











TPS22929D

SLVSB39A - DECEMBER 2011-REVISED JUNE 2014

# TPS22929D Ultra-Small, Low on Resistance Load Switch With Controlled Turn-on

#### **Features**

- Integrated Single Load Switch
- Small SOT23-6 Package
- Input Voltage Range: 1.4 V to 5.5 V
- Low ON-Resistance
  - r<sub>ON</sub> = 115 mΩ at VIN = 5 V
  - $r_{ON} = 115 \text{ m}\Omega$  at VIN = 3.3 V
  - $r_{ON} = 118 \text{ m}\Omega \text{ at VIN} = 2.5 \text{ V}$
  - $r_{ON} = 129 \text{ m}\Omega \text{ at VIN} = 1.5 \text{ V}$
- 1.8-A Continuous Switch Current (25°C)
- Low Threshold Control Input
- Controlled Slew-rate
- Under-Voltage Lock Out
- Quick Output Discharge
- **Reverse Current Protection**

## **Applications**

- Portable Industrial Equipment
- Portable Medical Equipment
- Portable Media Players
- Point Of Sales Terminal
- **GPS Devices**
- **Digital Cameras**
- Portable Instrumentation
- **Smartphones**

## 3 Description

The TPS22929D is a small, low r<sub>ON</sub> load switch with controlled turn on. The device contains a P-channel MOSFET that can operate over an input voltage range of 1.4 V to 5.5 V. The switch is controlled by an on/off input (ON), which is capable of interfacing directly with low-voltage control signals. TPS22929D is active high enable.

The TPS22929D contains a 150- $\Omega$  on-chip load resistor for quick output discharge when the switch is turned off. The rise time of the device is internally controlled in order to avoid inrush current.

The TPS22929D device provides circuit breaker functionality by latching off the power-switch during reverse voltage situations. An internal reverse voltage comparator disables the power-switch when the output voltage (V<sub>OUT</sub>) is driven higher than the input (V<sub>IN</sub>) to quickly (10 µs typ) stop the flow of current towards the input side of the switch. Reverse current is always active, even when the power-switch is disabled. Additionally, under-voltage lockout (UVLO) protection turns the switch off if the input voltage is too low.

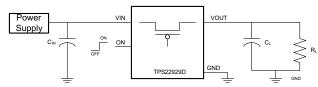
The TPS22929D is available in a small, space-saving 6-pin SOT23 package and is characterized for operation over the free-air temperature range of -40°C to 85°C.

#### Device Information<sup>(1)</sup>

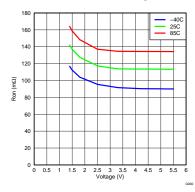
PART NUMBER	PACKAGE	BODY SIZE (NOM)
TPS22929D	SOT23 (6)	2.90 mm × 1.60 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.

# Simplified Schematic



#### On-State Resistance vs Input Voltage





# **Table of Contents**

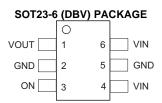
1	Features 1		9.1 Overview	12
2	Applications 1		9.2 Functional Block Diagram	
3	Description 1		9.3 Feature Description	12
4	Simplified Schematic		9.4 Device Functional Modes	13
5	Revision History2	10	Application and Implementation	14
6	Pin Configuration and Functions		10.1 Application Information	14
7	_		10.2 Typical Application	14
′	Specifications	11	Power Supply Recommendations	18
	7.1 Absolute Maximum Ratings	12	Layout	18
	7.2 Handling Ratings		12.1 Layout Guidelines	18
	7.3 Recommended Operating Conditions		12.2 Layout Example	19
	7.5 Electrical Characteristics	13	Device and Documentation Support	
			13.1 Trademarks	
	7.6 Switching Characteristics 6 7.7 Typical Characteristics 7		13.2 Electrostatic Discharge Caution	19
	Parametric Measurement Information		13.3 Glossary	
8		14	Mechanical, Packaging, and Orderable	
9	Detailed Description 12		Information	19

