

IFX24401

Low Dropout Voltage Regulator

IFX24401TEV50 IFX24401ELV50

Data Sheet

Rev. 1.02, 2009-12-10

Standard Power



Low Dropout Voltage Regulator

IFX24401



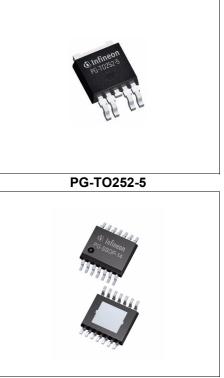
1 Overview

Features

- Output voltage 5 V ±2%
- Ultra low current consumption: 20 μA (typ.)
- 300 mA current capability
- Enable input
- Very low-drop voltage
- · Short circuit protection
- Overtemperature protection
- Low Dropout Voltage, 250mV (typ.)
- High Input Voltage 45 V
- Temperature Range -40 °C ≤ T_i ≤ 125 °C
- Green Product (RoHS compliant)

Applications

- Battery powered devices (e.g. Handheld GPS)
- Portable Radios
- HDTV Televisions
- Game Consoles
- Network Routers



PG-SSOP-14

For automotive and transportation applications, please refer to the Infineon TLE and TLF voltage regulator series.

Functional Description

The IFX24401 is a monolithic integrated low-drop voltage regulator for load currents up to 300 mA. The output voltage is regulated to $V_{Q,nom}$ = 5.0 V with an accuracy of ±2%. A sophisticated design allows stable operation with low ESR ceramic output capacitors down to 470 nF. The device is designed for the harsh environments. Therefore it is protected against overload, short circuit and overtemperature conditions. Due to its ultra low stand-by current consumption of 20 µA (typ.) the IFX24401 is ideal for use in battery powered applications. The regulator can be shut down via an Enable input which further reduces the current consumption to 5 µA (typ.). An integrated output sink current circuitry keeps the voltage at the Output pin Q below 5.5 V even when reverse currents are applied. Thus connected devices are protected from overvoltage damage.

Туре	Package	Marking
IFX24401TEV50	PG-TO252-5	2440150
IFX24401ELV50	PG-SSOP-14	24401V50



Block Diagram

2 Block Diagram

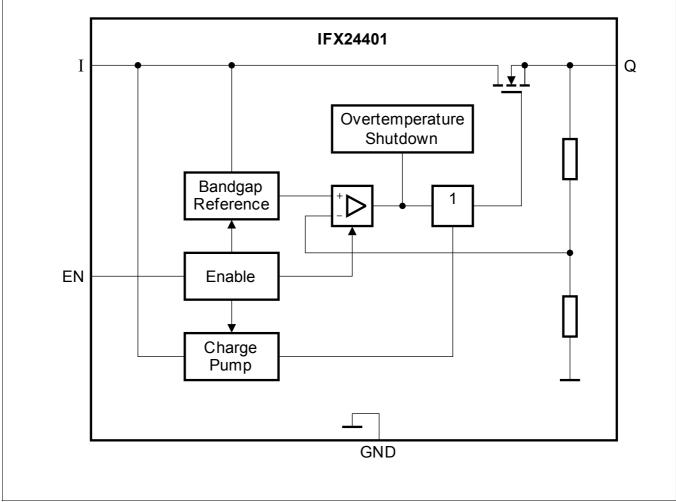
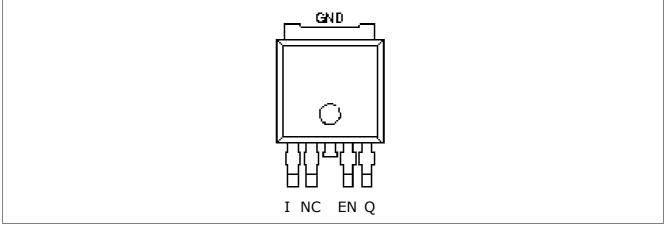


