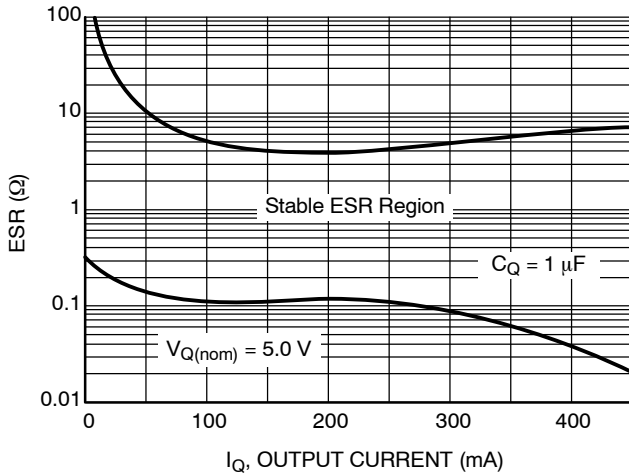


Figure 4. Output Stability with Output Capacitor ESR

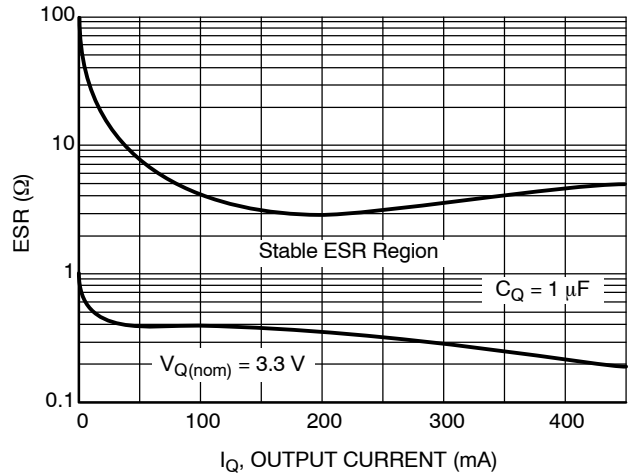


Figure 5. Output Stability with Output Capacitor ESR

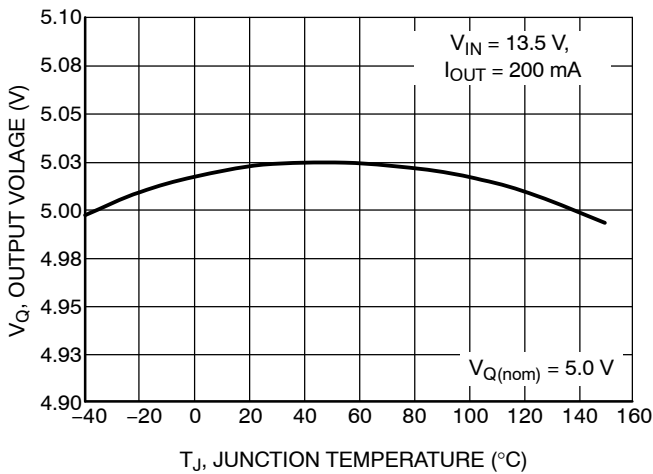


Figure 6. Output Voltage V_Q vs. Temperature T_J

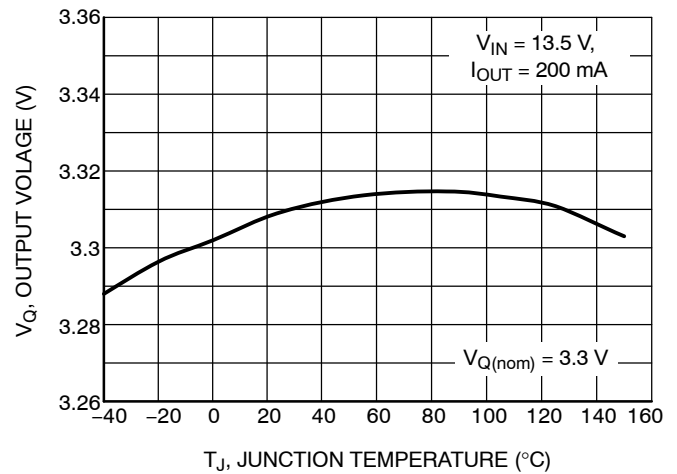


Figure 7. Output Voltage V_Q vs. Temperature T_J

TYPICAL PERFORMANCE CHARACTERISTICS

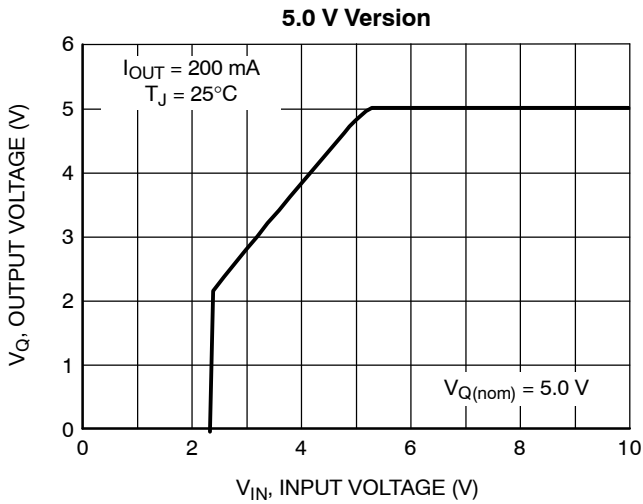


Figure 8. Output Voltage V_Q vs. Input Voltage V_{IN}

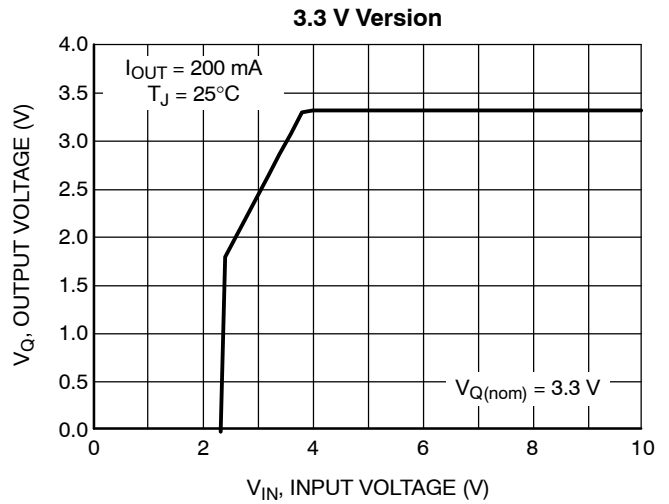


Figure 9. Output Voltage V_Q vs. Input Voltage V_{IN}

