

## CLC5526 Digital Variable Gain Amplifier (DVGA)

Check for Samples: [CLC5526](#)

### FEATURES

- **350 MHz Bandwidth**
- **Differential Input and Output**
- **Gain Control: Parallel w/data Latching**
- **Supply Voltage: +5V**
- **Supply Current: 48 mA**

### KEY SPECIFICATIONS

- **Low Two Tone Intermod:**
  - **Distortion: -64 dBc @ 1 VPP, 150 MHz**
  - **24.5 dBm IP3, 150 MHz**
- **Low Noise: 2.5 nV/√Hz (Max Gain), 9.3 dB Noise Figure (Max Gain)**
- **Wide Gain Range: +30 dB to -12 dB**
- **Gain Step Size: 6 dB**

### APPLICATIONS

- **Cellular/PCS Base Stations**
- **IF Sampling Receivers**
- **Infrared/CCD Imaging**
- **Back-channel Modems**
- **Electro-optics**
- **Instrumentation**
- **Medical Imaging**
- **High Definition Video**

### DESCRIPTION

The CLC5526 is a high performance, digitally controlled, variable-gain amplifier (DVGA). It has been designed for use in a broad range of mixed signal and digital communication applications such as mobile radio, cellular base stations and back-channel modems where automatic-gain-control (AGC) is required to increase system dynamic range.

The CLC5526 has differential input and output, allowing large signal swings on a single 5V rail. The input impedance is 200Ω. The differential output impedance is 600Ω and is designed to drive a 1 kΩ differential load. The output amplifier has excellent intermodulation performance. The CLC5526 is designed to accept signals from RF elements and maintain a terminated impedance environment.

The CLC5526 maintains a 350 MHz bandwidth over its entire gain and attenuation range from +30 dB to -12 dB. Internal clamping ensures very fast overdrive recovery. Two tone intermodulation distortion is excellent: at 150 MHz, 1 V<sub>pp</sub> it is -64 dBc.

Input signals to the CLC5526 are scaled by an accurate, differential R-2R resistive ladder with an input impedance of 200Ω. A scaled version of the input is selected under digital control and passed to the internal amplifier. The input common mode level is set at 2.4V via a bandgap referenced bias generator which can be overridden by an external input.

Following the resistive ladder is a fixed, 30 dB gain amplifier. The output stage common mode voltage of the CLC5526 is set to 3V, by internal, positive supply connected resistors.

Digital control of the CLC5526 is accomplished by a 3-bit parallel gain control input and a data valid pin to latch the data. If the data is not latched, the DVGA is transparent to gain control updates. All digital inputs are TTL/CMOS compatible.

A shutdown input reduces the CLC5526 supply current to a few mA. During shutdown, the input termination is maintained and current attenuation settings are held.

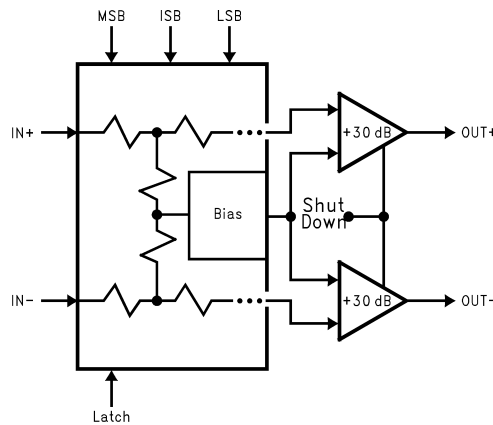
The CLC5526 operates over the industrial temperature range of -40°C to +85°C. The part is available in a 20-pin SSOP package.



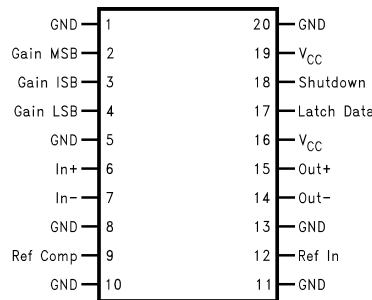
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### Block Diagram



### Pin Configuration



### Pin Descriptions

| Pin Name        | Pin No.                 | Description                             |
|-----------------|-------------------------|---|
| GND             | 1, 5, 8, 10, 11, 13, 20 | Circuit ground.                         |
| Gain MSB        | 2                       | Gain Selection Most Significant Bit     |
| Gain ISB        | 3                       | Gain Selection Data Bit                 |
| Gain LSB        | 4                       | Gain Selection Least Significant Bit    |
| In+             | 6                       | Positive Differential Input             |
| In-             | 7                       | Negative Differential Input             |
| Ref Comp        | 9                       | Reference Compensation                  |
| V <sub>CC</sub> | 16, 19                  | Positive Supply Voltage                 |
| Shutdown        | 18                      | Low Power Standby Control (Active High) |
| Latch Data      | 17                      | Data Latch Control (Active High)        |
| Out+            | 15                      | Positive Differential Output            |
| Out-            | 14                      | Negative Differential Output            |
| Ref In          | 12                      | External Reference Input                |