# 5 Revision History

CI	anges from Original (December 2011) to Revision A		
•	Updated to enhanced datasheet standards	1	
•	Added Handling Ratings table.	4	
•	Added Thermal Information table	4	
•	Added Detailed Description section.	12	
•	Added Application and Implementation section.	14	
•	Added Power Supply Recommendations section.	18	
•	Added Layout section.	18	



# 6 Pin Configuration and Functions



## **Pin Functions**

PIN		DESCRIPTION
NAME	DBV	DESCRIPTION
GND	2, 5	Ground
ON	3	Switch control input, active high. Do not leave floating
VOUT	1	Switch output
VIN	4, 6	Switch input, bypass this input with a ceramic capacitor to ground

# 7 Specifications

## 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
V <sub>IN</sub>	Input voltage range	-0.3	6	V
V <sub>OUT</sub>	Output voltage range	-0.3	6	V
V <sub>ON</sub>	Input voltage range	-0.3	6	V
	Maximum continuous power dissipation at 25°C		463	
$P_{MAX}$	Maximum continuous power dissipation at 70°C		254	mW
	Maximum continuous power dissipation at 85°C		185	
I <sub>MAX</sub>	Maximum continuous operating current		2	Α
T <sub>A</sub>	Operating free-air temperature range	-40	85	°C
T <sub>J</sub>	Maximum junction temperature		125	°C

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## 7.2 Handling Ratings

			MIN	MAX	UNIT
T <sub>stg</sub>	Storage temperature rang	ge	-65	150	°C
V	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	-2	2	14) /
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins (2)	-1	1	kV

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. Pins listed as 2 kV may actually have higher performance.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process. Pins listed as 1 kV may actually have higher performance.

## 7.3 Recommended Operating Conditions

			MIN	MAX	UNIT
V <sub>IN</sub>	Input voltage range		1.4	5.5	V
V <sub>ON</sub>	ON voltage range		0	5.5	V
V <sub>OUT</sub>	Output voltage range			$V_{IN}$	
$V_{IH}$	High-level input voltage, ON	VIN = 1.4 V to 5.5 V	1.1	5.5	V
.,	Law law line of walks and ON	VIN = 3.61 V to 5.5 V		0.6	V
$V_{IL}$	Low-level input voltage, ON	VIN = 1.4 V to 3.6 V		0.4	V
C <sub>IN</sub>	Input Capacitor		1 (1)		μF

<sup>(1)</sup> Refer to the application section.

#### 7.4 Thermal Information

		TPS22929D	
	THERMAL METRIC <sup>(1)</sup>	DBV	UNITS
		6 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	216	
$R_{\theta JCtop}$	Junction-to-case (top) thermal resistance	209	
$R_{\theta JB}$	Junction-to-board thermal resistance	131	°C/W
ΨЈТ	Junction-to-top characterization parameter	52	
ΨЈВ	Junction-to-board characterization parameter	110	

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.



# 7.5 Electrical Characteristics

 $VIN = 1.4 \text{ V to } 5.5 \text{ V}, T_A = -40^{\circ}\text{C} \text{ to } 85^{\circ}\text{C} \text{ (unless otherwise noted)}$ 