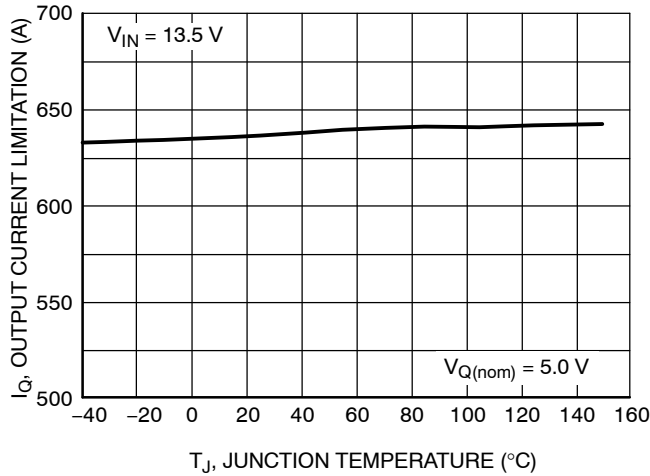


Figure 10. Output Current I_Q vs. Temperature T_J

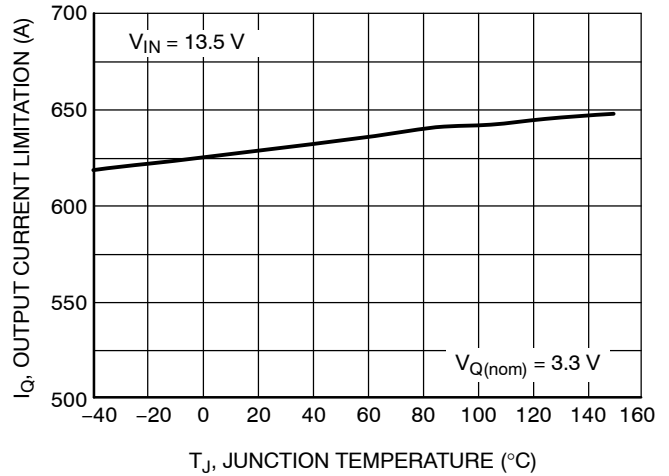


Figure 11. Output Current I_Q vs. Temperature T_J

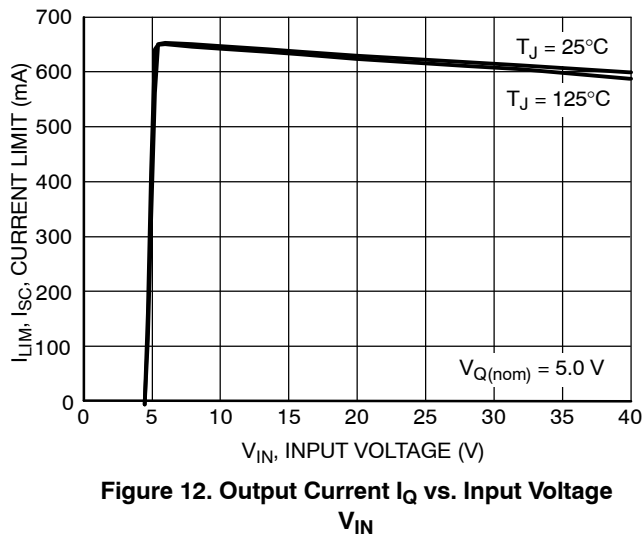


Figure 12. Output Current I_Q vs. Input Voltage V_{IN}

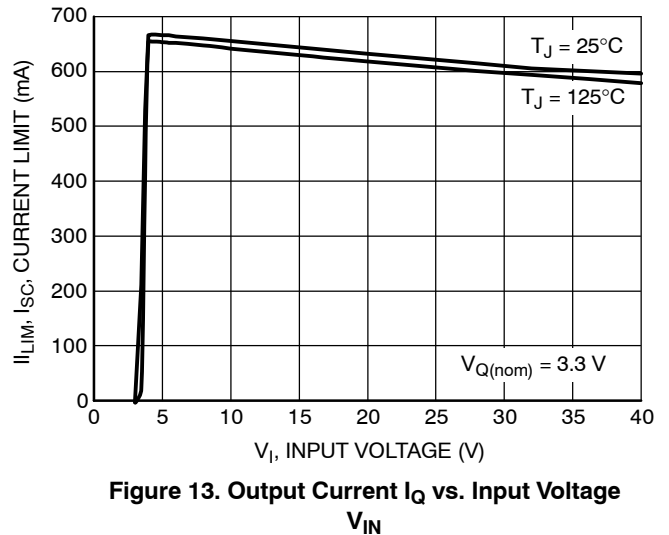


Figure 13. Output Current I_Q vs. Input Voltage V_{IN}

TYPICAL PERFORMANCE CHARACTERISTICS

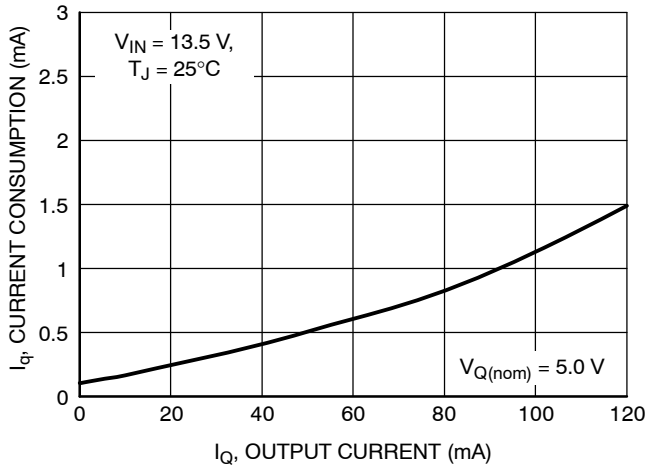


Figure 14. Current Consumption I_{q1} vs. Output Current I_Q

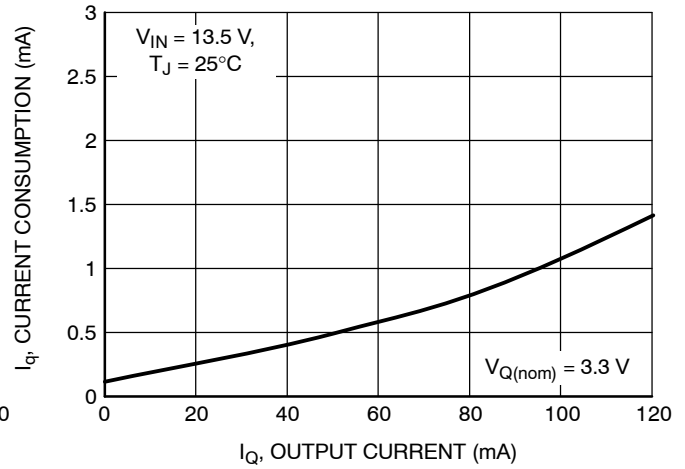