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

**Absolute Maximum Ratings<sup>(1)(2)</sup>**

|   |                     |
|---|---------------------|
| Positive Supply Voltage ( $V_{CC}$ )              | -0.5V to +6V        |
| Differential Voltage between any two Grounds      | <200 mV             |
| Analog Input Voltage Range                        | -0.5V to + $V_{CC}$ |
| Digital Input Voltage Range                       | -0.5V to + $V_{CC}$ |
| Output Short Circuit Duration (one-pin to ground) | Infinite            |
| Junction Temperature                              | 150°C               |
| Storage Temperature Range                         | -65°C to +150°C     |
| Lead Solder Duration (+300°C)                     | 10 sec              |

- (1) "Absolute Maximum Ratings" are limited values, to be applied individually, and beyond which the serviceability of the circuit may be impaired. Functional operability under any of these conditions is not necessarily implied. Exposure to maximum ratings for extended periods may affect device reliability.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.

**Recommended Operating Conditions**

|  |                |
|--|----------------|
| Positive Supply Voltage ( $V_{CC}$ )         | +5V $\pm$ 5%   |
| Differential Voltage between any two Grounds | <10 mV         |
| Analog Input Voltage Range, AC Coupled       | $\pm$ 0.5V     |
| Operating Temperature Range                  | -40°C to +85°C |

**Package Thermal Resistance**

| Package     | ( $\theta_{JA}$ ) | ( $\theta_{JC}$ ) |
|-------------|-------------------|-------------------|
| 20-Pin SSOP | 90°C/W            | 38°C/W            |

**Reliability Information**

|                  |     |
|------------------|-----|
| Transistor Count | 300 |
|------------------|-----|

**Electrical Characteristics**

The following specifications apply for  $V_{CC} = +5V$ ,  $R_L = 1\text{ k}\Omega$  maximum gain setting. **Boldface limits apply for  $T_A = T_{min} = -40^\circ\text{C}$  to  $T_{max} = +85^\circ\text{C}$** , all other limits  $T_A = 25^\circ\text{C}$ <sup>(1)(2)(3)</sup>.

| Symbol                      | Parameter               | Conditions                                 | Min | Typ | Max | Units |
|-----------------------------|-------------------------|--|-----|-----|-----|-------|
| <b>DYNAMIC PERFORMANCE</b>  |                         |  |     |     |     |       |
| BW                          | Small-Signal Bandwidth  |  |     | 350 |     | MHz   |
| <b>NOISE AND DISTORTION</b> |                         |  |     |     |     |       |
|                             | 2nd Harmonic Distortion | $f_{IN} = 150\text{ MHz}, 1\text{ V}_{PP}$ | 53  | 67  |     | dBc   |
|                             |                         | $f_{IN} = 250\text{ MHz}, 1\text{ V}_{PP}$ |     | 64  |     | dBc   |
|                             |                         | $f_{IN} = 150\text{ MHz}, 2\text{ V}_{PP}$ | 43  | 62  |     | dBc   |
|                             |                         | $f_{IN} = 250\text{ MHz}, 2\text{ V}_{PP}$ |     | 58  |     | dBc   |
|                             | 3rd Harmonic Distortion | $f_{IN} = 150\text{ MHz}, 1\text{ V}_{PP}$ | 53  | 71  |     | dBc   |
|                             |                         | $f_{IN} = 250\text{ MHz}, 1\text{ V}_{PP}$ |     | 70  |     | dBc   |
|                             |                         | $f_{IN} = 150\text{ MHz}, 2\text{ V}_{PP}$ | 43  | 57  |     | dBc   |
|                             |                         | $f_{IN} = 250\text{ MHz}, 2\text{ V}_{PP}$ |     | 56  |     | dBc   |

- (1) Limits are 100% tested at 25°C.
- (2) Typical specifications are the mean values of the distributions of deliverable amplifiers tested to date.
- (3) Outgoing quality levels are determined from tested parameters.