Figure 1 Block Diagram



Pin Configuration

3 Pin Configuration





3.1 Pin Definitions and Functions (PG-TO252-5)

Pin	Symbol	Function
1	1	Input
		Connect ceramic capcitor between I and GND
2	N.C.	No Connect
		May be open or connected to GND
3	GND	Ground
		Internally connected to heat slug
4	EN	Enable Input
		Low signal level disables the regulator. Pull-down resistor is integrated.
5	Q	Output
		Place capacitor between Q pin and GND. Capacitor placement should be close to pin.
		Refer to capacitance and ESR requirements in "Functional Range" on Page 6
Heat Slug		Heat Slug
		Connect to board GND and heatsink



IFX24401

Pin Configuration

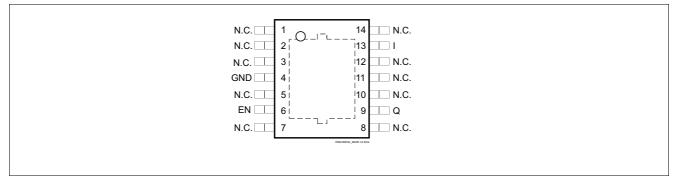


Figure 3 Pin Configuration PG-SSOP-14 (top view)

3.2 Pin Definitions and Functions (PG-SSOP-14)

Pin	Symbol	Function
1,2,3,5,7	N.C.	No Connect
		May be open or connected to GND
4	GND	Ground
6	EN	Enable Input
		Low signal level disables the regulator. Pull-down resistor is integrated.
8,10,11,1	N.C.	No Connect
2,14		May be open or connected to GND
9	Q	Output
		Place capacitor between Q pin and GND. Capacitor placement should be close to pin. Refer to capacitance and ESR requirements in "Functional Range" on Page 6
13	1	Input
		Connect ceramic capcitor between I and GND
Pad		Exposed Pad
		Connect to board GND and heatsink



4 General Product Characteristics

4.1 Absolute Maximum Ratings

Absolute Maximum Ratings¹⁾

 $T_{\rm j}$ = -40 °C to 150 °C; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

Parameter	Symbol	Limit Values		Unit	Test Condition
		Min.	Max.	_	
Input I	I				
Voltage	VI	-0.3	45	V	-
Current	I	-1	-	mA	-
Enable EN	I				
Voltage	V _{EN}	-0.3	45	V	Observe current limit $I_{\text{EN,max}}^{2)}$
Current	I _{EN}	-1	1	mA	-
Output Q		1	I	I	
Voltage	V _Q	-0.3	5.5	V	-
Voltage	VQ	-0.3	6.2	V	$t < 10 \text{ s}^{3)}$
Current	I _Q	-1	-	mA	-
Temperature		4			
Junction temperature	Tj	-40	150	°C	-
Storage temperature	T _{stg}	-50	150	°C	-
		1	1		

1) Not subject to production test, specified by design.

2) External resistor required to keep current below absolute maximum rating when voltages \geq 5.5 V are applied.

3) Exposure to these absolute maximum ratings for extended periods (t > 10 s) may affect device reliability.

Note: Stresses above the ones listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note: Integrated protection functions are designed to prevent IC destruction under fault conditions described in the data sheet. Fault conditions are considered as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.

4.2 Functional Range

Parameter	Symbol	Lim	it Values	Unit	Remarks
		Min.	Max.		
Input voltage	V	5.5	42	V	-
Junction temperature	Ti	-40	125	°C	-
Output Capacitor	C _Q	470	_	nF	1)
	ESR (C_Q)	-	10	Ω	f = 10 kHz

1) The minimum output capacitance requirement is applicable for a worst case capacitor tolerance of 30%

Note: In the operating range, the functions given in the circuit description are fulfilled.