Figure 15. Current Consumption I_{q1} vs. Output Current I_Q

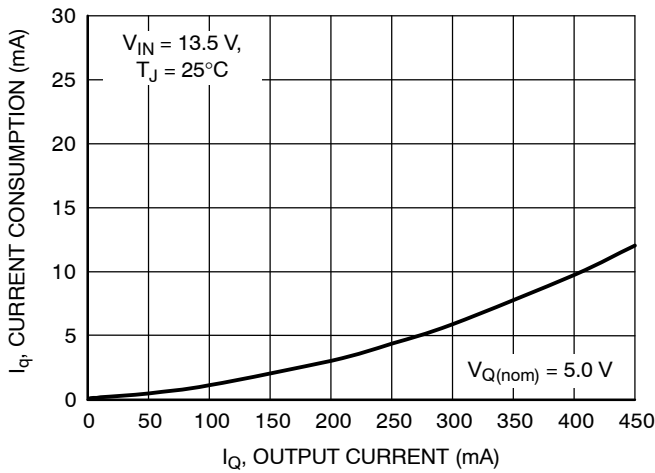


Figure 16. Current Consumption I_{q1} vs. Output Current I_Q

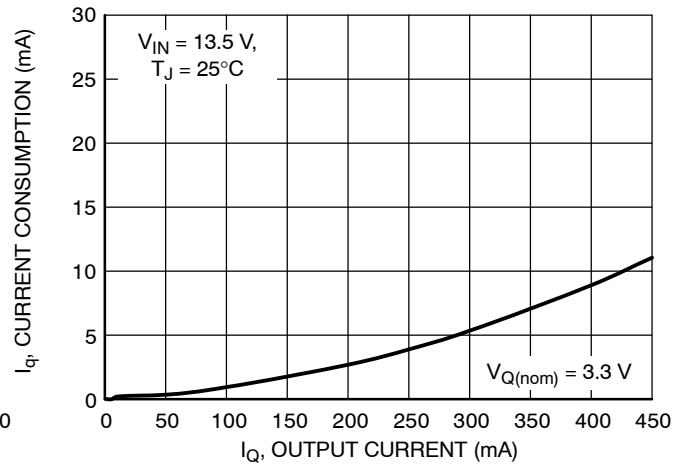


Figure 17. Current Consumption I_{q1} vs. Output Current I_Q

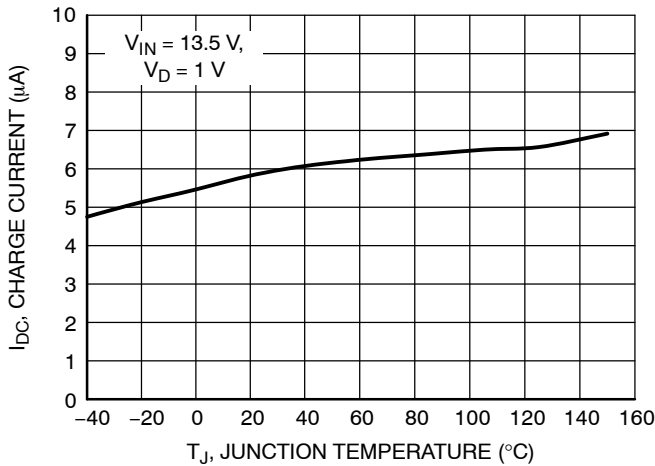


Figure 18. Charge Current $I_{D,C}$ vs. Temperature T_J

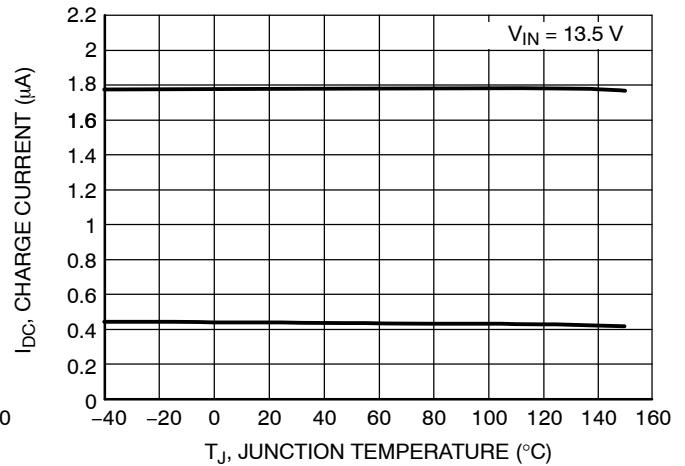


Figure 19. Delay Switching Threshold V_{DU} , V_{DL} vs. Temperature T_J

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TYPICAL PERFORMANCE CHARACTERISTICS