## Electrical Characteristics (continued)

The following specifications apply for  $V_{CC} = +5V$ ,  $R_L = 1\text{ k}\Omega$  maximum gain setting. **Boldface limits apply for  $T_A = T_{min} = -40^\circ\text{C}$  to  $T_{max} = +85^\circ\text{C}$** , all other limits  $T_A = 25^\circ\text{C}^{(1)(2)(3)}$ .

| Symbol                       | Parameter                           | Conditions  | Min        | Typ      | Max        | Units                  |
|------------------------------|-------------------------------------|---|------------|----------|------------|------------------------|
| IMD                          | Two Tone Intermodulation Distortion | $f_1 = 149.9\text{ MHz}$ , $f_2 = 150.1\text{ MHz}$ , $1\text{ V}_{PP}$ Composite |            | 64       |            | dBc                    |
|                              |                                     | $f_1 = 149.9\text{ MHz}$ , $f_2 = 150.1\text{ MHz}$ , $2\text{ V}_{PP}$ Composite |            | 61       |            | dBc                    |
|                              |                                     | $f_1 = 249.9\text{ MHz}$ , $f_2 = 250.1\text{ MHz}$ , $1\text{ V}_{PP}$ Composite |            | 63       |            | dBc                    |
|                              |                                     | $f_1 = 249.9\text{ MHz}$ , $f_2 = 250.1\text{ MHz}$ , $2\text{ V}_{PP}$ Composite |            | 54       |            | dBc                    |
|                              | Two Tone, 3rd Order Intermodulation | 150 MHz   |            | 24.5     |            | dBm                    |
| Thermal Noise                |                                     | Minimum Gain Setting  |            | 2.2      |            | nV/ $\sqrt{\text{Hz}}$ |
|                              |                                     | Maximum Gain Setting  |            | 2.5      |            | nV/ $\sqrt{\text{Hz}}$ |
| Noise Figure                 |                                     | Maximum Gain Setting  |            | 9.3      |            | dB                     |
| <b>ANALOG I/O</b>            |                                     |   |            |          |            |                        |
|                              | Differential Input Impedance        |   |            | 200      |            | $\Omega$               |
|                              | Differential Output Impedance       |   |            | 600      |            | $\Omega$               |
|                              | Input Signal Level (AC Coupled)     | Maximum Gain  |            | 126      |            | mV                     |
|                              | Maximum Input Signal Level          | Recommended   |            | 6        |            | $V_{PP}$               |
|                              | Maximum Output Signal Level         | Recommended   |            | 4        |            | $V_{PP}$               |
|                              | Output Clipping                     |   |            | 8        |            | $V_{PP}$               |
| <b>GAIN PARAMETERS</b>       |                                     |   |            |          |            |                        |
|                              | Maximum Gain                        |   |            | 30       |            | dB                     |
|                              | Minimum Gain                        |   |            | -12      |            | dB                     |
|                              | Gain Step Size                      |   |            | 6.02     |            | dB                     |
|                              | Gain Step Accuracy                  | (1 sigma)   |            | 0.03     |            | dB                     |
|                              | Cumulative Gain Step Error          | (1 sigma)   |            | 0.085    |            | dB                     |
| <b>DIGITAL INPUTS/TIMING</b> |                                     |   |            |          |            |                        |
|                              | Logic Compatibility                 |   |            | TTL/CMOS |            | V                      |
| $V_{IL}$                     | Logic Input Low Voltage             |   |            |          | <b>0.8</b> | V                      |
| $V_{IH}$                     | Logic Input High Voltage            |   | <b>2.0</b> |          |            | V                      |
| $T_{SU}$                     | Setup Time                          |   |            | 3        |            | ns                     |
| $T_{HOLD}$                   | Hold Time                           |   |            | 3        |            | ns                     |
| $T_{PW}$                     | Minimum Pulse Width                 |   |            | 3        |            | ns                     |
| <b>POWER REQUIREMENTS</b>    |                                     |   |            |          |            |                        |
| $I_{CC}$                     | +5V Supply Current                  |   |            | 48       | 60         | mA                     |
|                              | Shutdown                            |   |            | 9        |            | mA                     |

### Typical Performance Characteristics

( $V_{CC} = +5V$ ,  $R_L = 1\text{ k}\Omega$ , max gain; unless specified)

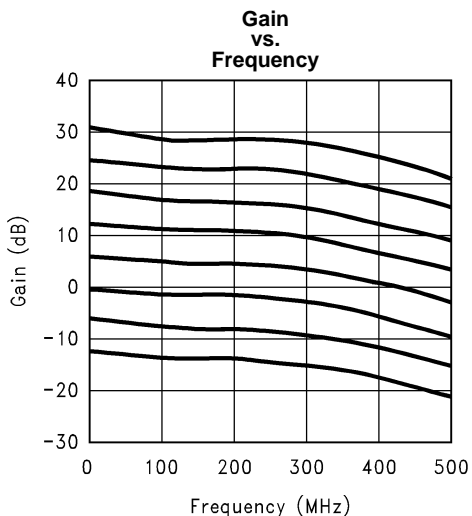


Figure 1.