General Product Characteristics

4.3 Thermal Resistance

Pos.	Parameter	Symbol	Limit Value			Unit	Conditions
			Min.	Тур.	Max.	1	
IFX244	401TEV50 (PG-TO252-5,)	I					
4.3.1	Junction to Case ¹⁾	R _{thJC}	-	4	-	K/W	measured to pin 5
4.3.2	Junction to Ambient ¹⁾	R_{thJA}	_	115	_	K/W	Footprint only ²⁾
4.3.3			-	57	-	K/W	300mm^2 heatsink area on PCB ²⁾
4.3.4	_		-	42	-	K/W	600mm ² heatsink area on PCB ²⁾
IFX244	401ELV50 (PG-SSOP-14)						
4.3.5	Junction to Case ¹⁾	$R_{ m thJC}$	-	7	_	K/W	measured to pin 5
4.3.6	Junction to Ambient ¹⁾	R_{thJA}	_	120	_	K/W	Footprint only ²⁾
4.3.7			-	59	-	K/W	300mm^2 heatsink area on PCB ²⁾
4.3.8			-	49	-	K/W	600mm^2 heatsink area on $\text{PCB}^{2)}$

1) not subject to production test, specified by design

2) EIA/JESD 52_2, FR4, 80 \times 80 \times 1.5 mm; 35 μ Cu, 5 μ Sn



General Product Characteristics

Table 1 Electrical Characteristics

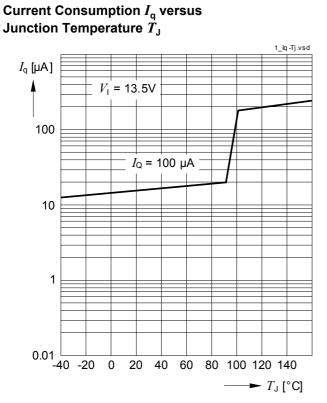
 $V_{\rm I}$ = 13.5 V; $V_{\rm EN}$ = 5 V; -40 °C < $T_{\rm i}$ < 125 °C (unless otherwise specified)

Parameter	Symbol		Limit Values			Measuring Condition
		Min.	Тур.	Max.		
Output Q	L.		1		1	
Output voltage	V _Q	4.9	5.0	5.1	V	0.1 mA < $I_{\rm Q}$ < 300 mA 6 V < $V_{\rm I}$ < 16 V
Output voltage	V _Q	4.9	5.0	5.1	V	0.1 mA < $I_{\rm Q}$ < 100 mA 6 V < $V_{\rm I}$ < 40 V
Output current limit	$I_{\rm Q,LIM}$	320	-	-	mA	1)
Output current limit	$I_{Q,LIM}$	-	_	800	mA	$V_{\rm Q} = 0 \rm V$
Current consumption; $I_q = I_1 - I_Q$	Iq	-	20	30	μΑ	I _Q = 0.1 mA; T _j = 25 °C
Current consumption; $I_q = I_1 - I_Q$	Iq	-	-	40	μA	$I_{\rm Q}$ = 0.1 mA; $T_{\rm j} \le 80 \ ^{\circ}{\rm C}$
Quiescent current; Disabled	Iq	-	5	9	μA	V _{EN} = 0 ∨; T _j < 80 °C
Drop voltage	V _{dr}	-	250	500	mV	$I_{\rm Q}$ = 200 mA; $V_{\rm dr}$ = $V_{\rm I}$ - $V_{\rm Q}^{(1)}$
Load regulation	$\Delta V_{Q, lo}$	-40	15	40	mV	$I_{\rm Q}$ = 5 mA to 250 mA
Line regulation	$\Delta V_{Q, li}$	-20	5	20	mV	$V_{\rm I}$ = 10V to 32 V; $I_{\rm Q}$ = 5 mA
Power supply ripple rejection	PSRR	-	60	-	dB	f _r = 100 Hz; V _r = 0.5 Vpp
Temperature output voltage drift	dV_Q/dT	-	0.5	-	mV/K	-
Enable Input EN	I					
Turn-on Voltage	V _{EN ON}	3.1	-	-	V	$V_Q \ge 4.9 V$
Turn-off Voltage	$V_{EN OFF}$	-	-	0.8	V	$V_Q \leq 0.3 V$
H-input current	$I_{\rm EN \ ON}$	-	3	4	μA	V _{EN} = 5 V
L-input current	$I_{EN OFF}$	-	0.5	1	μΑ	V _{EN} = 0 V; T _i < 80 °C