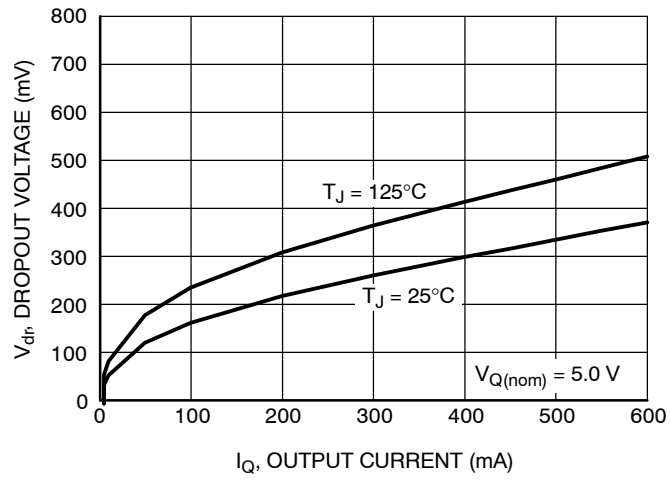


Figure 20. Drop Voltage V_{dr} vs. Output Current I_Q

APPLICATION INFORMATION

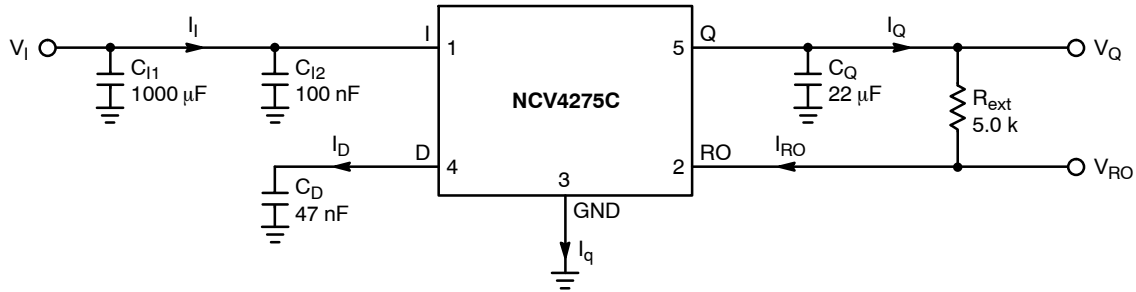


Figure 21. Test Circuit

Circuit Description

The NCV4275C is an integrated low dropout regulator that provides 5.0 V or 3.3 V, 450 mA protected output and a signal for power on reset. The regulation is provided by a PNP pass transistor controlled by an error amplifier with a bandgap reference, which gives it the lowest possible drop out voltage and best possible temperature stability. The output current capability is 450 mA, and the base drive quiescent current is controlled to prevent over saturation when the input voltage is low or when the output is overloaded. The regulator is protected by both current limit and thermal shutdown. Thermal shutdown occurs above 150°C to protect the IC during overloads and extreme ambient temperatures. The delay time for the reset output is adjustable by selection of the timing capacitor. See Figure 21, Test Circuit, for circuit element nomenclature illustration.

Regulator

The error amplifier compares the reference voltage to a sample of the output voltage (V_Q) and drives the base of a PNP series pass transistor by a buffer. The reference is a bandgap design to give it a temperature-stable output. Saturation control of the PNP is a function of the load current and input voltage. Over saturation of the output power device is prevented, and quiescent current in the ground pin is minimized.

Regulator Stability Considerations

The input capacitors (C_{I1} and C_{I2}) are necessary to stabilize the input impedance to avoid voltage line influences. Using a resistor of approximately 1.0 Ω in series with C_{I2} can stop potential oscillations caused by stray inductance and capacitance.

The output capacitor helps determine three main characteristics of a linear regulator: startup delay, load transient response and loop stability. The capacitor value and type should be based on cost, availability, size and temperature constraints. A tantalum, aluminum or ceramic capacitors can be used. The range of stability versus capacitance, load current and capacitive ESR is illustrated

in Figures 2 to 5. Minimum ESR for $C_Q = 22 \mu F$ is native ESR of ceramic capacitors. The aluminum electrolytic capacitor is the least expensive solution, but, if the circuit operates at low temperatures ($-25^\circ C$ to $-40^\circ C$), both the capacitance and ESR of the capacitor will vary considerably. The capacitor manufacturer’s data sheet usually provides this information.

The value for the output capacitor C_Q shown in Figure 21, Test Circuit, should work for most applications; however, it is not necessarily the optimized solution. Stability is guaranteed for $C_Q \geq 22 \mu F$ and an $ESR \leq 4.5 \Omega$ (5.0 V Version), 3.5Ω (3.3 V Version). ESR characteristics were measured with ceramic capacitors and additional resistors to emulate ESR. Murata ceramic capacitors were used, GRM32ER71A226ME20 (22 μF, 10 V, X7R, 1210), GRM31MR71E105KA01 (1 μF, 25 V, X7R, 1206).

Reset Output

The reset output is used as the power on indicator to the microcontroller. This signal indicates when the output voltage is suitable for reliable operation of the controller. It pulls low when the output is not considered to be ready. RO is pulled up to V_Q by an external resistor, typically 5.0 kΩ in value. The input and output conditions that control the Reset Output and the relative timing are illustrated in Figure 22, Reset Timing.

Output voltage regulation must be maintained for the delay time before the reset output signals a valid condition. The delay for the reset output is defined as the amount of time it takes the timing capacitor on the delay pin to charge from a residual voltage of 0.0 V to the upper timing threshold voltage V_{DU} . The charging current for this is $I_{D,C}$ and D pin voltage in steady state is typically 2.4 V. By using typical IC parameters with a 47 nF capacitor on the D pin, the following time delay for 5.0 V regulator is derived:

$$t_{RD} = C_D V_{DU} / I_{D,C}$$

$$t_{RD} = 47 \text{ nF} (1.8 \text{ V}) / 5.5 \mu A = 15.4 \text{ ms}$$

Other time delays can be obtained by changing the capacitor value.