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	MIN	TYP	MAX	UNIT
		$I_{OUT} = 0$ , $V_{IN} = V_{ON} = 5.25 \text{ V}$			2.2	10	
		$I_{OUT} = 0$ , $V_{IN} = V_{ON} = 4.2 \text{ V}$			2.1	7.0	
I <sub>IN</sub>	Quiescent current	$I_{OUT} = 0$ , $V_{IN} = V_{ON} = 3.6 \text{ V}$	Full		2.0	7.0	μΑ
		$I_{OUT} = 0$ , $V_{IN} = V_{ON} = 2.5 \text{ V}$			1.0	5.0	
		$I_{OUT} = 0$ , $V_{IN} = V_{ON} = 1.5 \text{ V}$		·	0.8	5.0	
		$V_{ON} = GND$ , $V_{OUT} = Open$ , $V_{IN} = 5.25 \text{ V}$			0.8	10	
		$V_{ON} = GND$ , $V_{OUT} = Open$ , $V_{IN} = 4.2 \text{ V}$			0.3	7.0	
$I_{IN(off)}$	Off supply current	$V_{ON} = GND$ , $V_{OUT} = Open$ , $V_{IN} = 3.6 V$	Full		0.2	7.0	μΑ
		$V_{ON} = GND$ , $V_{OUT} = Open$ , $V_{IN} = 2.5 V$		·	0.2	5.0	
		$V_{ON} = GND$ , $V_{OUT} = Open$ , $V_{IN} = 1.5 V$			0.1	5.0	
		$V_{ON} = GND, V_{OUT} = 0, V_{IN} = 5.25 V$			0.8	10	
		$V_{ON} = GND$ , $V_{OUT} = 0$ , $V_{IN} = 4.2 \text{ V}$			0.3	7.0	μА
I <sub>IN(Leakage)</sub>	Leakage current	$V_{ON} = GND$ , $V_{OUT} = 0$ , $V_{IN} = 3.6 V$	Full		0.2	7.0	
		$V_{ON} = GND$ , $V_{OUT} = 0$ , $V_{IN} = 2.5 \text{ V}$			0.2	5.0	
		$V_{ON} = GND$ , $V_{OUT} = 0$ , $V_{IN} = 1.5 V$			0.1	5.0	
		V - 5 25 V I - 200 mA	25°C		115	150	mΩ
		$V_{IN} = 5.25 \text{ V}, I_{OUT} = -200 \text{ mA}$	Full			175	
		$V_{IN} = 5.0 \text{ V}, I_{OUT} = -200 \text{ mA}$	25°C		115	150	
		V <sub>IN</sub> = 5.0 V, I <sub>OUT</sub> = -200 IIIA	Full			175	
		V <sub>IN</sub> = 4.2 V, I <sub>OUT</sub> = -200 mA	25°C		115	150	
			Full			175	
r <sub>ON</sub>	On-resistance	V = 3.3 V I = 200 mA	25°C		115	150	11122
		$V_{IN} = 3.3 \text{ V}, I_{OUT} = -200 \text{ mA}$	Full			175	
		V - 2.5.V I - 200 mA	25°C		118	155	
		$V_{IN} = 2.5 \text{ V}, I_{OUT} = -200 \text{ mA}$	Full			180	
		V - 15 V I - 200 mA	25°C		129	170	
		$V_{IN} = 1.5 \text{ V}, I_{OUT} = -200 \text{ mA}$	Full			200	
RPD	Output pull down resistance	$V_{IN} = 3.3 \text{ V}, V_{ON} = 0, I_{OUT} = 30 \text{ mA}$	25°C		150	200	Ω
UVLO	Under voltage lockout	$V_{IN}$ increasing, $V_{ON} = 3.6 \text{ V}$ , $I_{OUT} = -100 \text{ mA}$	Full			1.4	V
	-	$V_{IN}$ decreasing, $V_{ON}$ 3.6 V, $R_L$ = 10 $\Omega$		0.50			
I <sub>ON</sub>	ON input leakage current	V <sub>ON</sub> = 1.4 V to 5.25 V or GND	Full	<del></del> ,		1	μΑ
V <sub>RVP</sub>	Reverse Current Voltage Threshold				77		mV
t <sub>DELAY</sub>	Reverse Current Response Delay	V <sub>IN</sub> = 5V			10		μs