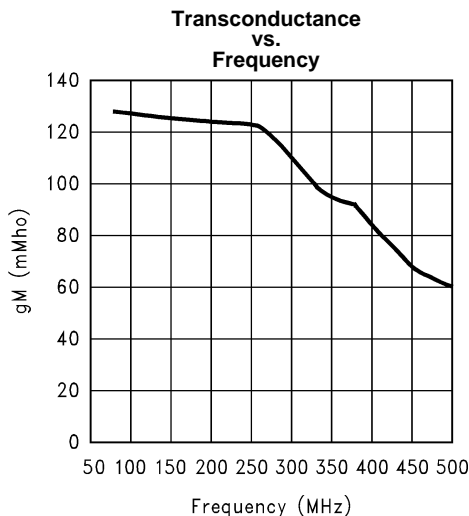


Figure 2.

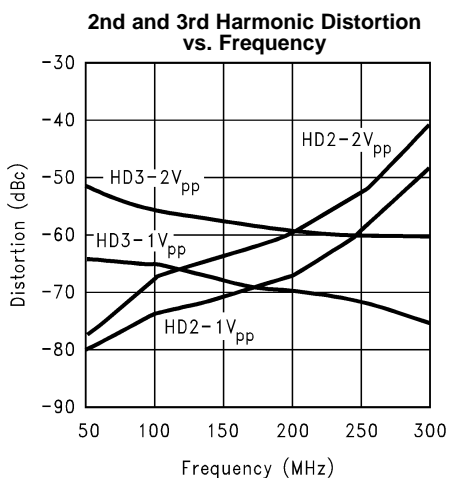


Figure 3.

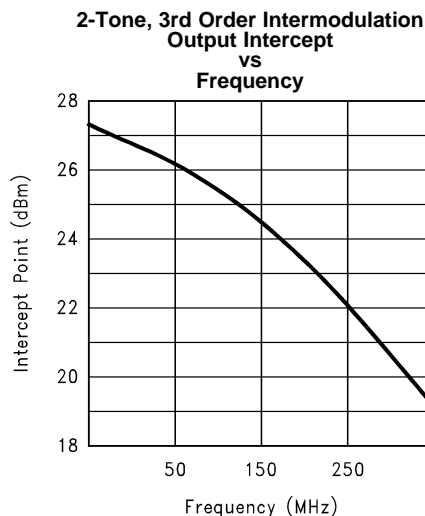


Figure 4.

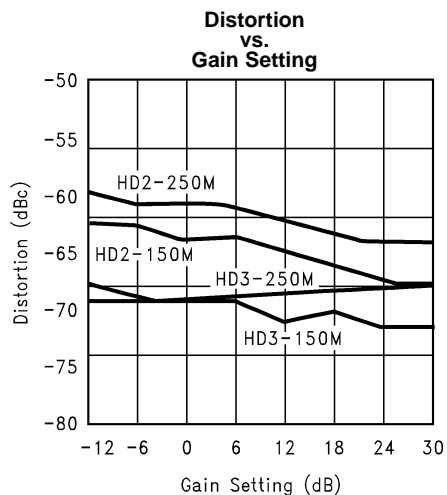


Figure 5.

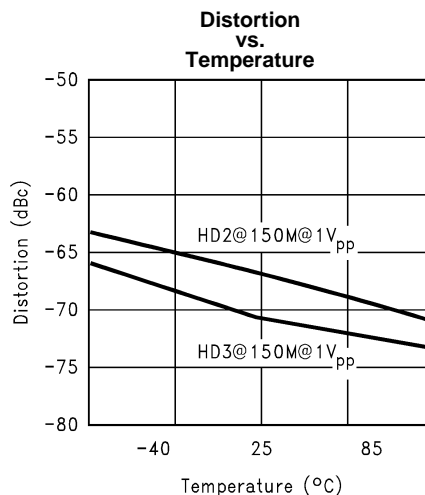


Figure 6.

**Typical Performance Characteristics (continued)**

( $V_{CC} = +5V$ ,  $R_L = 1\text{ k}\Omega$ , max gain; unless specified)

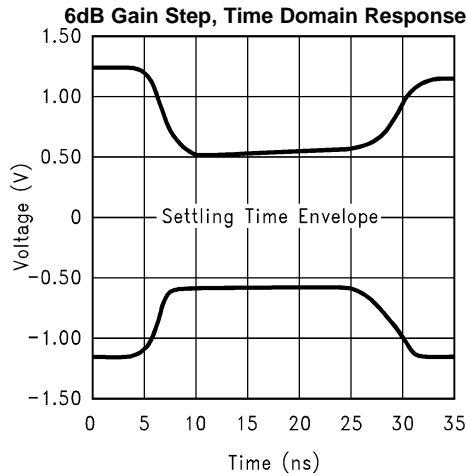


Figure 7.