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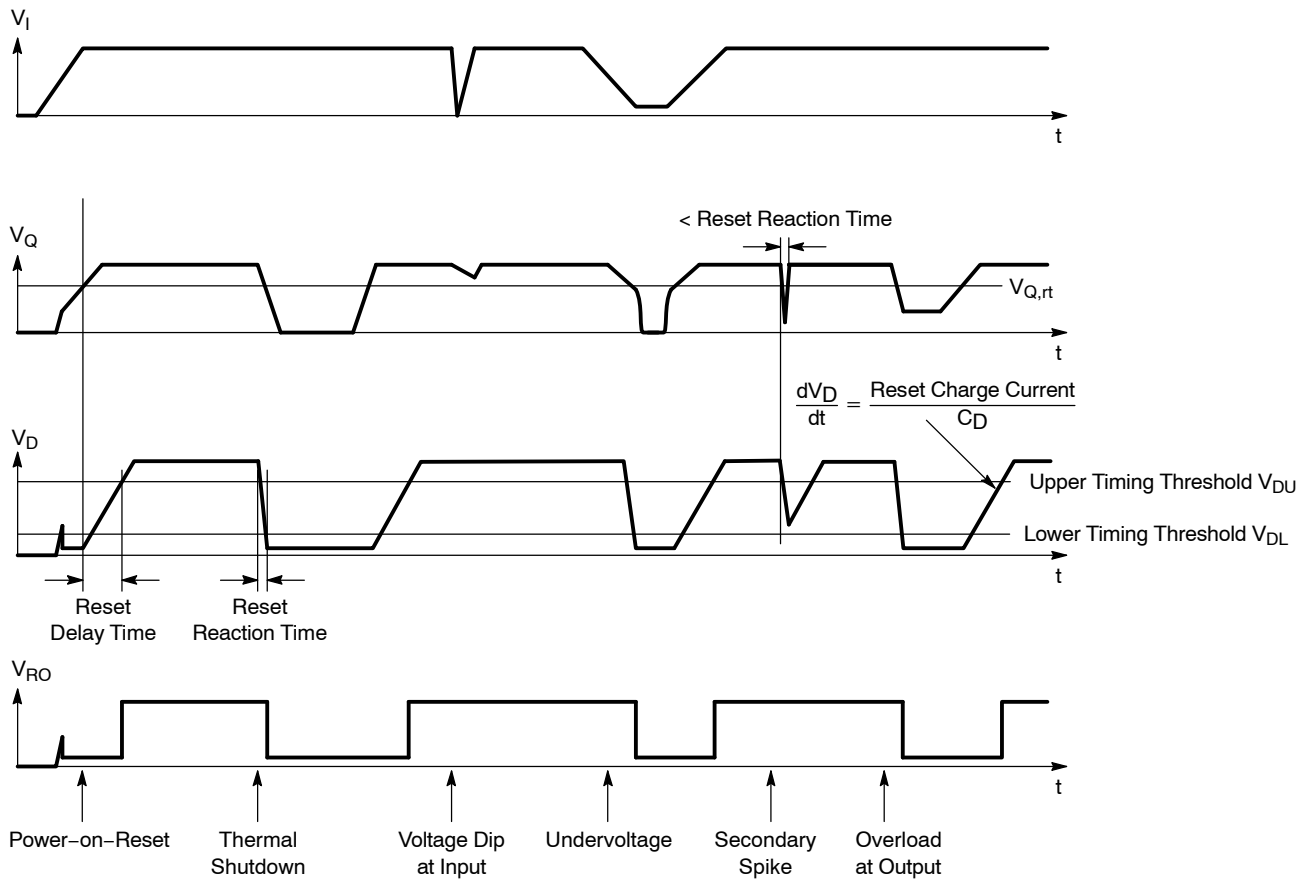


Figure 22. Reset Timing

Calculating Power Dissipation in a Single Output Linear Regulator

The maximum power dissipation for a single output regulator (Figure 23) is:

$$PD(max) = [V_{I(max)} - V_{Q(min)}] I_{Q(max)} + V_{I(max)} I_q \tag{1}$$

where

$V_{I(max)}$ is the maximum input voltage,

$V_{Q(min)}$ is the minimum output voltage,

$I_{Q(max)}$ is the maximum output current for the application,

I_q is the quiescent current the regulator consumes at $I_{Q(max)}$.

Once the value of $P_{D(max)}$ is known, the maximum permissible value of $R_{\theta JA}$ can be calculated:

$$R_{\theta JA} = \frac{150^{\circ}C - T_A}{P_D} \tag{2}$$

The value of $R_{\theta JA}$ can then be compared with those in the package section of the data sheet. Those packages with $R_{\theta JA}$'s less than the calculated value in Equation 2 will keep the die temperature below 150°C.

In some cases, none of the packages will be sufficient to dissipate the heat generated by the IC, and an external heatsink will be required.

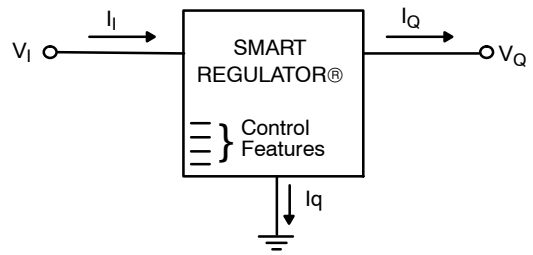


Figure 23. Single Output Regulator with Key Performance Parameters Labeled

Heatsinks

A heatsink effectively increases the surface area of the package to improve the flow of heat away from the IC and into the surrounding air.

Each material in the heat flow path between the IC and the outside environment will have a thermal resistance. Like series electrical resistances, these resistances are summed to determine the value of $R_{\theta JA}$:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CS} + R_{\theta SA} \tag{3}$$

where

$R_{\theta JC}$ is the junction-to-case thermal resistance,

$R_{\theta CS}$ is the case-to-heatsink thermal resistance,

$R_{\theta SA}$ is the heatsink-to-ambient thermal resistance.

$R_{\theta JC}$ appears in the package section of the data sheet. Like $R_{\theta JA}$, it too is a function of package type. $R_{\theta CS}$ and $R_{\theta SA}$ are functions of the package type, heatsink and the interface between them. These values appear in heatsink data sheets of heatsink manufacturers.

Thermal, mounting, and heatsinking considerations are discussed in the ON Semiconductor application note AN1040/D.