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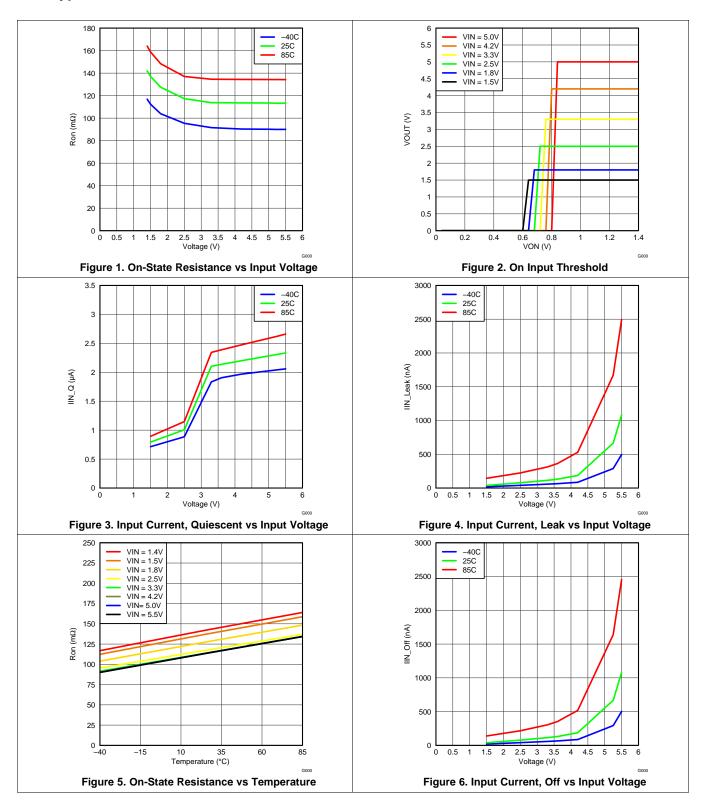


# 7.6 Switching Characteristics

	DADAMETED	TEGT COMPITION	TPS22929D		
PARAMETER		TEST CONDITION	TYP	UNIT	
VIN = 5	5 V, T <sub>A</sub> = 25°C (unless otherwise noted	i)			
t <sub>ON</sub>	Turn-ON time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	3315		
t <sub>OFF</sub>	Turn-OFF time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	7.4		
t <sub>R</sub>	VOUT rise time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	3660	μs	
t <sub>F</sub>	VOUT fall time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	6.1		
VIN = 3	3.3 V, T <sub>A</sub> = 25°C (unless otherwise not	ed)	•		
t <sub>ON</sub>	Turn-ON time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	4655		
t <sub>OFF</sub>	Turn-OFF time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	9.5		
t <sub>R</sub>	VOUT rise time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	4150	μs	
t <sub>F</sub>	VOUT fall time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	3.0		
VIN = 1	1.5 V, T <sub>A</sub> = 25°C (unless otherwise not	ed)	•		
t <sub>ON</sub>	Turn-ON time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	10175		
t <sub>OFF</sub>	Turn-OFF time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	18.3		
t <sub>R</sub>	VOUT rise time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	7812	μs	
t <sub>F</sub>	VOUT fall time	$R_L = 10 \Omega, C_L = 0.1 \mu F$	3.0		