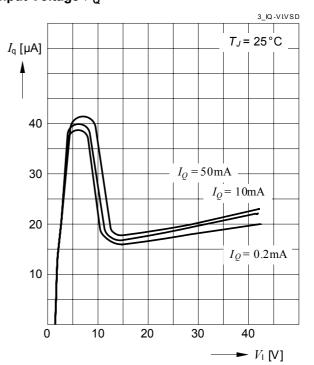
1) Measured when the output voltage $V_{\rm Q}$ has dropped 100 mV from the nominal value obtained at $V_{\rm I}$ = 13.5 V.



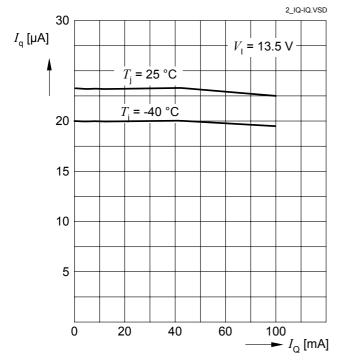
5 Typical Performance Characteristics



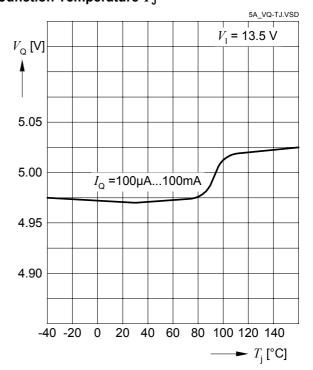
Current Consumption I_q versus Input Voltage V_q



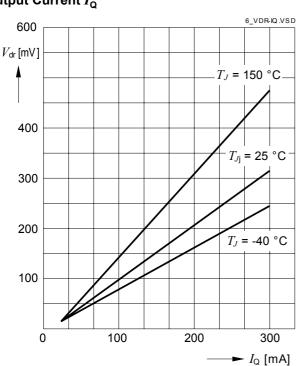
Current Consumption I_q versus Output Current I_Q



Output Voltage V_{Q} versus Junction Temperature T_{J}

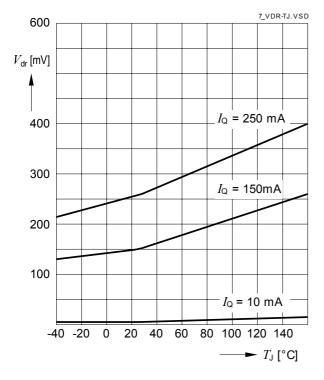




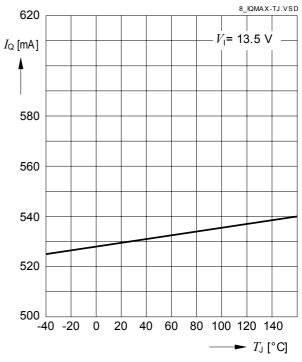


Dropout Voltage V_{dr} versus Output Current I_Q

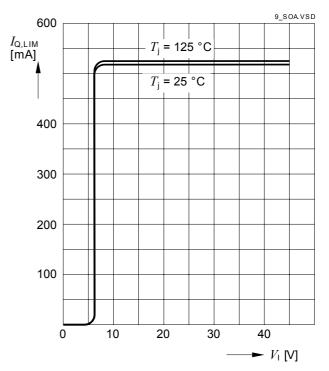
Dropout Voltage $V_{\rm dr}$ versus Junction Temperature



Maximum Output Current I_{Q} versus Junction Temperature T_{j}



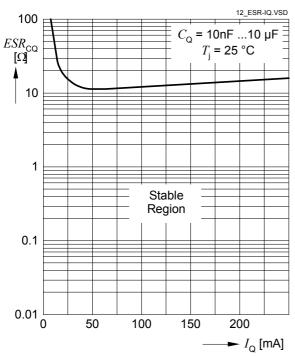
Maximum Output Current I_{Q} versus Input Voltage V_{I}