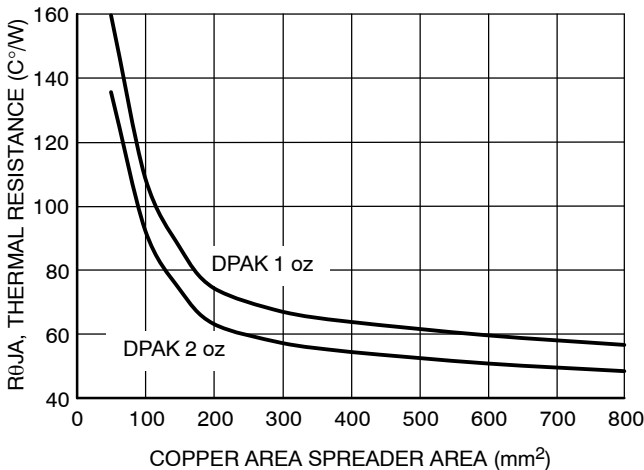


Figure 24. θ_{JA} vs. Copper Spreader Area, DPAK 5-Lead

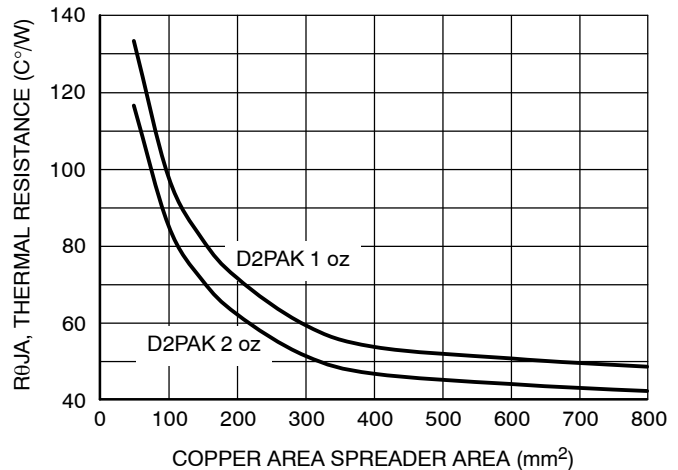


Figure 25. θ_{JA} vs. Copper Spreader Area, D²PAK 5-Lead

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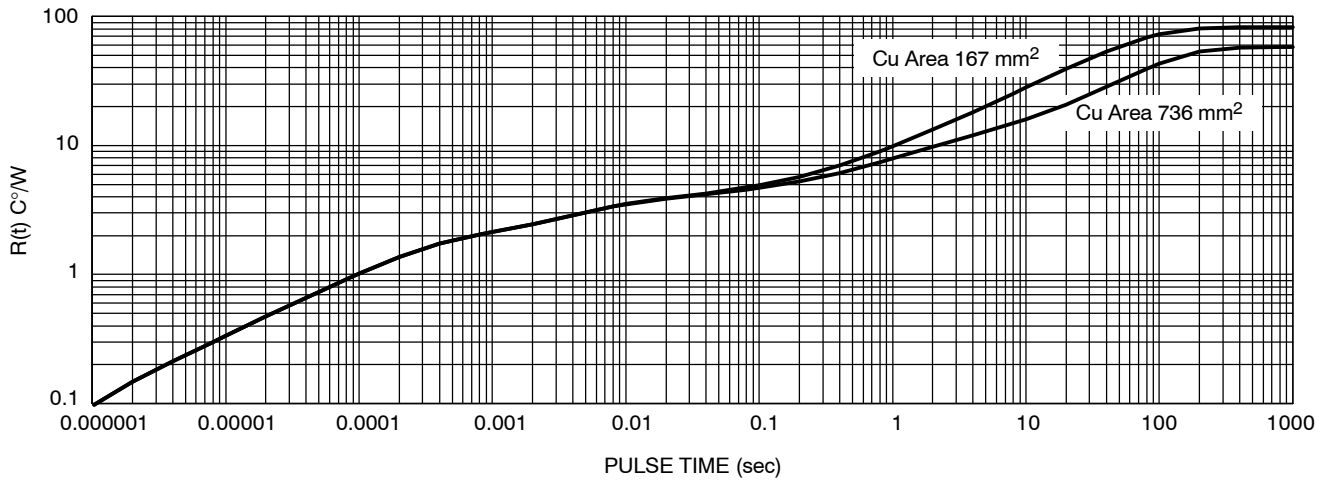