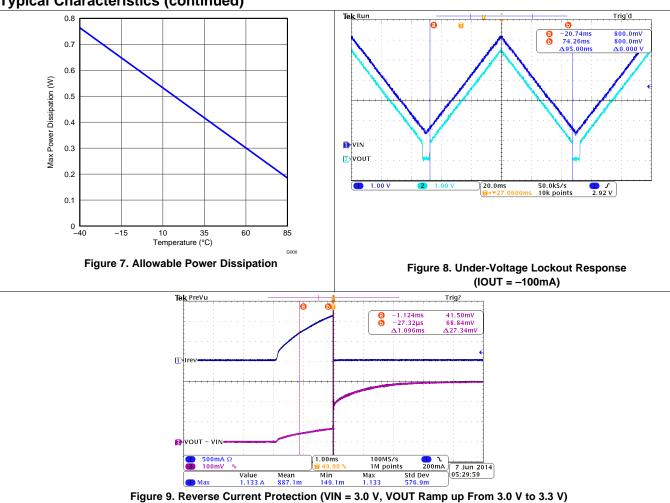


## 7.7 Typical Characteristics



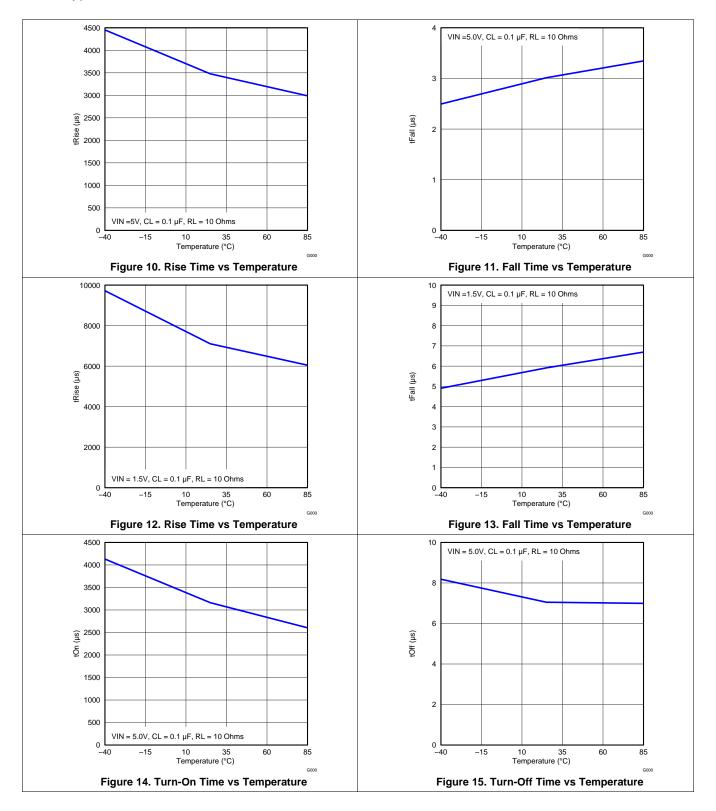


## **Typical Characteristics (continued)**



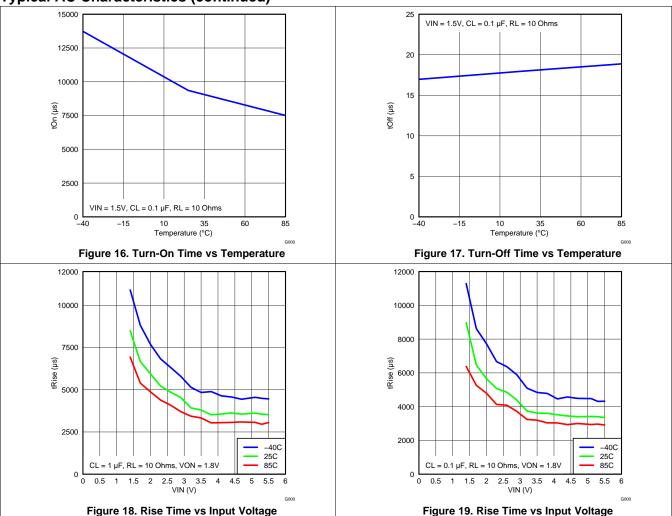


## 7.7.1 Typical AC Characteristics

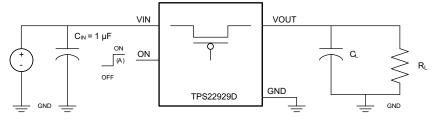




## **Typical AC Characteristics (continued)**



## 8 Parametric Measurement Information



A. Rise and fall times of the control signal is 100 ns.

Figure 20. Test Circuit



# **Parametric Measurement Information (continued)**

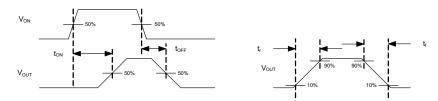


Figure 21.  $T_{ON}/T_{OFF}$  Waveforms



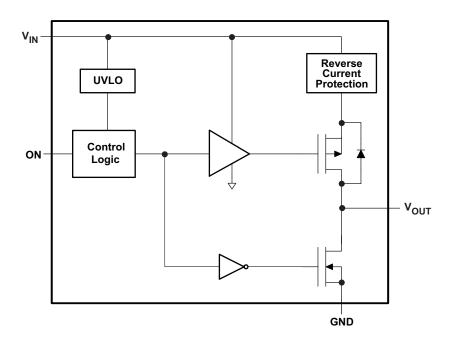
#### 9 Detailed Description

#### 9.1 Overview

The TPS22929D is a single channel, 1.8-A load switch in a small, space-saving 6-pin SOT23-6 package. This devices implement a low resistance P-channel MOSFET with a controlled rise time for applications that need to limit the inrush current.