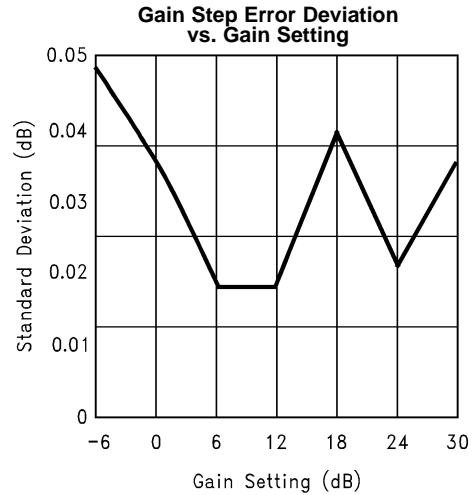


Figure 8.

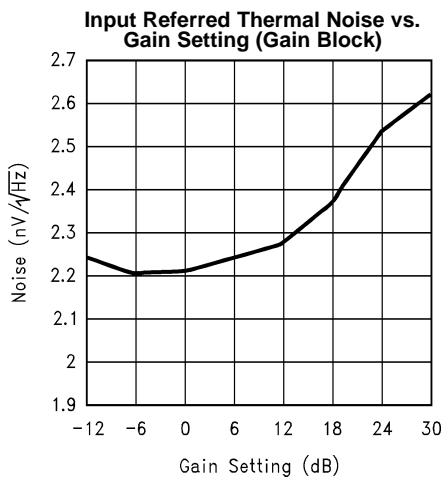


Figure 9.

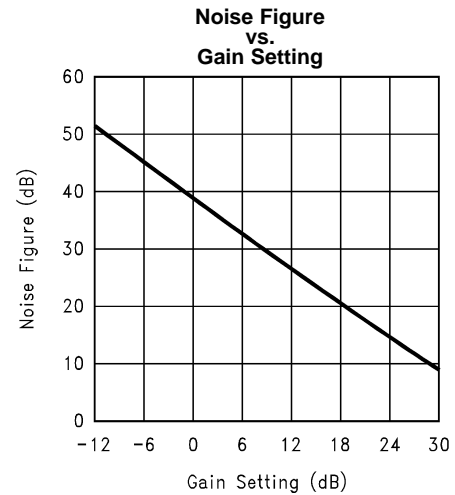


Figure 10.

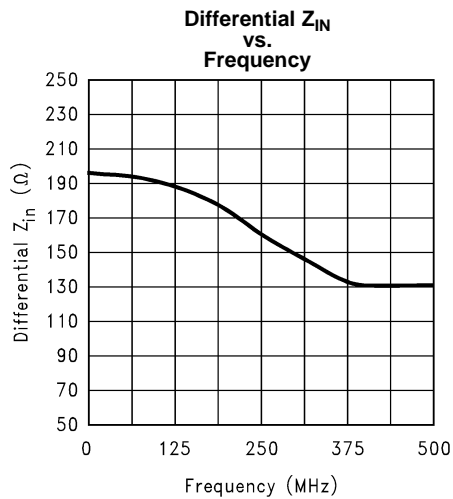


Figure 11.

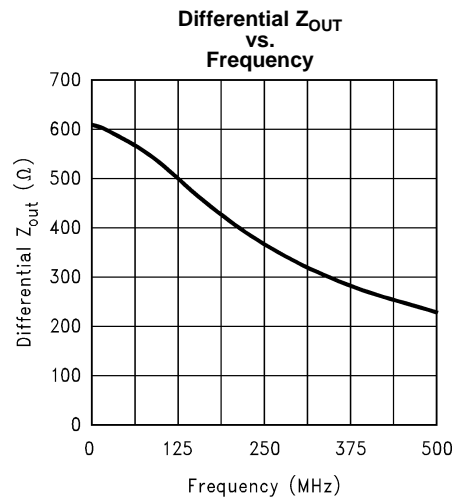
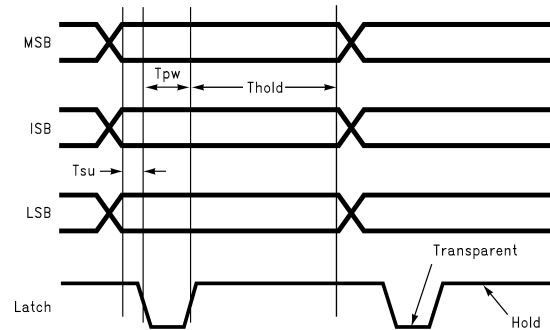


Figure 12.