Figure 26. Single-Pulse Heating Curves, DPAK 5-Lead

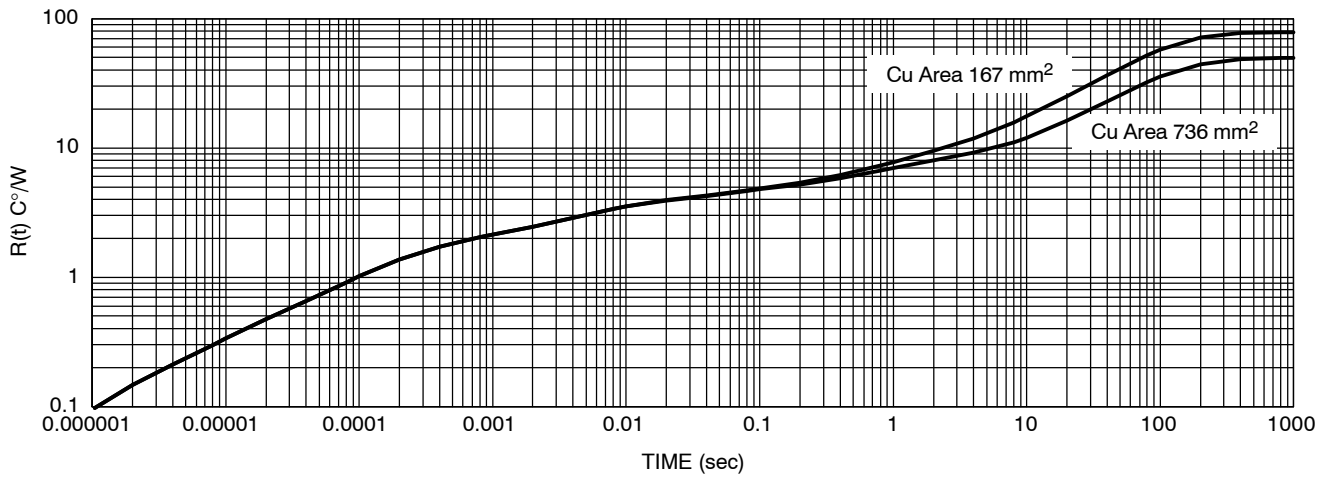


Figure 27. Single-Pulse Heating Curves, D²PAK 5-Lead

NCV4275C

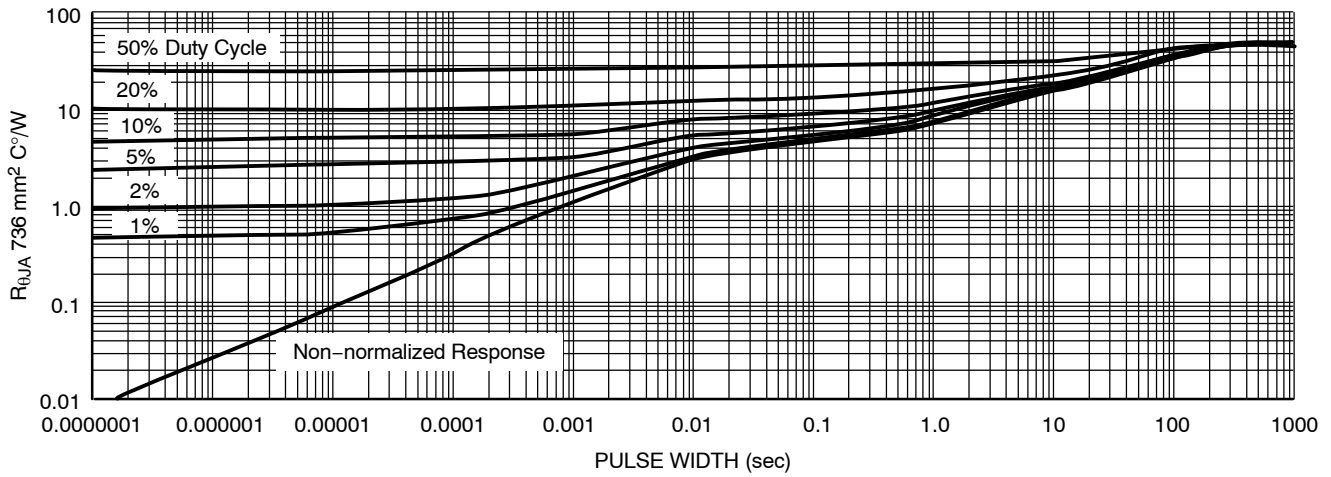


Figure 28. Duty Cycle for 1" Spreader Boards, DPAK 5-Lead

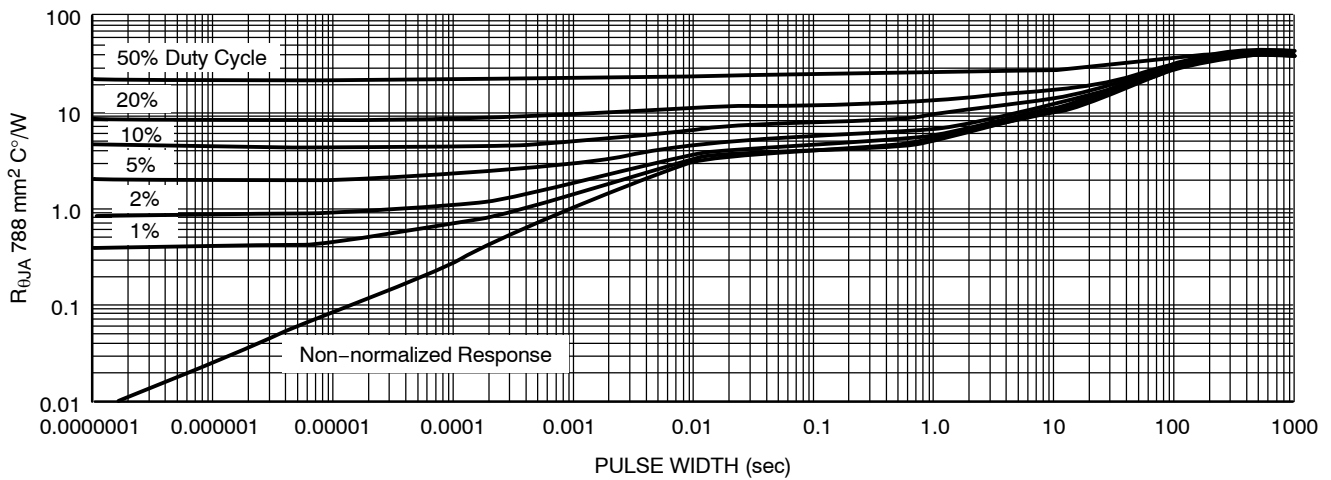


Figure 29. Duty Cycle for 1" Spreader Boards, D²PAK 5-Lead

ORDERING INFORMATION

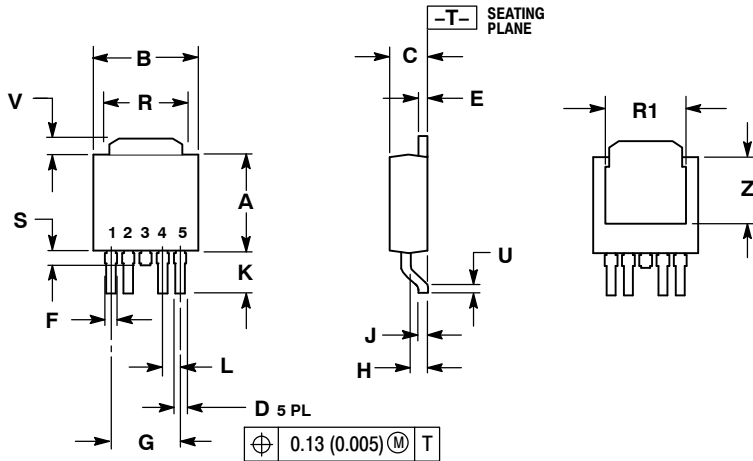
Device	Output Voltage	Package	Shipping [†]
NCV4275CDS50R4G	5.0 V	D2PAK (Pb-Free)	800 / Tape & Reel
NCV4275CDT50RKG		DPAK (Pb-Free)	2500 / Tape & Reel
NCV4275CDS33R4G	3.3 V	D2PAK (Pb-Free)	800 / Tape & Reel
NCV4275CDT33RKG		DPAK (Pb-Free)	2500 / Tape & Reel