This device is also designed to have very low leakage current during off state. This prevents downstream circuits from pulling high standby current from the supply. Integrated control logic, driver, power supply, and output discharge FET eliminates the need for additional external components, which reduces solution size and bill of materials (BOM) count.

#### 9.2 Functional Block Diagram



#### 9.3 Feature Description

#### 9.3.1 On/Off Control

The ON pin controls the state of the switch. Asserting ON high enables the switch. ON is active high and has a low threshold, making it capable of interfacing with low-voltage signals. The ON pin is compatible with standard GPIO logic threshold. It can be used with any microcontroller with 1.2-V, 1.8-V, 2.5-V or 3.3-V GPIOs.

#### 9.3.2 Output Pull-Down

The output pull-down is active when the user is turning off the main pass FET. The pull-down discharges the output rail to approximately 10% of the rail, and then the output pull-down is automatically disconnected to optimize the shutdown current.

#### 9.3.3 Under-Voltage Lockout

The under-voltage lockout turns-off the switch if the input voltage drops below the under-voltage lockout threshold. During under-voltage lockout (UVLO), if the voltage level at  $V_{OUT}$  exceeds the voltage level at  $V_{IN}$  by the Reverse Current Voltage Threshold ( $V_{RVP}$ ), the body diode will be disengaged to prevent any current flow to  $V_{IN}$ . With the ON pin active the input voltage rising above the under-voltage lockout threshold will cause a controlled turn-on of the switch which limits current over-shoots.



## **Feature Description (continued)**

## 9.3.4 Reverse Current Protection

In a scenario where  $V_{OUT}$  is greater than  $V_{IN}$ , there is potential for reverse current through the pass FET or the body diode. The TPS22929D monitors  $V_{IN}$  and  $V_{OUT}$  voltage levels. When the reverse current voltage threshold  $(V_{RVP})$  is exceeded, the switch is disabled (within 10  $\mu$ s typ). Additionally, the body diode is disengaged so as to prevent any reverse current flow to  $V_{IN}$ . The FET, and the output  $(V_{OUT})$ , will resume normal operation when the reverse voltage scenario is no longer present.

Use the following formula to calculate the amount of reverse current required to activate the protection circuit for a particular application:

$$I_{RC} = \frac{0.077V}{R_{ON(VIN)}}$$

Where,

IRC is the amount of reverse current,

R<sub>ON(VIN)</sub> is the on-resistance as determined by the input voltage.

## 9.4 Device Functional Modes

**Table 1. Function Table** 

ON	VIN to VOUT	VOUT to GND <sup>(1)</sup>
L	OFF	ON
Н	ON	OFF

(1) See Output Pull-Down section.



## 10 Application and Implementation

#### 10.1 Application Information

#### 10.1.1 VIN to VOUT Voltage Drop

The VIN to VOUT voltage drop in the device is determined by the  $R_{ON}$  of the device and the load current. The  $R_{ON}$  of the device depends upon the VIN condition of the device. Refer to the  $R_{ON}$  specification of the device in the Electrical Characteristics table of this datasheet. Once the  $R_{ON}$  of the device is determined based upon the VIN conditions, use Equation 1 to calculate the VIN to VOUT voltage drop:

$$\Delta V = I_{LOAD} \times R_{ON} \tag{1}$$

Where.

 $\Delta V$  = Voltage drop from VIN to VOUT

 $I_{LOAD}$  = Load current

 $R_{ON}$  = On-resistance of the device for a specific  $V_{IN}$ 

An appropriate I<sub>LOAD</sub> must be chosen such that the I<sub>MAX</sub> specification of the device is not violated.

#### 10.1.2 Input Capacitor

To limit the voltage drop on the input supply caused by transient inrush currents, when the switch turns on into a discharged load capacitor or short-circuit, a capacitor needs to be placed between VIN and GND. A 1-µF ceramic capacitor, CIN, placed close to the pins is usually sufficient. Higher values of CIN can be used to further reduce the voltage drop.