**Timing Diagram**



**Truth Table**

| Gain Word | MSB | ISB | LSB | Gain (dB) |
|-----------|-----|-----|-----|-----------|
| 0         | 0   | 0   | 0   | -12       |
| 1         | 0   | 0   | 1   | -6        |
| 2         | 0   | 1   | 0   | 0         |
| 3         | 0   | 1   | 1   | +6        |
| 4         | 1   | 0   | 0   | +12       |
| 5         | 1   | 0   | 1   | +18       |
| 6         | 1   | 1   | 0   | +24       |
| 7         | 1   | 1   | 1   | +30       |

## APPLICATIONS

### DESCRIPTION

The CLC5526 is a digitally programmable, variable gain amplifier with the following features:

- 8 gain settings ranging from –12 to +30 dB in 6dB steps
- Differential inputs and outputs (externally AC coupled)
- Self biased input common-mode voltage
- 3-bit parallel digital control
- Single +5V supply
- Low-Power standby mode

Please refer to [Figure 13](#) for a representative block diagram.

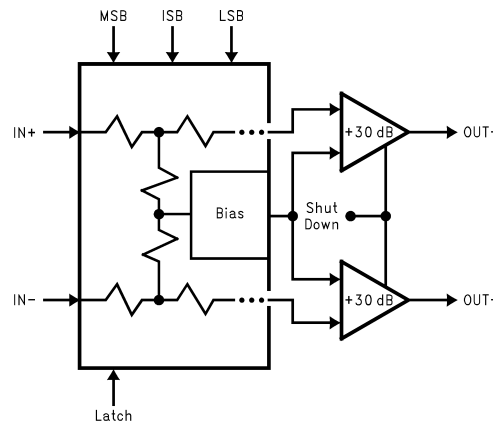


Figure 13. CLC5526 Block Diagram

### GAIN SELECTION

Gain levels can be decreased from the maximum value in –6dB steps via the 3-bit digital inputs. [Table 1](#) shows the gain selection truth table for a 1000Ω differential load.

Table 1. Gain Selection Truth Table

| Gain Word | MSB | ISB | LSB | Gain (dB) |
|-----------|-----|-----|-----|-----------|
| 0         | 0   | 0   | 0   | –12       |
| 1         | 0   | 0   | 1   | –6        |
| 2         | 0   | 1   | 0   | 0         |
| 3         | 0   | 1   | 1   | +6        |
| 4         | 1   | 0   | 0   | +12       |
| 5         | 1   | 0   | 1   | +18       |
| 6         | 1   | 1   | 0   | +24       |
| 7         | 1   | 1   | 1   | +30       |

Gain settings can be calculated as follows:

$$\text{GAIN} = -12 \text{ dB} + (\text{Gain Word}) * 6.02 \text{ dB} \quad (1)$$

Gain selection has two modes: Transparent or latched, depending on the LATCH input. If the LATCH input is held LOW, then the device is in the transparent mode. Changes on data inputs will result in direct changes to the gain setting.

Input data will be latched upon the LOW to HIGH transition of LATCH. While LATCH is HIGH, digital data will be ignored until LATCH is strobed low again.



**Note:** Upon power-up the analog inputs are disconnected from the internal amplifier. LATCH will need to be strobed LOW before an analog output will be present!

### DIFFERENTIAL I/O CONSIDERATIONS

Analog inputs and outputs need to be AC coupled to prevent DC loading of the common-mode voltages. If driving the CLC5526 from a single-ended 50Ω source is required, a 1:2 transformer should be used to generate the differential inputs. As the differential input impedance of the CLC5526 is 200Ω, the 1:4 impedance ratio will allow for optimum matching to the 50Ω source. The secondary outputs of the transformer should be AC coupled to the CLC5526 analog inputs, while the secondary center tap of the transformer should be directly connected to the system ground.

The CLC5526 is designed to drive differential circuits, such as the CLC5957 Analog to Digital convertor. Figure 14 below shows a typical application of the CLC5526.

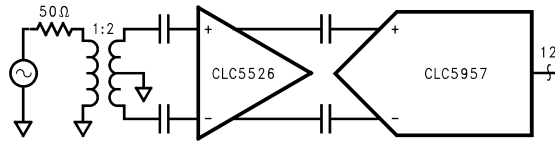


Figure 14. Differential I/O Connections

### DRIVING LOADS

Actual gain of the CLC5526 will vary with the output load. The device is designed to provide +30 dB maximum gain with a 1000Ω differential load.