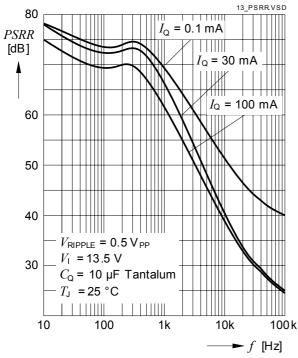


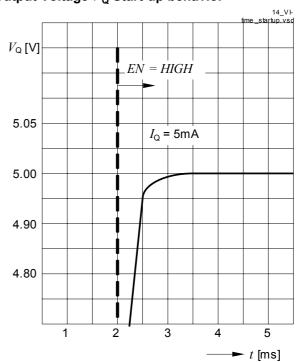


Region of Stability



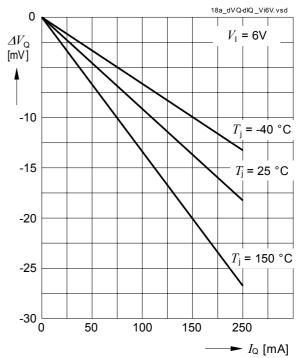
Power Supply Ripple Rejection PSRR versus Frequency f



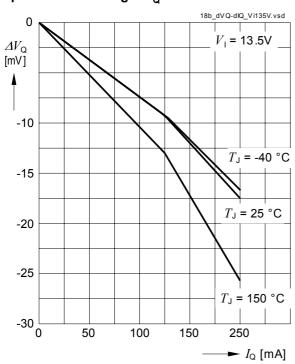


Output Voltage $V_{\rm Q}$ Start-up behavior



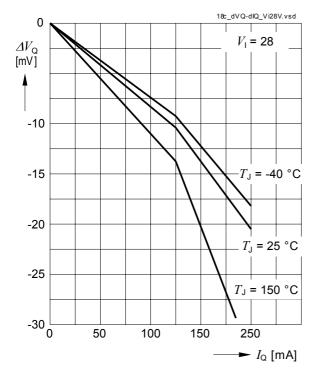




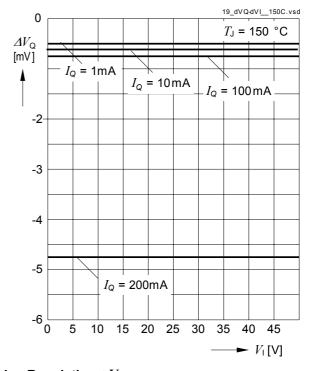


Load Regulation ΔV_{Q} versus Output Current Change dI_{Q}

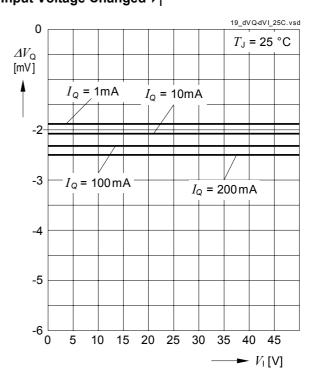
Load Regulation ΔV_{Q} versus Output Current Change ΔI_{Q}