#### 10.1.3 Output Capacitor

A C<sub>IN</sub> to C<sub>L</sub> ratio of 10 to 1 is recommended for minimizing V<sub>IN</sub> dip caused by inrush currents during startup.

## 10.2 Typical Application

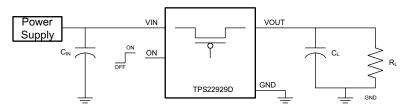


Figure 22. Typical Application Schematic

## 10.2.1 Design Requirements

Design Parameter	Example Value
VIN	1.5 V to 5 V
CL	0.1 µF to 1 µF
Maximum Acceptable Inrush Current	10 mA



#### 10.2.2 Detailed Design Procedure

#### 10.2.2.1 Managing Inrush Current

When the switch is enabled, the output capacitors must be charged up from 0-V to VIN voltage. This charge arrives in the form of inrush current. Inrush current can be calculated using the following equation:

Inrush Current = 
$$C \times \frac{dv}{dt}$$
 (2)

Where,

C = Output capacitance

d٧

dt = Output slew rate

The TPS22929D offers a very slow controlled rise time for minimizing inrush current. This device can be selected based upon the maximum acceptable slew rate which can be calculated using the design requirements and the inrush current equation. An output capacitance of 1.0  $\mu$ F will be used since the amount of inrush increases with output capacitance:

$$10 \text{ mA} = 1.0 \text{ } \mu\text{F} \times \frac{\text{dv}}{\text{dt}} \tag{3}$$

$$\frac{dv}{dt} = 10 \text{ V/ms} \tag{4}$$

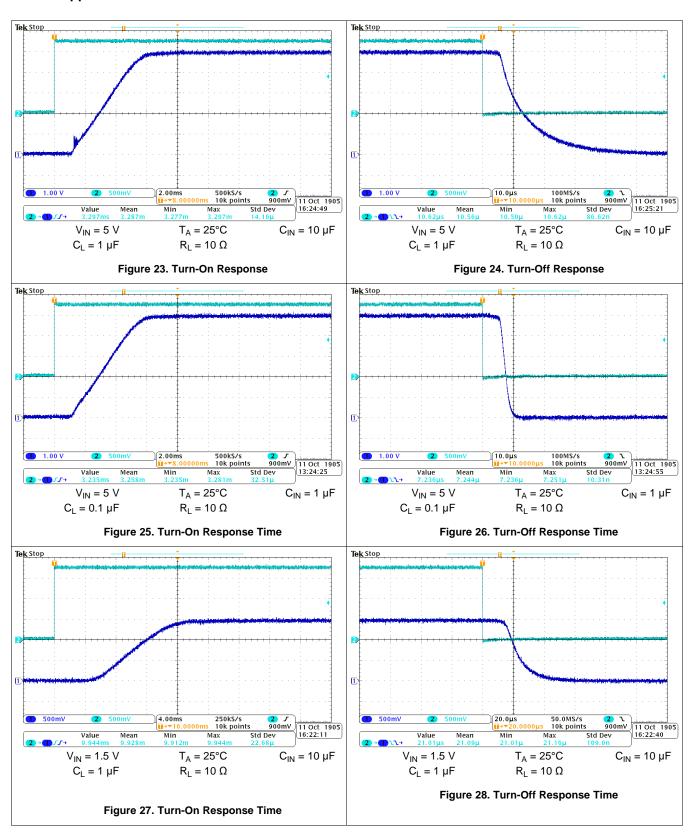
To ensure an inrush current of less than 10 mA, a device with a slew rate less than 10 V/ms must be used.

The TPS22929D has a typical rise time of 4500  $\mu s$  at 3.3 V . This results in a slew rate of 733 mV/ms which meets the above design requirements.