Each output of the CLC5526 contains an internal 300Ω resistor to the V<sub>CC</sub> rail. Actual gain calculations need to take this in account with a given external load resistor. The effective load resistance can be used with the following equation to calculate max gain values.

$$A_v = 20 \log (0.0843 * R_{\text{leff}})$$

where

- $R_{\text{leff}} = R_{\text{int}} \parallel R_{\text{ext(diff)}}$
- $R_{\text{int}} = 600\Omega$  differential (2)

Figure 15 shows maximum gain values over output load. Resistor values are for differential loads.

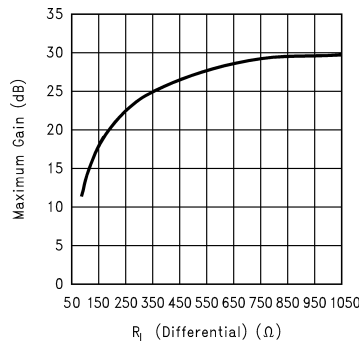


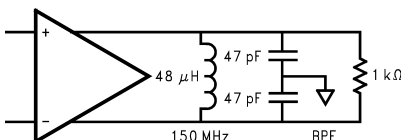
Figure 15. Maximum Gain vs R<sub>LOAD</sub>

Stray capacitance at the output, along with the output load value will form a pole, which can degrade the CLC5526 bandwidth. For a narrow-band application this problem can be alleviated by using a tuned load, which will incorporate any stray parasitic impedance into a resonant circuit. By tuning the resonant load, full gain can be achieved with a given resistive load.

A typical tuned load is shown below in Figure 16, where the resonant frequency is tuned about 150 MHz.

The 1000Ω load in this circuit can represent the input impedance of the CLC5957 Analog to Digital converter. Actual values for the reactive components may vary slightly to account for board and device parasitic elements.

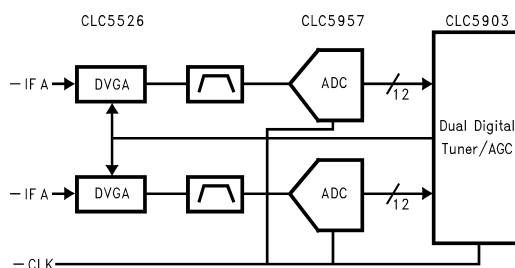
The Diversity Receiver Chipset may also use the ADC12L066 A/D converter in place of the CLC5957. Please refer to the Low Power Diversity Receiver Chipset (LDRCS) User's Guide for input matching between the CLC5526 and ADC12L066.



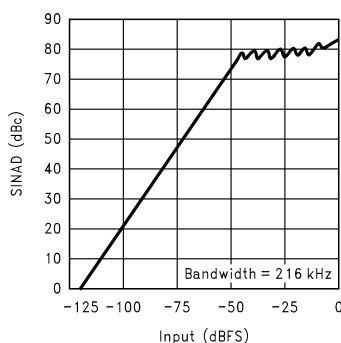
**Figure 16. CLC5526 Driving a Tuned Load**

## Typical Application

Although the CLC5526 can be used as a general purpose digital variable gain amplifier, it was specifically designed to provide the variable gain function in TI's Diversity Receiver Chipset. In this application, the CLC5526 drives a tuned BPF and the CLC5957 Analog to Digital converter. Digitized IF data is downsampled and tuned with the CLC5903 dual digital tuner which also provides the AGC control function. AGC data is fed back to the CLC5526. The CLC5957 differential input impedance is 1000Ω, so with the tuned load, full gain of the CLC5526 is achieved. Figure 17 shows the block diagram of the Diversity Receiver Chipset application. Figure 18 shows the SINAD vs Input Power of the diversity receiver chipset. For input power levels ranging from 0 dB to -110 dB, the chip set provides a signal to noise ratio in excess of the 9 dB required for a typical GSM system.



**Figure 17. Diversity Receiver Chipset Block Diagram**



**Figure 18. Diversity Receiver Chipset  
SINAD vs Input Power**

## Layout Considerations

A proper printed circuit layout is essential for achieving high frequency performance. Texas Instruments provides evaluation boards for the CLC5526, which include input and output transformers for impedance matching and single to differential signal conversion.

Supply bypassing is required for best performance. Provide a 6.8 μF Tantalum and 0.1 μF ceramic capacitor as close as possible to the supply pin.

In addition, a 100 pF ceramic capacitor should be placed between the COMP pin (pin 9) and the system ground. This will filter high frequency noise from the common-mode level.

Ceramic coupling capacitors should be used to AC couple both the input and output. Actual values will depend upon the signal frequency.