Line Regulation ΔV_{Q} versus Input Voltage Changed V_{I}

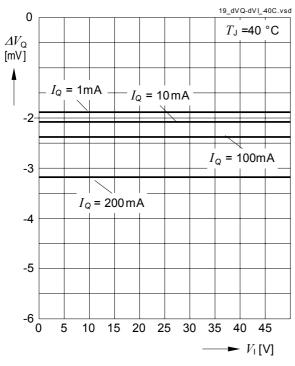


Line Regulation ΔV_{Q} versus Input Voltage Changed V_{I}



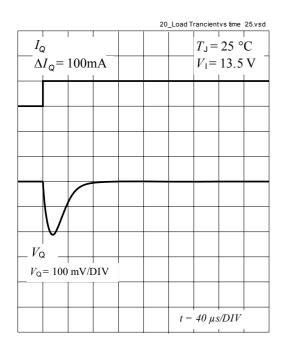
Data Sheet



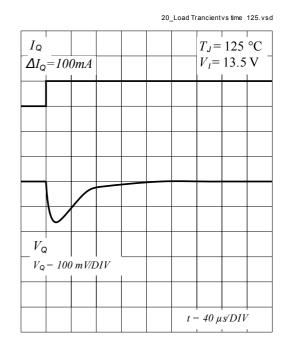


Line Regulation ΔV_{Q} versus Input Voltage Change V_{I}

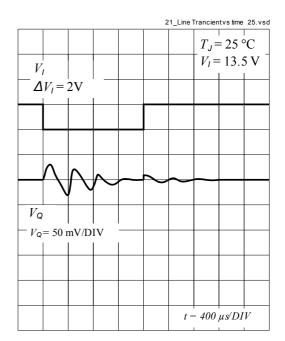
Load Transient Response Peak Voltage ΔV_{Q}



Load Transient Response Peak Voltage $\Delta V_{\rm Q}$



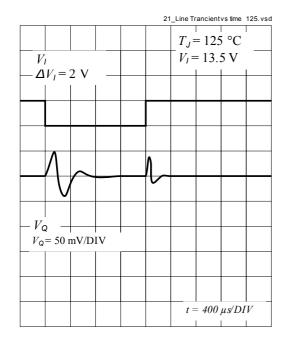
Line Transient Response Peak Voltage ΔV_{Q}





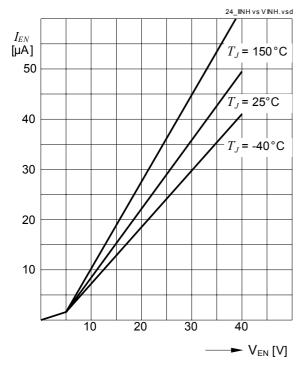
I

Typical Performance Characteristics

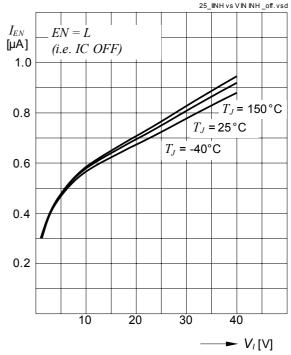


Line Transient Response Peak Voltage $\Delta V_{ m Q}$

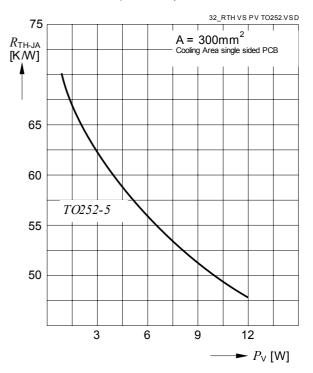
Enabled Input Current $I_{\rm EN}$ versus Enabled Input Voltage $V_{\rm EN}$



Enabled Input Current $I_{\rm EN}$ versus Input Voltage $V_{\rm I}$, EN=Off



Thermal Resistance Junction-Ambient R_{THJA} versus Power Dissipation P_{V}





IFX24401

Application Information

6 Application Information

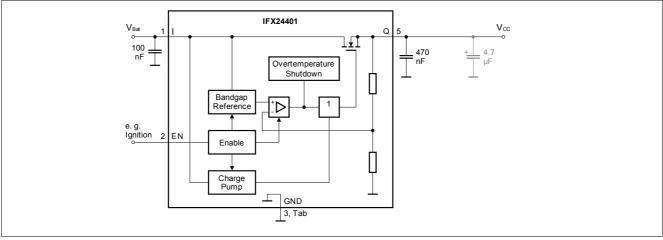


Figure 4 Application Diagram

Input, Output

An input capacitor is necessary for damping line influences. A resistor of approx. 1 Ω in series with $C_{\rm p}$, can damp the LC of the input inductivity and the input capacitor.