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# TEXAS INSTRUMENTS

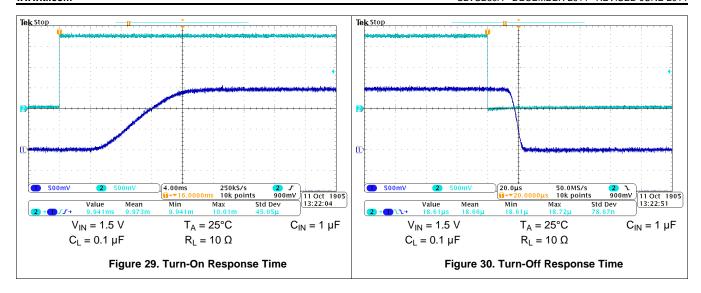
#### 10.2.3 Application Curves



Submit Documentation Feedback

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## 11 Power Supply Recommendations

The device is designed to operate with a VIN range of 1.4 V to 5.5 V.

## 12 Layout

#### 12.1 Layout Guidelines

For best performance, all traces should be as short as possible. To be most effective, the input and output capacitors should be placed close to the device to minimize the effects that parasitic trace inductances may have on normal operation. Using wide traces for  $V_{IN}$ ,  $V_{OUT}$ , and GND helps minimize the parasitic electrical effects along with minimizing the case to ambient thermal impedance.

#### 12.1.1 Thermal Considerations

For best device performance, be sure to follow the thermal guidelines in the *Thermal Information* table. To calculate max allowable continuous current for your application for a specific  $V_{IN}$  and ambient temperature, use the following formula:

$$I_{MAX} = \sqrt{\frac{T_J - T_A}{\theta_{JA}}} \\ R_{ON}$$

Where:

I<sub>MAX</sub>= Max allowable continuous current

 $T_J$ = Max thermal junction temperature (125°C)

T<sub>A</sub>= Ambient temperature of the application

 $\theta_{JA}$ = Junction-to-air thermal impedance (216°C/W)

R<sub>ON</sub>= R<sub>ON</sub> at a specified input voltage VIN (see *Electrical Characteristics*)



#### 12.2 Layout Example

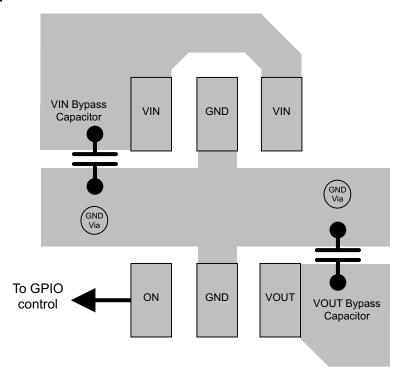


Figure 31. Layout Drawing

## 13 Device and Documentation Support

#### 13.1 Trademarks

All trademarks are the property of their respective owners.

## 13.2 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## 13.3 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

## 14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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## PACKAGE OPTION ADDENDUM

6-Feb-2020

#### **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty		Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
TPS22929DDBVR	ACTIVE	SOT-23	DBV	6	3000	Green (RoHS & no Sb/Br)	NIPDAU	Level-1-260C-UNLIM	-40 to 85	NF4F	Samples
TPS22929DDBVT	ACTIVE	SOT-23	DBV	6	250	Green (RoHS & no Sb/Br)	NIPDAU	Level-1-260C-UNLIM	-40 to 85	NF4F	Samples

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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6-Feb-2020

# **PACKAGE MATERIALS INFORMATION**

www.ti.com 24-Apr-2020

## TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS22929DDBVR	SOT-23	DBV	6	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TPS22929DDBVT	SOT-23	DBV	6	250	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3

# **PACKAGE MATERIALS INFORMATION**

www.ti.com 24-Apr-2020



#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins SPQ		Length (mm)	Width (mm)	Height (mm)	
TPS22929DDBVR	SOT-23	DBV	6	3000	180.0	180.0	18.0	
TPS22929DDBVT	SOT-23	DBV	6	250	180.0	180.0	18.0	



SMALL OUTLINE TRANSISTOR



#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.

- 4. Leads 1,2,3 may be wider than leads 4,5,6 for package orientation. 5. Refernce JEDEC MO-178.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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