[†]For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

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PACKAGE DIMENSIONS

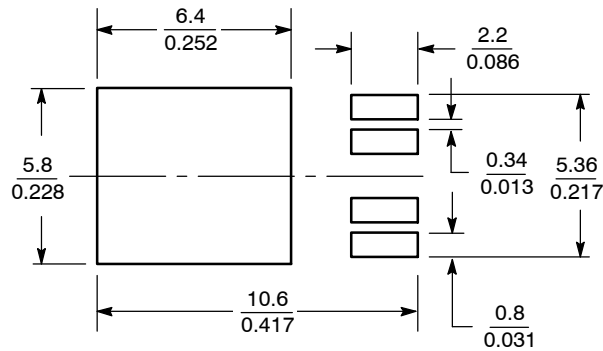
DPAK 5, CENTER LEAD CROP DT SUFFIX CASE 175AA ISSUE A



- NOTES:
 1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: INCH.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.235	0.245	5.97	6.22
B	0.250	0.265	6.35	6.73
C	0.086	0.094	2.19	2.38
D	0.020	0.028	0.51	0.71
E	0.018	0.023	0.46	0.58
F	0.024	0.032	0.61	0.81
G	0.180 BSC	4.56 BSC		
H	0.034	0.040	0.87	1.01
J	0.018	0.023	0.46	0.58
K	0.102	0.114	2.60	2.89
L	0.045 BSC	1.14 BSC		
R	0.170	0.190	4.32	4.83
R1	0.185	0.210	4.70	5.33
S	0.025	0.040	0.63	1.01
U	0.020	---	0.51	---
V	0.035	0.050	0.89	1.27
Z	0.155	0.170	3.93	4.32

SOLDERING FOOTPRINT*



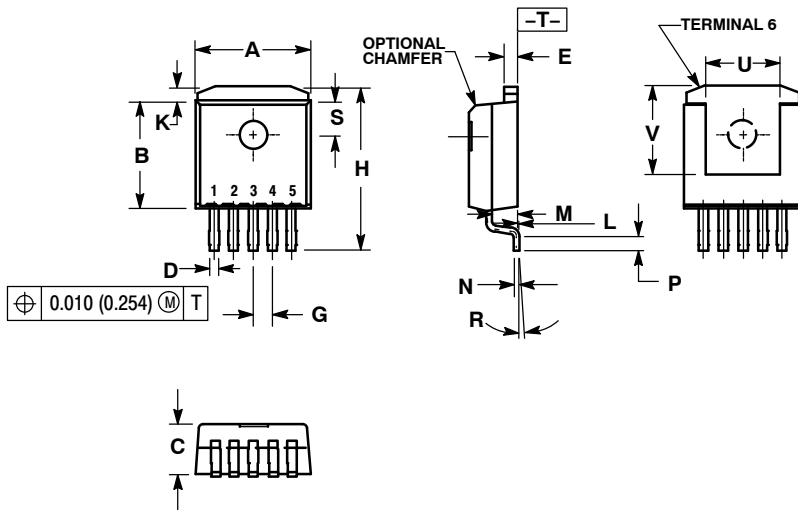
SCALE 4:1 $\left(\frac{\text{mm}}{\text{inches}}\right)$

*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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PACKAGE DIMENSIONS

**D²PAK, 5 LEAD
DS SUFFIX
CASE 936A-02
ISSUE C**

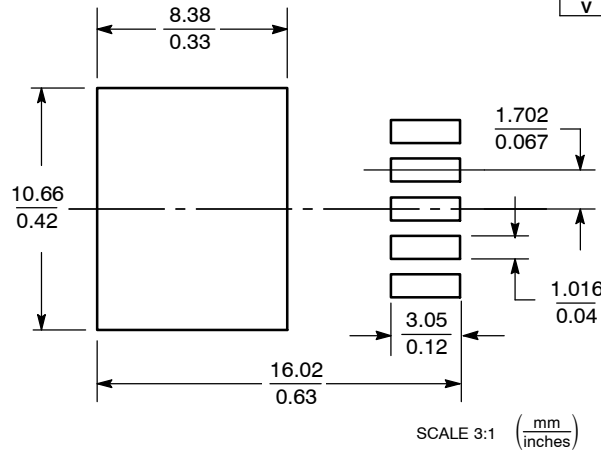


NOTES:

1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.
3. TAB CONTOUR OPTIONAL WITHIN DIMENSIONS A AND K.
4. DIMENSIONS U AND V ESTABLISH A MINIMUM MOUNTING SURFACE FOR TERMINAL 6.
5. DIMENSIONS A AND B DO NOT INCLUDE MOLD FLASH OR GATE PROTRUSIONS. MOLD FLASH AND GATE PROTRUSIONS NOT TO EXCEED 0.025 (0.635) MAXIMUM.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.386	0.403	9.804	10.236
B	0.356	0.368	9.042	9.347
C	0.170	0.180	4.318	4.572
D	0.026	0.036	0.660	0.914
E	0.045	0.055	1.143	1.397
G	0.067 BSC		1.702 BSC	
H	0.539	0.579	13.691	14.707
K	0.050 REF		1.270 REF	
L	0.000	0.010	0.000	0.254
M	0.088	0.102	2.235	2.591
N	0.018	0.026	0.457	0.660
P	0.058	0.078	1.473	1.981
R	5° REF		5° REF	
S	0.116 REF		2.946 REF	
U	0.200 MIN		5.080 MIN	
V	0.250 MIN		6.350 MIN	

SOLDERING FOOTPRINT*



*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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