### Evaluation Board Layout and Schematic Diagram

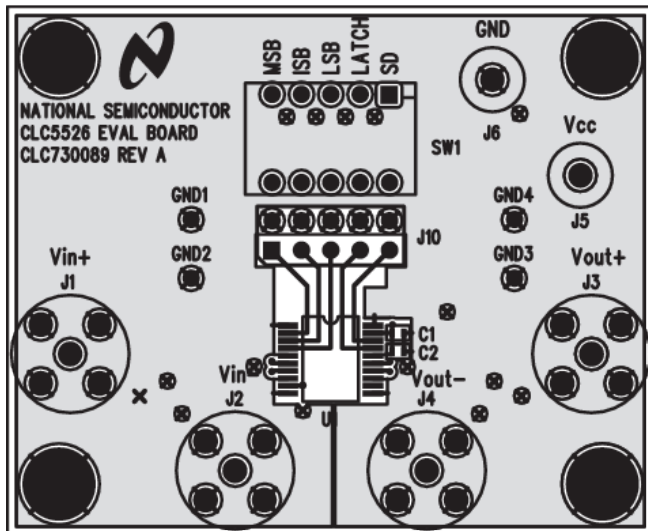


Figure 19. CLC5526 Layer 1

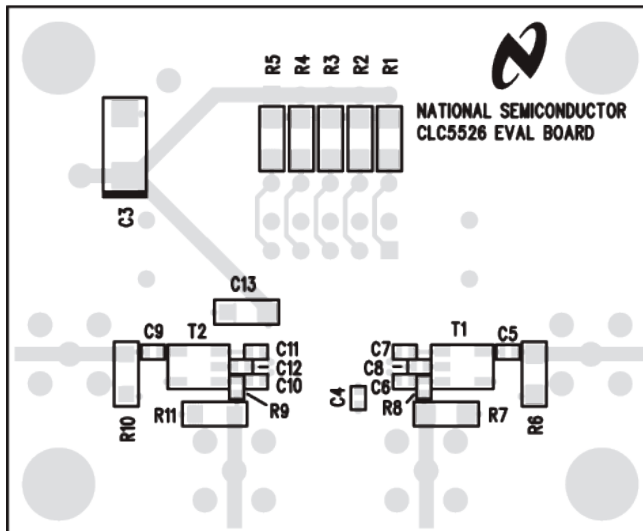


Figure 20. CLC5526 Layer 2

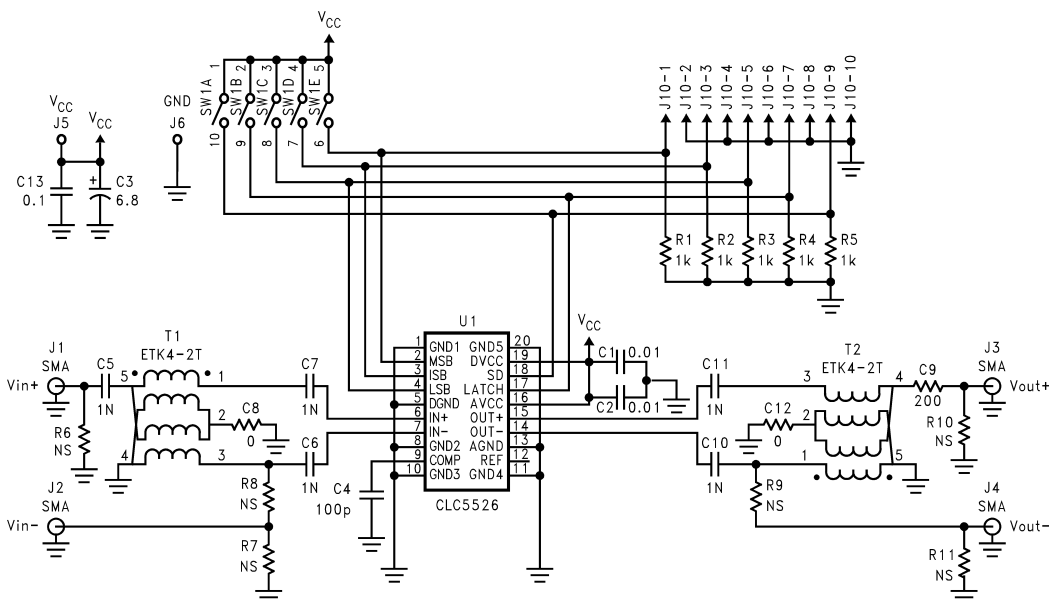


Figure 21. Evaluation Board Schematic

## REVISION HISTORY

| Changes from Revision D (March 2013) to Revision E         | Page               |
|--|--------------------|
| • Changed layout of National Data Sheet to TI format ..... | <a href="#">11</a> |

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