The IFX24401 requires a ceramic output capacitor of at least 470 nF. In order to damp influences resulting from load current surges it is recommended to add an additional electrolytic capacitor of 4.7 μ F to 47 μ F at the output as shown in Figure 4.

Additionally a buffer capacitor C_B of > 10 μ F should be used for the output to suppress influences from load surges to the voltage levels. This one can either be an aluminum electrolytic capacitor or a tantalum capacitor following the application requirements.

A general recommendation is to keep the drop over the equivalent serial resistor (ESR) together with the discharge of the blocking capacitor below the allowed Headroom of the Application to be supplied (e.g. typ. $dV_Q = 350$ mV).

Since the regulator output current roughly rises linearly with time the discharge of the capacitor can be calculated as follows:

 $dVC_B = dI_Q^* dt/C_B$

The drop across the ESR calculates as:

 $dV_{ESR} = dI^*ESR$

To prevent a reset the following relationship must be fullfilled:

 $\mathrm{dV}_{\mathrm{C}} + \mathrm{dV}_{\mathrm{ESR}} < \mathrm{V}_{\mathrm{RH}} = 350 \mathrm{mV}$

Example: Assuming a load current change of $dI_Q = 100$ mA, a blocking capacitor of $C_B = 22\mu$ F and a typical regulator reaction time under normal operating conditions of dt ~ 25µs and for special dynamic load conditions, such as load step from very low base load, a reaction time of dt ~ 75µs.

$$dV_{c} = dI_{Q}*dt/C_{B} = 100mA * 25\mu s/22\mu F = 113mV$$

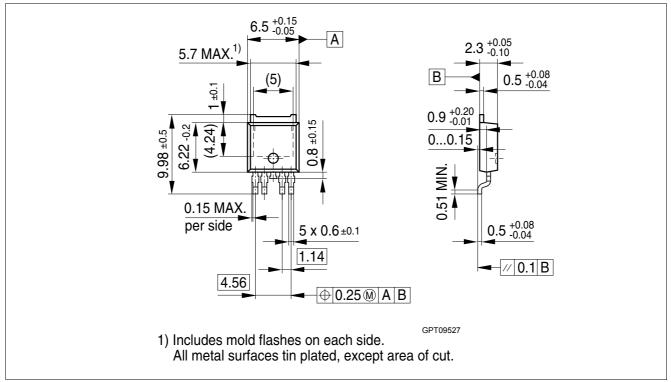
So for the ESR we can allow $dV_{ESR} = V_{RH2} - dV_{C} = 350 \text{mV} - 113 \text{mV} = 236 \text{mV}$

The permissible ESR becomes: ESR = dV_{ESR} / dI_{O} = 236mV/100mA = 2.36Ohm



Package Outlines

7 Package Outlines





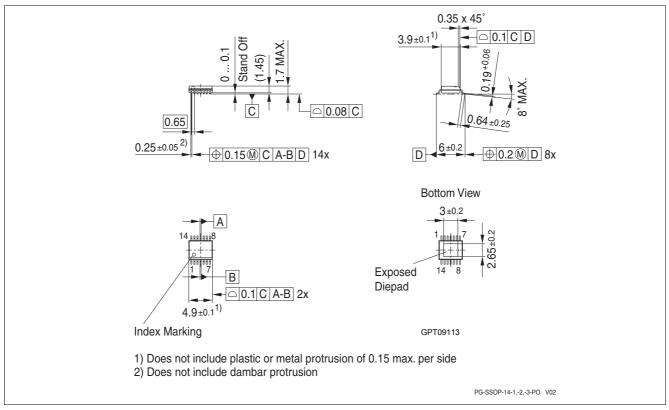


Figure 6 PG-SSOP-14



Revision History

8 Revision History

Revision	Date	Changes
1.02	2009-12-10	Corrections to pin assignment
1.01	2009-10-19	Coverpage changed Overview page: Inserted reference statement to TLE/TLF series.
1.0	2009-04-28	Initial Release

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